

# Z8051 Series 8-Bit Microcontrollers

# Z51F0811

# **Product Specification**

PS029602-0212

PRELIMINARY



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# **Revision History**

Each instance in this document's revision history reflects a change from its previous edition. For more details, refer to the corresponding page(s) or appropriate links furnished in the table below.

	Revision		
Date	Level	Description	Page
Feb 2012	02	Removed references to SOP, LQFP packages.	All
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# Z51F0811

# CMOS SINGLE-CHIP 8-BIT MICROCONTROLLER WITH 12-BIT A/D CONVERTER

#### 1. Overview

#### **1.1 Description**

The Z51F0811 is advanced CMOS 8-bit microcontroller with 8K bytes of Flash. This is powerful microcontroller which provides a highly flexible and cost effective solution to many embedded control applications. This provides the following features: 8K bytes of Flash, 256 bytes of internal SRAM, 256 bytes of external SRAM, 512 bytes of Data EEPROM, general purpose I/O, 8/16-bit timer/counter, watchdog timer, watch timer, SPI, USART, I2C, on-chip POR, BOD, 12-bit A/D converter, analog comparator, buzzer driving port, 10-bit high speed PWM output, on-chip oscillator and clock circuitry. The Z51F0811 also supports power saving modes to reduce power consumption.

Device Name	Flash	EEPROM	RAM	ADC	I/O Port	Package			
	11 9K bytes 512 bytes	9K butos 512 butos	bytes 512 bytes 256				15ch	30	32-pin QFN
Z51F0811				256 bytes	12ch	26	28-pin TSSOP		
2311 0011	on bytes	512 bytes		012 59100		200 Dytes	9ch	18	20-pin TSSOP
					8ch	14	16-pin TSSOP		

#### 1.2 Features

• CPU

- 8 Bit CISC Core (8051 Compatible,2 clock per cycle)

- 8K Bytes On-chip Flash
- Optional boot code section with protection
- Endurance : 10,000 times at room temp
- Retention : 10 years
- 256 Bytes SRAM
- 256 Bytes XRAM
- 512 Bytes Data EEPROM
- Endurance : 300,000 times at room temp
- Retention : 10 years
- General Purpose I/O
- 30 Ports (P0[7:0], P1[6:0], P2[6:0], P3[7:0]): 32 Pin
- 26 Ports (P0[7:0], P1[6:0], P2[2:0], P3[7:0]): 28 Pin
- 18 Ports (P0[7:0], P1[6:0], P2[2:0]): 20 Pin
- 14 Ports (P0[7:0], P1[2:0], P2[2:0]): 16 Pin
- One Basic Interval Timer
- Timer/ Counter
- -8Bitx4ch(16Bitx2ch) + 16Bitx1ch
- 3 High Frequency 10-bit PWM (Using Timer1)
- 10-bit PWM (Using Timer3)
- Watch Dog Timer
- Watch Timer
- SPI
- USART (2ch)
- I2C
- Buzzer Driving Port
- 12 Bit A/D Converter
- 15 Input channels
- Analog Comparator
- On Chip Analog Comparator
- Interrupt Sources

- External (8)
- Pin Change Interrupt(P0) (1)
- USART0,1 (4)
- SPI (1)
- Timer (5)
- I2C (1)
- Data EEPROM (1)
- ADC (1)
- Analog Comparator(1)
- WDT (1)
- WT (1)
- BIT (1)
- On-Chip RC-Oscillator
- -8MHz(±3%)
- Power On Reset
- Programmable Brown-Out Detector
- Minimum Instruction Execution Time
- 200ns (@10MHz, NOP Instruction)
- Power down mode
- IDLE, STOP1, STOP2 mode
- Sub-Active mode
- System used external 32.768KHz crystal or system used internal 125KHz Ring oscillator
- Operating Frequency
- 1MHz ~ 12MHz
- Operating Voltage
- 1.8V ~ 5.5V
- Operating Temperature : -40 ~ +85℃
- Package Type
- 32-pin QFN
- 28-pin TSSOP
- 20-pin TSSOP
- 16-pin TSSOP
- Pb free package

#### **1.3 Ordering Information**

Device name	ROM size	RAM size	EEPROM size	Package
Z51F0811QUX				32 QFN
Z51F0811RJX	8K bytes Flash		510 by too	28 TSSOP
Z51F0811RHX		256 bytes	512 bytes	20 TSSOP
Z51F0811RFX				16 TSSOP

Table 1-1 Ordering Information for Z51F0811 parts

#### 1.3.1 Part Number Suffix Designation

Zilog part numbers consist of a number of components, as indicated in the following example.

**Example:** Part number Z51F0811RFX is an 8-bit MCU with 8 KB of Flash memory and 512 bytes of RAM in a 16-pin TSSOP package and operating within a -40°C to +85°C temperature range. In accordance with RoHS standards, this device has been built using lead-free solder.



#### **1.4 Development Tools**

#### 1.4.1 Compiler

We do not provide the compiler. Please contact third parties.

The Z51F0811 core is Mentor 8051. Anyway, device ROM size is smaller than 64KB. Developer can use all kinds of third party's standard 8051 compiler.

#### 1.4.2 OCD emulator and debugger

The OCD (On Chip Debug) emulator supports Zilogs 8051 series MCU emulation.

The OCD interface uses two wires interfacing between PC and MCU which is attached to user's system. The OCD can read or change the value of MCU internal memory and I/O peripherals. And also the OCD controls MCU internal debugging logic, it means OCD controls emulation, step run, monitoring, etc.

The OCD Debugger program works on Microsoft-Windows NT, 2000, XP, Vista (32bit) operating system.

If you want to see more details, please refer OCD debugger manual. You can download debugger S/W and manual from our web-site.

Connection:

- SCLK (Z51F0811 P06 port)
- SDATA (Z51F0811 P07 port)

#### 1.4.3 Programmer

Single programmer:

PGMplus USB: It programs MCU device directly.



Figure 1-1 Single Programmer

OCD emulator: It can write code in MCU device too.

Because of, OCD debugging supports ISP (In System Programming).

It does not require additional H/W, except developer's target system.

Note) If you produce semiconductor and measure the stop current, use OCD ISP

Gang programmer:

It programs 8 MCU devices at once. So, it is mainly used in mass production line.

Gang programmer is standalone type, it means it does not require host PC.

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### 2. Block Diagram



Figure 2-1 Z51F0811 block diagram

### 3. Pin Assignment



Figure 3-1 Z51F0811 32 QFN Pin assignment



Figure 3-2 Z51F0811 28 TSSOP Pin assignment

- NAME



Figure 3-3 Z51F0811 20 TSSOP Pin assignment



Figure 3-4 Z51F0811 16 TSSOP Pin assignment





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SEATING PLANE

SIDE VIEW

- ALL DIMENSIONS ARE IN MILLIMETERS.
  . XXX HATCH AREA IS SOLDERABLE EXPOSED PAD.
- 5. MAXIMUM ALLOWABLE BURR IS 0.050mm IN ALL DIRECTIONS.



Figure 4-1 32 pin QFN package



Figure 4-2 28 pin TSSOP package





Figure 4-3 20 pin TSSOP package





Figure 4-4 16 pin TSSOP package

## 5. Pin Description

Table 5-1 Normal pin description

PIN Name	I/O	Function	@RESET	Shared with
P00		Port P0	it units software port ister it units software port it units software port it units software port it units software port it units software port it units software port it units software port it units software port port port port port port port port	Avref / AN0 / SS0
P01		8-Bit I/O Port		AN1 / SCK0 / ACK0
P02		Can be set in input or output mode in 1-bit units Internal pull-up register can be used via software		AN2 / T2O / MOSI0 / TxD0
P03		when this port is used as input port	Land	AN3 / EC2 / MISO0 / RxD0
P04	I/O	Open Drain enable register can be used via software when this port is used as output port	Input	SXIN / AC- / AN4
P05		AN0~AN7 can be selected by ADCM register		SXOUT / AC+ / AN5
P06				DSCL / ACOUT/ AN6/ T0O / SCL
P07				DSDA / AN7/ EC0 / SDA
P10		Port P1		PWM1AA / T1O / INT0
P11		7-Bit I/O Port		PWM1AB / INT1
P12		Can be set in input or output mode in 1-bit units Internal pull-up register can be used via software		PWM1BA / BUZ / INT2
P13	I/O	when this port is used as input port	Input	PWM1BB
P14		Open Drain enable register can be used via software when this port is used as output port		PWM1CA
P15		AN8, AN9 can be selected by ADCM register		AN8 / PWM1CB
P16				AN9 / PWM3 / T3O / INT3
P20		Port P2		XIN
P21		7-Bit I/O Port		XOUT
P22		Can be set in input or output mode in 1-bit units Internal pull-up register can be used via software		RESETB
P23	I/O	when this port is used as input port	Input	AN12
P24		Open Drain enable register can be used via software when this port is used as output port		AN13
P25		AN12, AN13, AN14 can be selected by ADCM		AN14 / (SCL)
P26		register		(SDA)
P30		Port P3		INT4
P31		8-Bit I/O Port		INT5
P32		Can be set in input or output mode in 1-bit units		INT6
P33		Internal pull-up register can be used via software when this port is used as input port		INT7
P34	I/O	Open Drain enable register can be used via software when this port is used as output port	Input	SS1
P35		AN10, AN11 can be selected by ADCM register		ACK1
P36				AN10 / TxD1
P37				AN11 / RxD1

#### 6. Port Structures

#### 6.1 General Purpose I/O Port



Figure 6-1 General Purpose I/O Port

#### 6.2 External Interrupt I/O Port



Figure 6-2 External Interrupt I/O Port

### 7. Electrical Characteristics

#### 7.1 Absolute Maximum Ratings

Table 7-1 Absolute Maximum Rationgs

Parameter	Symbol	Rating	Unit
O marke ) (alta na	VDD	-0.3~+6.5	V
Supply Voltage	VSS	-0.3~+0.3	V
	VI	-0.3~VDD+0.3	V
	VO	-0.3~VDD+0.3	V
	IOH	10	mA
Normal Voltage Pin	ΣΙΟΗ	80	mA
	IOL	20	mA
	ΣIOL	160	mA
Total Power Dissipation	PT	600	mW
Storage Temperature	TSTG	-45~+125	°C

Note) Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at any other conditions beyond those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

#### 7.2 Recommended Operating Conditions

Parameter	Symbol	Condition MIN		TYP	MAX	Unit
		fXIN=1~12MHz	4.5			
Supply Voltage	VDD	fXIN=1~8MHz	10	-	5.5	V
		fSUB=32.768KHz	1.8		85 12 -	
Operating Temperature	TOPR	VDD=1.8~5.5V -40 - 85		85	°C	
	FOPR	fXIN	1	-	12	MHz
Operating Frequency		fSUB	-	32.768	-	KHz
Operating Frequency		Internal RC-OSC	7.76		8.24	MHz
		Internal Ring-OSC	-	1		MHz

Table 7-2 Recommended Operation Conditions

### 7.3 A/D Converter Characteristics

Table 7-3 A/D Converter	Characteristics
-------------------------	-----------------

(TA=-40℃ ~ +85℃, VDD=AVDD=2.7V ~ 5.5V, VSS=0V)

Parameter	Symbol	Condition	MIN	TYP	MAX	Unit
Resolution		-	-	12	-	bits
Total Accuracy				-	±3	lsb
Integral Linear Error	INL		-	-	±2	lsb
Differential Linearity Error	DLE	AVDD=VDD=5.12V fXIN=4MHz	-	-	±2	lsb
Zero Offset Error	ZOE		-		±3	lsb
Full Scale Error	FSE		-		±3	lsb
Conversion Time	tCON	12bit conversion max 3MHz	-	60	-	cycle
Analog Input Voltage	VAN	-	VSS	-	AVDD=VDD	V
Analog Power Voltage	AVDD	-	-	*AVDD=VDD	-	V
Analog Reference Voltage	AVREF	-	2.7	-	5.5	V
Analog Ground Voltage	AVSS	-	-	VSS	-	V
Analog Input Leakage Current		AVDD=VDD=5.12V	-	-	10	uA
	IDD		-	1	3	mA
ADC Operating Current	SIDD	AVDD=VDD=5.12V	-	-	1	uA

### 7.4 Analog Comparator Characteristics

Table 7-4 Analog Comparator Characteristics

Parameter	Symbol	Condition	MIN	TYP	MAX	Unit
Input Leakage Current	ΙL	VDDEXT=5V, Vin=1/2VDDEXT	-50	-	50	nA
Input Offset Voltage	V <sub>offset</sub>	VDDEXT=5V, Vin=1/2VDD	10	-	40	±mV
Operating Current	IOP	COMP_EN=H	-	1	-	mA
Power Down Current	IPD	COMP_EN=L	-	1	-	uA
Response Time	V <sub>RT</sub>	CL= 50pF, VDDEXT=5V	-	-	500	nS

### 7.5 Voltage Dropout Converter Characteristics

Table 7-5 Voltage Dropout Converter Characteristics

Parameter	Symbol	Condition	MIN	TYP	MAX	Unit
Operating Voltage		-	1.8	-	5.5	V
Operating Temperature		-	-40	-	+85	°C
Regulation Voltage		-	1.62	1.8	1.98	V
Drop-out Voltage		-	-	-	0.02	V
- · •		RUN/IDLE	-	20	-	mA
		SUB-ACTIVE	-	1	-	mA
Current Drivability		STOP1	-	50	-	uA
		STOP2	-    1.8    -    5.5    N      -    -40    -    +85    *      -    1.62    1.8    1.98    N      -    1.62    1.8    1.98    N      -    -    -    0.02    N      VIDLE    -    20    -    m      ACTIVE    -    1    -    m      FOP1    -    50    -    u      VIDLE    -    10    -    u      VIDLE    -    10    -    u      FOP2    -    10    -    u      VIDLE    -    -    1    m      ACTIVE    -    -    0.1    m      FOP2    -    -    0.1    m      FOP2    -    -    0.1    u      FOP2    -    -    0.1    u      FOP2    -    -    1    u	uA		
	IDD1	RUN/IDLE	-	-	1	mA
On exerting Comment	IDD2	SUB-ACTIVE	-	-	0.1	mA
Operating Current	SIDD1	STOP1	-	-	5	uA
	SIDD2	STOP2	-	-	5.5 +85 1.98 0.02 - - - 1 0.1 5 0.1 1	uA
Drivebility Transition Time	TRAN1	SUB to RUN	-	-	1	uS
Drivability Transition Time	TRAN2	-    1.8    -    5.      -    -40    -    +8      -    1.62    1.8    1.9      -    1.62    1.8    1.9      -    -    -    0.0      RUN/IDLE    -    20    -      SUB-ACTIVE    -    1    -      STOP1    -    50    -      STOP2    -    10    -      RUN/IDLE    -    -    1      STOP2    -    10    -      SUB-ACTIVE    -    0.0    -      STOP2    -    10    -      SUB-ACTIVE    -    -    0.0      SUB to RUN    -    -    1	200	uS		

Note) -STOP1: WDT running - STOP2: WDT disable

#### 7.6 Power-On Reset Characteristics

Table 7-6 Power-On Reset Characteristics

Parameter	Symbol	Condition	MIN	TYP	MAX	Unit
Operating Voltage		-	VSS	-	5.5	V
Operating Temperature		-	-40	-	+85	°C
RESET Release Level		-	1.3	1.4	1.5	V
On and the One and	IDD	-	-	-	10	uA
Operating Current	SIDD	-	-	-	1	uA

#### 7.7 Brown Out Detector Characteristics

Table 7-7 Brown Out Detector Characteristics

Parameter	Symbol	Condition	MIN	TYP	MAX	Unit
Operating Voltage		-	VSS	-	5.5	V
Operating Temperature		-	-40	-	+85	°C
	4.2V	-	4.0	-	4.4	V
Detection have	3.6V	-	3.4	-	3.8	V
Detection Level	2.5V	-	2.3	-	2.7	V
	1.6V	-	1.4	-	1.8	V
Hysteresis		-	-	-	-	mV
Operating Current	IDD	-	-	-	50	uA
Operating Current	SIDD	-	-	-	1	uA

#### 7.8 Internal RC Oscillator Characteristics

Table 7-8 Internal RC Oscillator Characteristics

Parameter	Symbol	Condition	MIN	TYP	MAX	Unit
Operating Voltage		-	1.8	-	5.5	V
Operating Temperature		-	-40	-	+85	°C
Frequency		-	7.76		8.24	MHz
Stabilization Time		-	-	-	10	mS
	IDD	-	-	-	-	uA
Operating Current	SIDD	-	-	-	1	uA

#### 7.9 Ring-Oscillator Characteristics

Table 7-9 Ring-Oscillator Characteristics

Parameter	Symbol	Condition	MIN	TYP	MAX	Unit
Operating Voltage		-	1.8	-	5.5	V
Operating Temperature		-	-40	-	+85	°C
Frequency		-	-	1	-	MHz
Stabilization Time		-	-	-	-	mS
Operating Current	IDD	-	-	-	-	uA
	SIDD	-	-	-	1	uA

#### 7.10 PLL Characteristics

Table 7-10 PLL Characteristics

Parameter	Symbol	Condition	Min.	Тур.	Max.	Unit
PLL current	IPLL	-	-	1.5	TBD	mA
Input clock frequency	fxin	-	2	-	16	MHz
Output clock frequency	fout	-	6.25	-	128	MHz
Output clock duty	-	-	40	-	60	%
Setting time	tD	-	-	1	-	mS
Accuracy	-	-	-	2	-	%

 $(TA = 0^{\circ}C \sim +70^{\circ}C, VDD18 = 1.6V \sim 2.0V, VSS = 0V)$ 

### 7.11 DC Characteristics

Table 7-11 DC Characteristics

Parameter	Symbol	Condition	MIN	TYP	MAX	Unit
Input Low Voltage	VIL1	P2[2]	-0.5	-	0.2VDD	V
	VIL2	All others PAD	-0.5	-	0.2VDD	V
Input High Voltage	VIH1	P2[2]	0.8VDD	-	VDD	V
	VIH2	All others PAD	0.7VDD	-	VDD	V
Output Low Voltage	VOL1	ALL I/O (IOL=20mA, VDD=4.5V)	-	-	1	V
Output High Voltage	VOH1	ALL I/O (IOH=-8.57mA, VDD=4.5V)	3.5	-	-	V
Input High Leakage Current	ШΗ	ALL PAD	-	-	1	uA
Input Low Leakage Current	IIL	ALL PAD	-1	-	-	uA
Pull-Up Resister	RPU	ALL PAD	20	-	50	kΩ
Power Supply Current	IDD1	Run Mode, fXIN=12MHz @5V	-	*2.6	10	mA
	IDD2	Sleep Mode, fXIN=12MHz @5V	-	*1.5	5	mA
	IDD3	Sub Active Mode, fSUBXIN=32.768KHz @5V	-	*71	500	uA
	IDD4	STOP1 Mode, WDT Active @5V (BOD enable)	-	*45	200	uA
	IDD5	STOP1 Mode, WDT Active @5V (BOD disable)	-	*20	100	uA
	IDD6	STOP2 Mode, WDT Disable @5V (BOD enable)	-	*27	100	uA
	IDD7	STOP2 Mode, WDT Disable @5V (BOD disable)	-	*1	7 (room temp)	uA

(VDD =2.7~5.5V, VSS =0V, fXIN=10.0MHz, TA=-40~+85℃)

Note) - STOP1: WDT running, STOP2: WDT disable.

- (\*) typical test condition : VDD=5V, Internal RC-OSC=8MHz, ROOM TEMP, all PORT output LOW, Timer0 Active, 1PORT toggling.
# 7.12 AC Characteristics

Table 7-12 AC Characteristics

<sup>(</sup>VDD=5.0V±10%, VSS=0V, TA=-40~+85℃)

Parameter	Symbol	PIN	MIN	TYP	MAX	Unit
Operating Frequency	fMCP	XIN	1	-	10	MHz
System Clock Cycle Time	tSYS	-	100	-	1000	ns
Oscillation Stabilization Time (8MHz)	tMST1	XIN, XOUT	-	-	10	ms
External Clock "H" or "L" Pulse Width	tCPW	XIN	90	-	-	ns
External Clock Transition Time	tRCP,tFCP	XIN	-	-	10	ns
External Interrupt Input Width	tIW	INT0~INTx	2	-	-	tSYS
External Interrupt Transition Time	tFI,tRI	INT0~INTx			1	us
nRESET Input Pulse "L" Width	tRST	nRESET	8	-	-	tSYS
External Counter Input "H" or "L" Pulse Width	tECW	EC0~ECx	2	-	-	tSYS
Event Counter Transition Time	tREC,tFEC	EC0~ECx	-	-	20	ns



Figure 7-1 AC Timing

# 7.13 SPI Characteristics

Table 7-13 SPI Characteristics

Parameter	Symbol	PIN	MIN	TYP	MAX	Unit
Output Clock Pulse Period	tSCK	SCK	-	SPI clock mode	-	ns
Input Clock Pulse Period	tSCK	SCK	2• tSYS	-	-	ns
Input Clock "H" or "L" Pulse Width	tSCKL, tSCKH	SCK		50% duty	-	ns
Input Clock Pulse Transition Time	tFSCK,tRSCK	SCK	-	-	30	ns
Output Clock "H" or "L" Pulse Width	tSCKL, tSCKH	SCK	tSYS-30	-	-	ns
Output Clock Pulse Transition Time	tFSCK,tRSCK	SCK	-	-	30	ns
First Output Clock Delays Time	tFOD	OUTPUT				
Output Clock Delay Time	tDS	OUTPUT	-	-	100	ns
Input Pulse Transition Time	tFSIN,tRSIN	INPUT	-	-	30	ns
Input Setup Time	tDIS	INPUT	100		-	ns
Input Hold Time	tDIH	INPUT	tSYS+70	-	-	ns

(VDD=5.0V±10%, VSS=0V, TA=-40~+85℃)



Figure 7-2 SPI Timing

# 7.14 Typical Characteristics

These graphs and tables provided in this section are for design guidance only and are not tested or guaranteed. In some graphs or tables the data presented are outside specified operating range (e.g. outside specified VDD range). This is for information only and devices are guaranteed to operate properly only within the specified range.

The data presented in this section is a statistical summary of data collected on units from different lots over a period of time. "Typical" represents the mean of the distribution while "max" or "min" represents (mean +  $3\sigma$ ) and (mean -  $3\sigma$ ) respectively where  $\sigma$  is standard deviation.



Figure 7-3 8MHz Internal OSC Freq.(OSC\_HOT: 85°C)

# 8. Memory

The Z51F0811 addresses two separate address memory stores: Program memory and Data memory. The logical separation of Program and Data memory allows Data memory to be assessed by 8-bit addresses, which can be more quickly stored and manipulated by 8-bit CPU. Nevertheless, 16-bit Data memory addresses can also be generated through the DPTR register.

Program memory can only be read, not written to. There can be up to 64K bytes of Program memory. In the Z51F0811 Flash version of these devices the 8K bytes of Program memory are provided onchip. Data memory can be read and written to up to 256 bytes internal memory (DATA) including the stack area.

# 8.1 Program Memory

A 16-bit program counter is capable of addressing up to 64K bytes, but this device has just 8K bytes program memory space.

Figure 8-1 shows a map of the lower part of the program memory. After reset, the CPU begins execution from location 0000H. Each interrupt is assigned a fixed location in program memory. The interrupt causes the CPU to jump to that location, where it commences execution of the service routine. External interrupt 0, for example, is assigned to location 0003H. If external interrupt 0 is going to be used, its service routine must begin at location 0003H. If the interrupt is not going to be used, its service location is available as general purpose program memory. If an interrupt service routine is short enough (as is often the case in control applications), it can reside entirely within that 8 byte interval. Longer service routines can use a jump instruction to skip over subsequent interrupt locations, if other interrupts are in use.

Z51F0811 Product Specification



Figure 8-1 Program memory

- User Function Mode: 8KBytes Included Interrupt Vector Region
- Non-volatile and reprogramming memory: Flash memory based on EEPROM cell

# 8.2 Data Memory



Figure 8-2 shows the internal Data memory space available.

Figure 8-2 Data memory map

The internal memory space is divided into three blocks, which are generally referred to as the lower 128, upper 128, and SFR space.

Internal Data memory addresses are always one byte wide, which implies an address space of only 256 bytes. However, the addressing modes for internal RAM can in fact accommodate 384 bytes, using a simple trick. Direct addresses higher than 7FH access one memory space and indirect addresses higher than 7FH access a different memory space. Thus Figure 8-2 shows the upper 128 and SFR space occupying the same block of addresses, 80H through FFH, although they are physically separate entities.

The lower 128 bytes of RAM are present in all 8051 devices as mapped in Figure 8-3. The lowest 32 bytes are grouped into 4 banks of 8 registers. Program instructions call out these registers as R0 through R7. Two bits in the Program Status Word select which register bank is in use. This allows more efficient used of code space, since register instructions are shorter than instructions that use direct addressing.

The next 16 bytes above the register banks form a block of bit-addressable memory space. The 8051 instruction set includes a wide selection of single-bit instructions, and the 128 bits in this area can be directly addressed by these instructions. The bit addresses in this area are 00H through 7FH.

All of the bytes in the lower 128 can be accessed by either direct or indirect addressing. The upper 128 bytes RAM can only be accessed by indirect addressing. These spaces are used for user RAM and stack pointer.





Figure 8-3 Lower 128 bytes RAM

# 8.3 EEPROM Data Memory

Z51F0811 has 512 bytes EEPROM Data memory. This area has no relation with RAM/Flash. It can read and write through SFR with 8-bit unit.

For more information about EEPROM Data memory, see chapter 15.



Figure 8-4 XDATA memory area

# 8.4 SFR Map

### 8.4.1 SFR Map Summary

Table 8-1 SFR Map Summary

	0H/8H	1H/9H	2H/AH	3H/BH	4H/CH	5H/DH	6H/EH	7H/FH
2F58H	-	FUSE_PKG	FUSE_CAL2	FUSE_CAL1	FUSE_CAL0	FUSE_CON F	TEST_REG B	TEST_REG A
2F50H	PSR0	PSR1	-	-	-	-	-	-
2F48H	-	-	-	-	-	-	-	-
2F40H	-	-	-	-	-	-	-	-
2F38H	-	-	-	-	-	-	-	-
2F30H	-		-	-	-	-	-	-
2F28H	-	-	-	-	-	-	-	-
2F20H	-	-	-	-	-	-	-	-
2F18H	P0DB	P1DB	P2DB	P3DB	-	-	-	-
2F10H	-	-	-	-	-	-	-	-
2F08H	-	-	-	-	P00D	P10D	P2OD	P3OD
2F00H	P0PU	P1PU	P2PU	P3PU	-	-	-	-
			,					
F8H	IP1	ACCSR	UCTRL11	UCTRL12	UCTRL13	USTAT1	UBAUD1	UDATA1
F0H	В	-	FEARL	FEARM	FEARH	FEDR	FETR	-
E8H	-	-	FEMR	FECR	FESR	FETCR	-	-
E0H	ACC	-	UCTRL1	UCTRL2	UCTRL3	USTAT	UBAUD	UDATA
D8H	-	-	I2CMR	I2CSR	I2CSCLLR	I2CSCLHR	I2CSDHR	I2CDR
D0H	PSW	-	SPICR	SPIDR	SPISR	T4H	I2CSAR1	I2CSAR
C8H	-	-	T3CR	T3DR / PWM3PR	T3 / PWM3DR / CDR3	PWM3HR	T4CR	T4L
C0H	-	-	T2DLYB	T1DLYC	T1ISR	T1IMSK	T2CR	T2 / T2DR / CDR2
B8H	IP	-	T1BDR	T1CDR	T1PHR	T1PCR2	T1PCR3	T1DLYA
B0H	-	-	TOCR	T0 / T0DR / CDR0	T1CR	T1DR /T1PPR	T1 / T1ADR / CDR1	T1PCR
A8H	IE	IE1	IE2	IE3	IE4	IE5	PCI0	TMISR
A0H	-	P3IO	EO	EIENAB	EIFLAG	EIEDGE	EIPOLA	EIBOTH
98H	P3	P2IO	ADCM	ADCM2 / ADCRH	ADCRL	WTMR	WTR / WTCR	BUZCR
90H	P2	P1IO	-	-	-	-	-	-
88H	P1	P0IO	SCCR	BCCR	BITR	WDTMR	WDTR / WDTCR	BUZDR
80H	P0	SP	DPL	DPH	-	PLLCR	BODR	PCON

Note: 1) The registers of which lower 3-bit address are 000 are bit-addressable (except for XSFR)

### 8.4.2 Compiler Compatible SFR

# ACC (Accumulator) : E0H

Z51F0811 Product Specification



DPH Data Pointer High Byte

### PSW (Program Status Word) : D0H

7	6	5	4	3	2	1	0
CY	AC	F0	RS1	RS0	OV	F1	Р
RW	RW	RW	RW	RW	RW	RW	RW
						I	nitial value : 00

- CY Carry Flag
- AC Auxiliary Carry Flag
- **F0** General Purpose User-Definable Flag
- RS1 Register Bank Select bit 1
- RS0 Register Bank Select bit 0
- OV Overflow Flag

0

1

1

1

1

- F1 User-Definable Flag
- P Parity Flag. Set/cleared by hardware each instruction cycle to indicate an odd/even number of '1' bits in the accumulator

### EO (Extended Operation Register) : A2H

	7	6	5	4	3	2	1	0	
	-	-	-	TRAP_EN	-	DPSEL2	DPSEL1	DPSEL0	i i
	R	R	R	RW	R	RW	RW	RW	
							I	nitial value : 0	ЭН
TRAP_EN				Select the instr	uction				

	0 S	elect MOVC	@(DPTR++)	), A				
	1 S	elect Softwa	re TRAP inst	ruction				
DPSEL[2:0]	Select Bar	Select Banked Data Point Register						
	DPSEL2	DPSEL1	DPSEL0					
	0	0	0	DPTR0				
	0	0	1	DPTR1				
	0	1	0	DPTR2				

1

0

1

0

1

DPTR3

DPTR4

DPTR5

DPTR6

DPTR7

1

0

0

1

1

# 9. I/O Ports

# 9.1 I/O Ports

The Z51F0811 has four I/O ports (P0, P1, P2 and P3). Each port can be easily configured by software as I/O pin, internal pull up and open drain pin to meet various system configurations and design requirements. Also P0 includes function that can generate interrupt according to change of state of the pin.

# 9.2 Port Register

# 9.2.1 Data Register (Px)

Data Register is a bidirectional I/O port. If ports are configured as output ports, data can be written to the corresponding bit of the Px. If ports are configured as input ports, the data can be read from the corresponding bit of the Px.

# 9.2.2 Direction Register (PxIO)

Each I/O pin can independently used as an input or an output through the PxIO register. Bits cleared in this read/write register will select the corresponding pin in Px to become an input, setting a bit sets the pin to output. All bits are cleared by a system reset.

# 9.2.3 Pull-up Resistor Selection Register (PxPU)

The on-chip pull-up resistor can be connected to them in 1-bit units with a pull-up resistor selection register (PxPU). The pull-up register selection controls the pull-up resister enable/disable of each port. When the corresponding bit is 1, the pull-up resister of the pin is enabled. When 0, the pull-up resister is disabled. All bits are cleared by a system reset. (Only port pull-up resistor selection have default ON state for unused pins in 32-pin package for 16, 20, 28-pin package).

# 9.2.4 Open-drain Selection Register (PxOD)

There is internally open-drain selection register (PxOD) in P0, P1, P2 and P3. The open-drain selection register controls the open-drain enable/disable of each port. Ports become push-pull by a system reset.

# 9.2.5 Debounce Enable Register (PxDB)

P0, P1, P2 and P3 support debounce function. Debounce time of each ports has about 5us, but if P2[2] uses external reset function, it has about 7us debounce time. (except P2[2], other port initialization state is OFF)

# 9.2.6 Pin Change Interrupt Enable Register (PCI0)

The P0 can support Pin Change Interrupt function. Pin Change Interrupts PCI will trigger if any enabled P0[7:0] pin toggles. The PCI0 Register control which pins contribute to the pin change interrupts.

### 9.2.7 Port Selection Register (PSRx)

PSRx registers prevent the input leakage current when ports are connected to analog inputs. If the bit of PSRx is '1', the dynamic current path of the schmitt OR gate of the port is cut off and the digital input of the corresponding port is always '1'.

### 9.2.8 Register Map

Table	9-1	Register	Map
rubio	<b>U</b> 1	regiotor	inap

Name	Address	Dir	Default	Description
P0	80H	R/W	00H	P0 Data Register
P0IO	89H	R/W	00H	P0 Direction Register
POPU	2F00H	R/W	00H	P0 Pull-up Resistor Selection Register
P0OD	2F0CH	R/W	00H	P0 Open-drain Selection Register
P0DB	2F18H	R/W	00H	P0 Debounce Enable Register
PCI0	AEH	R/W	00H	P0 Pin Change Interrupt Enable Register
P1	88H	R/W	00H	P1 Data Register
P1IO	91H	R/W	00H	P1 Direction Register
P1PU	2F01H	R/W	00H	P1 Pull-up Resistor Selection Register
P10D	2F0DH	R/W	00H	P1 Open-drain Selection Register
P1DB	2F19H	R/W	00H	P1 Debounce Enable Register
P2	90H	R/W	00H	P2 Data Register
P2IO	99H	R/W	00H	P2 Direction Register
P2PU	2F02H	R/W	00H	P2 Pull-up Resistor Selection Register
P2OD	2F0EH	R/W	00H	P2 Open-drain Selection Register
P2DB	2F1AH	R/W	00H	P2 Debounce Enable Register
P3	98H	R/W	00H	P3 Data Register
P3IO	A1H	R/W	00H	P3 Direction Register
P3PU	2F03H	R/W	00H	P3 Pull-up Resistor Selection Register
P3OD	2F0FH	R/W	00H	P3 Open-drain Selection Register
P3DB	2F1BH	R/W	00H	P3 Debounce Enable Register
PSR0	2F50H	R/W	00H	Port Selection Register 0
PSR1	2F51H	R/W	00H	Port Selection Register 1,2,3

# 9.3 Px Port

### 9.3.1 Px Port Description

Px is 8-bit I/O port. Px control registers consist of Data register (Px), direction register (PxIO), debounce enable register (PxDB), pull-up register selection register (PxPU), open-drain selection register (PxOD), pin change interrupt register (PCIO)

### 9.3.2 Register description for Px

### Px (Px Data Register) : 80H, 88H, 90H, 98H



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| RW                 |
|----|----|----|----|----|----|----|--------------------|
|    |    |    |    |    |    |    | nitial value : 00H |

Px[7:0] I/O Data

## PxIO (Px Direction Register) : 89H, 91H, 99H, A1H

7	6	5	4	3	2	1	0	_
Px710	Px6lO	Px5IO	Px4IO	Px3IO	Px210	Px1IO	Px0IO	
RW								
						I	nitial value : 0	ΟH

PxIO[7:0] Px data I/O direction.

0 Input

1 Output

### PxPU (Px Pull-up Resistor Selection Register) : 2F00H ~ 2F03H

7	6	5	4	3	2	1	0	
Px7PU	Px6PU	Px5PU	Px4PU	Px3PU	Px2PU	Px1PU	Px0PU	ĺ
RW								
						I	nitial value : 0	OН

PxPU[7:0] Configure pull-up resistor of Px port

0 Disable

1 Enable

### PxOD (Px Open-drain Selection Register) : 2F0CH ~ 2F0FH

7	6	5	4	3	2	1	0
Px70D	Px6OD	Px5OD	Px4OD	Px3OD	Px2OD	Px10D	Px0OD
RW	RW	RW	RW	RW	RW	RW	RW
	nitial value : 00H						
	P	xOD[7:0]	Configure ope	en-drain of Px	port		
			0 Disat	ble			
				le			

### PxDB (Px Debounce Enable Register) : 2F18H ~ 2F1BH



# PCI0 (P0 Pin Change Interrupt Enable Register) : AEH

7	6	5	4	3	2	1	0
PC107	PC106	PC105	PC104	PC103	PCI02	PC101	PCI00
RW							

Initial value : 00H

### PCI0[7:0] Configure Pin Change Interrupt of P0 port

0 Disable

1 Enable

### PSR0 (P0 Port Selection Register) : 2F50H

7	6	5	4	3	2	1	0
PSR07	PSR06	PSR05	PSR04	PSR03	PSR02	PSR01	PSR00
RW							
						I	nitial value : 00H

PSR0[7:0]

0] P07~P00 port selection register

0 Disable analog channel AN[7:0] (default)

1 Enable analog channel AN[7:0]

### PSR1 (Port Selection Register 1, 2, 3) : 2F51H

7	6	5	4	3	2	1	0
PSR17	PSR16	PSR15	PSR14	PSR13	PSR12	PSR11	PSR10
RW	RW	RW	RW	RW	RW	RW	RW
						I	nitial value : 00H
PSR1[7]			-	ction register 6] for I2C (def 5] for I2C	ault)		
	P		P25,P24,P23, 0 Disat	-	nnel AN[14:8]	0	

## 9.4 Port RESET Noise Canceller

The Figure 13-21 is the Noise canceller diagram for Noise cancel of Pore RESET. It has the Noise cancel value of about 5us ( $@V_{DD}=5V$ ) to input of Port Reset.



Figure 9-1 Port Reset noise canceller time diagram

# **10. Interrupt Controller**

### 10.1 Overview

The Z51F0811 supports up to 32 interrupt sources. The interrupts have separate enable register bits associated with them, allowing software control. They can also have four levels of priority assigned to them. The interrupt controller has following features:

- receive the request from 32 interrupt source
- 8 group priority
- 4 priority levels
- Multi Interrupt possibility
- If the requests of different priority levels are received simultaneously, the request of higher priority level is serviced
- Each interrupt source can control by EA bit and each IEx bit
- Interrupt latency: 5~8 machine cycles in single interrupt system

The maskable interrupts are enabled through six of interrupt enable registers (IE, IE1, IE2, IE3, IE4, IE5). Bits of IE, IE1, IE2, IE3, IE4, IE5 register each individually enable/disable a particular interrupt source. Overall control is provided by bit 7 of IE (EA). When EA is set to '0', all interrupts are disabled: when EA is set to '1', interrupts are individually enabled or disabled through the other bits of the interrupt enable registers. The Z51F0811 supports a four-level priority scheme. Each maskable interrupt is individually assigned to one of four priority levels by writing to IP or IP1.

Interrupt default mode is level-trigger basically but if needed, it is able to change edge-trigger mode. Table 10-1 shows the Interrupt Group Priority Level that is available for sharing interrupt priority. Priority sets two bit which is to IP and IP1 register about group. Interrupt service routine services higher priority. If two requests of different priority levels are received simultaneously, the request of higher priority level is serviced. If the request of same or lower priority level is received, that request is not serviced.

Interrupt Group	Highest			Lowest	
0 (Bit0)	Interrupt0	Interrupt8	Interrupt16	Interrupt24	Highest
1 (Bit1)	Interrupt1	Interrupt9	Interrupt17	Interrupt25	
2 (Bit2)	Interrupt2	Interrupt10	Interrupt18	Interrupt26	
3 (Bit3)	Interrupt3	Interrupt11	Interrupt19	Interrupt27	
4 (Bit4)	Interrupt4	Interrupt12	Interrupt20	Interrupt28	
5 (Bit5)	Interrupt5	Interrupt13	Interrupt21	Interrupt29	
6 (Bit6)	Interrupt6	Interrupt14	Interrupt22	Interrupt30	
7 (Bit7)	Interrupt7	Interrupt15	Interrupt23	Interrupt31	Lowest

# **10.2 External Interrupt**

The external interrupt on INT0, INT1, INT2, INT3, INT4, INT5, INT6 and INT7 pins receive various interrupt request depending on the EIEDGE (External Interrupt Edge register) and EIPOLA (External

Interrupt Polarity register) and EIBOTH(External Interrupt Both Edge register) as shown in Figure 10-1. Also each external interrupt source has control setting bits. The EIFLAG (External interrupt flag register) register provides the status of external interrupts.



Figure 10-1 External Interrupt Description

## 10.3 Block Diagram



Figure 10-2 Block Diagram of Interrupt

## **10.4 Interrupt Vector Table**

The interrupt controller supports 32 interrupt sources as shown in the Table 10-2 below. When interrupt becomes service, long call instruction (LCALL) is executed in the vector address. Interrupt request 32 has a decided priority order.

Interrupt Source	Symbol	Interrupt Enable Bit	Priority	Mask	Vector Address
Hardware Reset	RESETB	0	0	Non-Maskable	0000H
External Interrupt 0	INT0	IE0.0	1	Maskable	0003H
External Interrupt 1	INT1	IE0.1	2	Maskable	000BH
External Interrupt 2	INT2	IE0.2	3	Maskable	0013H
External Interrupt 3	INT3	IE0.3	4	Maskable	001BH
Pin Change Interrupt (P0)	INT4	IE0.4	5	Maskable	0023H
Reserved	INT5	-	6	-	002BH
USART0 Rx	INT6	IE1.0	7	Maskable	0033H
USART0Tx	INT7	IE1.1	8	Maskable	003BH
SPI0	INT8	IE1.2	9	Maskable	0043H
l <sup>2</sup> C	INT9	IE1.3	10	Maskable	004BH
USART1 Rx	INT10	IE1.4	11	Maskable	0053H
USART1 Tx	INT11	IE1.5	12	Maskable	005BH
TO	INT12	IE2.0	13	Maskable	0063H
T1	INT13	IE2.1	14	Maskable	006BH
T2	INT14	IE2.2	15	Maskable	0073H
T3	INT15	IE2.3	16	Maskable	007BH
T4	INT16	IE2.4	17	Maskable	0083H
EEPROM	INT17	IE2.5	18	Maskable	008BH
ADC	INT18	IE3.0	19	Maskable	0093H
Comparator	INT19	IE3.1	20	Maskable	009BH
ŴT	INT20	IE3.2	21	Maskable	00A3H
WDT	INT21	IE3.3	22	Maskable	00ABH
BIT	INT22	IE3.4	23	Maskable	00B3H
Reserved	INT23	-	24	-	00BBH
Reserved	INT24	-	25	-	00C3H
Reserved	INT25	-	26	-	00CBH
Reserved	INT26	-	27	-	00D3H
Reserved	INT27	-	28	-	00DBH
External Interrupt 4	INT28	IE4.4	29	Maskable	00E3H
External Interrupt 5	INT29	IE4.5	30	Maskable	00EBH
External Interrupt 6	INT30	IE5.0	31	Maskable	00F3H
External Interrupt 7	INT31	IE5.1	32	Maskable	00FBH

For maskable interrupt execution, first EA bit must set '1' and specific interrupt source must set '1' by writing a '1' to associated bit in the IEx. If interrupt request is received, specific interrupt request flag set '1'. And it remains '1' until CPU accepts interrupt. After that, interrupt request flag will be cleared automatically.

### 10.5 Interrupt Sequence

An interrupt request is held until the interrupt is accepted or the interrupt latch is cleared to '0' by a reset or an instruction. Interrupt acceptance always generates at last cycle of the instruction. So instead of fetching the current instruction, CPU executes internally LCALL instruction and saves the PC stack. For the interrupt service routine, the interrupt controller gives the address of LJMP instruction to CPU. After finishing the current instruction, at the next instruction to go interrupt service routine needs 5~8 machine cycle and the interrupt service task is terminated upon execution of an interrupt return instruction [RETI]. After generating interrupt, to go to interrupt service routine, the following process is progressed

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Figure 10-3 Interrupt Vector Address Table

# 10.6 Effective Timing after Controlling Interrupt bit



Figure 10-4 Interrupt Enable Register effective Timing

# 10.7 Multi Interrupt

If two requests of different priority levels are received simultaneously, the request of higher priority level is serviced. If requests of the interrupt are received at the same time simultaneously, an interrupt polling sequence determines by hardware which request is serviced. However, multiple processing through software for special features is possible.



Figure 10-5 Execution of Multi Interrupt

Following example is shown to service INT0 routine during INT1 routine in Figure 10-6. In this example, INT0 interrupt priority is higher than INT1 interrupt priority. If some interrupt is lower than INT1 priority, it can't service its interrupt routine.

Example) Software Multi Interrupt:

```
INT1:
MOV
         IE, #01H
                      ; Enable INT0 only
MOV
         IE1, #00H
                      : Disable others
SETB
         ΕA
                      ; Enable global interrupt (necessary for multi interrupt)
÷
MOV
         IE, #0FFH
                       ; Enable all Interrupts
MOV
         IE1, #0FFH
RETI
```

## **10.8 Interrupt Enable Accept Timing**



Figure 10-6 Interrupt Response Timing Diagram

# **10.9 Interrupt Service Routine Address**



Figure 10-7 Correspondence between vector Table address and the entry address of ISP

## 10.10 Saving/Restore General-Purpose Registers



Figure 10-8 Saving/Restore Process Diagram & Sample Source

# Interrupt sampled here CLP1 CLP2 CLP2 C1P1 C1P2 C2P1 C2P2 SCLK INT\_SRC INTR\_ACK LAST\_CYC INTR\_LCALL 8-Bit interrupt Vector INT\_VEC {8'h00, INT\_VEC} PROGA

# **10.11 Interrupt Timing**

Figure 10-9 Timing chart of Interrupt Acceptance and Interrupt Return Instruction

Interrupt source sampled at last cycle of the command. When sampling interrupt source, it is decided to low 8-bit of interrupt vector. M8051W core makes interrupt acknowledge at first cycle of command, executes long call to jump interrupt routine as INT\_VEC.

Note) command cycle C?P?: L=Last cycle, 1=1<sup>st</sup> cycle or 1<sup>st</sup> phase, 2=2<sup>nd</sup> cycle or 2<sup>nd</sup> phase

## 10.12 Interrupt Register Overview

## 10.12.1 Interrupt Enable Register (IE, IE1, IE2, IE3, IE4, IE5)

Interrupt enable register consists of Global interrupt control bit (EA) and peripheral interrupt control bits. Totally 32 peripheral are able to control interrupt.

## 10.12.2 Interrupt Priority Register (IP, IP1)

The 32 interrupt divides 8 groups which have each 4 interrupt sources. A group can decide 4 levels interrupt priority using interrupt priority register. Level 3 is the high priority, while level 0 is the low priority. Initially, IP, IP1 reset value is '0'. At that initialization, low interrupt number has a higher priority than high interrupt number. If decided the priority, low interrupt number has a higher priority than high interrupt number in that group.

# 10.12.3 External Interrupt Flag Register (EIFLAG)

The external interrupt flag register is set to '1' when the external interrupt generating condition is satisfied. The flag is cleared when the interrupt routine is executed. Alternatively, the flag can be cleared by writing a '0' to it.

### 10.12.4 External Interrupt Edge Register (EIEDGE)

The External interrupt edge register determines which type of edge or level sensitive interrupt. Initially, default value is level. For level, write '0' to related bit. For edge, write '1' to related bit.

### 10.12.5 External Interrupt Polarity Register (EIPOLA)

According to EIEDGE register, the external interrupt polarity (EIPOLA) register has a different meaning. If EIEDGE is level type, EIPOLA is able to have Low/High level value. If EIEGDE is edge type, EIPOLA is able to have rising/falling edge value.

### 10.12.6 External Interrupt Enable Register (EIENAB)

When the external interrupt enable register is written to '1', the corresponding external pin interrupt is enabled. The EIEDGE and EIPOLA register defines whether the external interrupt is activated on rising or falling edge or level sensed.

### 10.12.7 External Interrupt Both Edge Enable Register (EIBOTH)

When the external interrupt both edge enable register is written to '1', the corresponding external pin interrupt is enabled by both edges. Initially, default value is disabled.

### 10.12.8 Register Map

Name	Address	Dir	Default	Description
IE	A8H	R/W	00H	Interrupt Enable Register
IE1	A9H	R/W	00H	Interrupt Enable Register 1
IE2	ААН	R/W	00H	Interrupt Enable Register 2
IE3	ABH	R/W	00H	Interrupt Enable Register 3
IE4	ACH	R/W	00H	Interrupt Enable Register 4
IE5	ADH	R/W	00H	Interrupt Enable Register 5
IP	B8H	R/W	00H	Interrupt Priority Register
IP1	F8H	R/W	00H	Interrupt Priority Register 1
EIENAB	A3H	R/W	00H	External Interrupt Enable Register
EIFLAG	A4H	R/W	00H	External Interrupt Flag Register
EIEDGE	A5H	R/W	00H	External Interrupt Edge Register
EIPOLA	A6H	R/W	00H	External Interrupt Polarity Register
EIBOTH	A7H	R/W	00H	External Interrupt Both Edge Enable Register

Table 10-3 Register Map

# **10.13 Interrupt Register Description**

The Interrupt Register is used for controlling interrupt functions. Also it has External interrupt control registers. The interrupt register consists of Interrupt Enable Register (IE), Interrupt Enable Register 1 (IE1), Interrupt Enable Register 2 (IE2), Interrupt Enable Register 3 (IE3), Interrupt Enable Register 4 (IE4) and Interrupt Enable Register 5 (IE5). For external interrupt, it consists of External Interrupt Flag Register (EIFLAG), External Interrupt Edge Register (EIEDGE), External Interrupt Polarity Register (EIPOLA), External Interrupt Enable Register (EIENAB) and External Interrupt Both Edge Enable Register(EIBOTH).

### 10.13.1 Register description for Interrupt

7	6	5	4	3	2	1	0
EA	-	INT5E	INT4E	INT3E	INT2E	INT1E	INTOE
RW	-	RW	RW	RW	RW	RW	RW
							nitial value : 00
		EA	Enable or disa	ble all interrup	ot bits		
			0 All In	terrupt disable	•		
			1 All In	terrupt enable			
INT5E Reserved							
			0 Disat	ole			
			1 Enab	le			
		INT4E	Enable or disa	ble Pin Chang	ge Interrupt 0	(Port 0)	
			0 Disat	ole			
			1 Enab	le			
		INT3E	Enable or disa	ble External I	nterrupt 3		
			0 Disat	ole			
PS020602-0	212		PRELI	MINARY			E

# IE (Interrupt Enable Register) : A8H

INT2E Enable or disable External Interrupt 2 0 Disable 1 Enable	
	INT2E
1 Enable	
INT1E Enable or disable External Interrupt 1	INT1E
0 Disable	
1 Enable	
INTOE Enable or disable External Interrupt 0	INT0E
0 Disable	
1 Enable	

# IE1 (Interrupt Enable Register 1) : A9H

7	6	5	4	3	2	1	0
-	-	INT11E	INT10E	INT9E	INT8E	INT7E	INT6E
-	-	RW	RW	RW	RW	RW	RW
						I	nitial value : 00H
		INT11E	Enable or disa	ble USART1	Tx Interrupt		
			0 Disat	ole			
			1 Enab	le			
		INT10E	Enable or disa	ble USART1	Rx Interrupt		
			0 Disat	ole			
			1 Enab	le			
		INT9E	Enable or disa	ble I <sup>2</sup> C Interru	ıpt		
			0 Disat	ole			
			1 Enab	le			
		INT8E	Enable or disa	ble SPI0 Inter	rupt		
			0 Disat	ole			
			1 Enab	le			
		INT7E	Enable or disa	ble USART0	Tx Interrupt		
			0 Disat	ole			
			1 Enab	le			
		INT6E	Enable or disa	ble USART0	Rx Interrupt		
			0 Disat	ole			
			1 Enab	le			

# IE2 (Interrupt Enable Register 2) : AAH

7	6	5	4	3	2	1	0		
-	-	INT17E	INT16E	INT15E	INT14E	INT13E	INT12E		
-	-	RW	RW	RW	RW	RW	RW		
						I	nitial value : 00H		
INT17E Enable or disable EEPROM Interrupt									
			0 Disable						
			1 Enab	le					
		INT16E	Enable or disa	ble Timer 4 In	iterrupt				
			0 Disab	ole					
			1 Enab	le					
	l	INT15E	Enable or disable Timer 3 Interrupt						

	0	Disable
	1	Enable
INT14E	Enable	or disable Timer 2 Interrupt
	0	Disable
	1	Enable
INT13E	Enable	or disable Timer 1 Interrupt
	0	Disable
	1	Enable
INT12E	Enable	or disable Timer 0 Interrupt
	0	Disable
	1	Enable

# IE3 (Interrupt Enable Register 3) : ABH

7	6	5	4	3	2	1	0
-	-	INT23E	INT22E	INT21E	INT20E	INT19E	INT18E
R	R	RW	RW	RW	RW	RW	RW
						I	nitial value : 00

INT23E	Reserv	ved
	0	Disable
	1	Enable
INT22E	Enable	or disable BIT Interrupt
	0	Disable
	1	Enable
INT21E	Enable	or disable WDT Interrupt
	0	Disable
	1	Enable
INT20E	Enable	or disable WT Interrupt
	0	Disable
	1	Enable
INT19E	Enable	or disable Analog Comparator Interrupt
	0	Disable
	1	Enable
INT18E	Enable	or disable ADC Interrupt
	0	Disable
	1	Enable

# IE4 (Interrupt Enable Register 4) : ACH

7	6	5	4	3	2	1	0			
-	-	INT29E	INT28E	INT27E	INT26E	INT25E	INT24E			
R	R	RW	RW	RW	RW	RW	RW			
	Initial value : 00H									
INT29E Enable or disable External Interrupt 5										
			0 Disable							
			1 Enab	le						
		INT28E	Enable or disa	ble External Ir	nterrupt 4					
			0 Disable							
			1 Enab	le						

INT27E	Reserved			
	0	Disable		
	1	Enable		
INT26E	Reserv	red		
	0	Disable		
	1	Enable		
INT25E	Reserv	red		
	0	Disable		
	1	Enable		
INT24E	Reserv	red		
	0	Disable		
	1	Enable		

# IE5 (Interrupt Enable Register 5) : ADH

7	6	5		4	3	2	1	0
-	-	INT35E	IN	T34E	INT33E	INT32E	INT31E	INT30E
R	R	RW	F	RW .	RW	RW	RW	RW
							I	nitial value : 00ł
		INT35E	Reserv	ved				
			0	Disab	le			
			1	Enab	le			
		INT34E	Reserv	ved				
			0	Disab	le			
			1	Enab	le			
		INT33E	Reserv	ved				
			0	Disab	ole			
			1	Enab	le			
		INT32E	Reserv	ved				
			0	Disab	ole			
			1	Enab	le			
		INT31E	Enable	e or disa	ble External I	nterrupt 7		
			0	Disab	ole			
			1	Enab	le			
		INT30E	Enable	e or disa	ble External I	nterrupt 6		
			0	Disab	ole			
			1	enabl	е			

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### IP (Interrupt Priority Register) : B8H

7	6	5	4	3	2	1	0	
IP7	IP6	IP5	IP4	IP3	IP2	IP1	IP0	
RW								
						I	nitial value : 0	OН

### IP1 (Interrupt Priority Register 1) : F8H

7	6	5	4	3	2	1	0
IP17	IP16	IP15	IP14	IP13	IP12	IP11	IP10
RW							

Initial value : 00H

IP[7:0], IP1[7:0] Select Interrupt Group Priority

IP1x	IPx	Description
0	0	level 0 (lowest)
0	1	level 1
1	0	level 2
1	1	level 3 (highest)

### EIFLAG (External Interrupt Flag Register) : A4H

7	6	5	4	3	2	1	0
FLAG7	FLAG6	FLAG5	FLAG4	FLAG3	FLAG2	FLAG1	FLAG0
RW							
							altial contract 001

Initial value : 00H

**FLAG[7:0]** If External Interrupt is occurred, the flag becomes '1'. The flag can be cleared by writing a '0' to bit

0 External Interrupt not occurred

1 External Interrupt occurred

### EIEDGE (External Interrupt Edge Register) : A5H

7	6	5	4	3	2	1	0		
EDGE7	EDGE6	EDGE5	EDGE4	EDGE3	EDGE2	EDGE1	EDGE0		
RW	RW	RW	RW	RW	RW	RW	RW		
	Initial value : 00H								
	EI	DGE[7:0]	Determines v ur.	which type of e	edge or level s	sensitive interr	rupt may occ		
		(	D Level	(default)					
			1 Edge						

#### EIPOLA (External Interrupt Polarity Register) : A6H

7	6	5	4	3	2	1	0
POLA7	POLA6	POLA5	POLA4	POLA3	POLA2	POLA1	POLA0
RW							

Initial value : 00H

**POLA[7:0]** According to EIEDGE, External interrupt polarity register has a different means. If EIEDGE is level type, external interrupt polarity is able to have Low/High level value. If EIEGDE is edge type, external interrupt polarity is able to have rising/ falling edge value.



Level case:

- 0 When High level, Interrupt occurred (default)
- 1 When Low level, Interrupt occurred Edge case:
- 0 When Rising edge, Interrupt occurred (default)
- 1 When Falling edge, Interrupt occurred

### EIENAB (External Interrupt Enable Register) : A3H

7	6	5	4	3	2	1	0	
ENAB7	ENAB6	ENAB5	ENAB4	ENAB3	ENAB2	ENAB1	ENAB0	
RW								

Initial value : 00H

ENAB[7:0] Control External Interrupt

0 Disable (default)

1 Enable

### EIBOTH (External Interrupt Both Edge Enable Register) : A7H

7	6	5	4	3	2	1	0	
BOTH7	BOTH6	BOTH5	BOTH4	BOTH3	BOTH2	BOTH1	BOTH0	
RW	-							
						I	nitial value : 0	OH

**BOTH[7:0]** Determines which type of interrupt may occur, EIBOTH or EIEDGE+EIPOLA. if EIBOTH is enable, EIEDGE and EIPOLA register value don't matter.

0 Disable (default)

1 Enable

# 11. Peripheral Hardware

# **11.1 Clock Generator**

# 11.1.1 Overview

As shown in Figure 11-1, the clock generator produces the basic clock pulses which provide the system clock to be supplied to the CPU and the peripheral hardware. It contains main-frequency clock oscillator. The system clock operation can be easily obtained by attaching a crystal between the XIN and XOUT pin, respectively. The system clock can also be obtained from the external oscillator. In this case, it is necessary to put the external clock signal into the XIN pin and open the XOUT pin. The default system clock is INT-RC Oscillator and the default division rate is one. In order to stabilize system internally, use 1MHz RING oscillator for BIT, WDT and ports de-bounce.

- Calibrated Internal RC Oscillator (8 MHz)
  - . INT-RC OSC/1 (Default system clock)
  - . INT-RC OSC/2 (4 MHz)
  - . INT-RC OSC/4 (2 MHz)
  - . INT-RC OSC/8 (1 MHz)
- Crystal Oscillator (1~16 MHz)
- Sub-Clock Crystal Oscillator (32.768 KHz)



# 11.1.2 Block Diagram

Figure 11-1 Clock Generator Block Diagram

# 11.1.3 Register Map

Table 11-1 Register Map

Name	Address	Dir	Default	Description
SCCR	8AH	R/W	04H	System and Clock Control Register

### 11.1.4 Clock Generator Register description

The Clock Generation Register uses clock control for system operation. The clock generation consists of System and Clock register.

### 11.1.5 Register description for Clock Generator

### SCCR (System and Clock Control Register) : 8AH

7	6	5		4	3	2	1	0		
STOP1	DIV1	DIV0	C	BYS	ISTOP	XSTOP	CS1	CS0		
RW	RW	RW	F	W.	RW	RW	RW	RW		
								Initial value : 0		
		STOP1	Contro	I the ST	OP Mode					
			Note) set this		CON=0x03, It	is applied. Bu	t when PCON	l=0x01, don't		
			0 STOP2 Mode (at PCON=0x03) (default)							
			1	STOP	1 Mode (at P	CON=0x03)				
		DIV[1:0]		whenus		system clock, is system cloc				
			Note)	To chang	ge by softwar	e, CBYS set to	o '1'			
			DIV1	DIV0	description	า				
			0	0	fINTRC/1	(8MHz)				
			0	1	fINTRC/2	(4MHz)				
			1	0	fINTRC/4	(2MHz)				
			1	1	fINTRC/8	(1MHz)				
		CBYS	change is cont to '0', then w	e is cont rolled by it is not hen wak	rolled by hard y software. E: changed rigi æ-up, it applie	ck change. If dware. But if th x) when settin ht now, CPU es to clock cha ep other bits in	his set to '1', og CS[1:0], if goes to STO ange.	clock change CBYS bit set		
			0	Clock	changed by I	nardware durir	ng stop mode	(default)		
			1	Clock	changed by s	software				
		ISTOP			eration of INT 3YS='1', It is a	-RC Oscillatio	n			
			0	RC-O	scillation enal	ble (default)				
			1	RC-O	scillation disa	ble				
		XSTOP	Control the operation of X-Tal Oscillation Note1) when CBYS='1', It is applied Note2) if XINENA bit in FUSE_CONF to '0', XSTOP is fix					ixed to '1'		
			0	X-Tal	Oscillation er	nable				
			1	X-Tal	Oscillation dis	sable (default)				
020602-0	040									

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CS[1:0]	Determine System Clock Note) by CBYS bit, reflection point is decided					
	CS1	CS0	Description			
	0	0	fINTRC INTRC (8 MHz)			

- 0 0 fXIN Main Clock (1~16 MHz)
- 0 1 0 1 1 fSUB (32.768 KHz)
- 1 fRING (125 KHz)

# 11.2 BIT

## 11.2.1 Overview

The Z51F0811 has one 8-bit Basic Interval Timer that is free-run and can't stop. Block diagram is shown in Figure 11-2. In addition, the Basic Interval Timer generates the time base for watchdog timer counting. It also provides a basic interval timer interrupt (BITF).

The Z51F0811 has these Basic Interval Timer (BIT) features:

- During Power On, BIT gives a stable clock generation time
- On exiting Stop mode, BIT gives a stable clock generation time
- As clock function, time interrupt occurrence





Figure 11-2 BIT Block Diagram

# 11.2.3 Register Map

Table 11-2 Register Map

Name	Address	Dir	Default	Description
BCCR	8BH	R/W	05H	BIT Clock Control Register
BITR	8CH	R	00H	Basic Interval Timer Register

### 11.2.4 Bit Interval Timer Register description

The Bit Interval Timer Register consists of BIT Clock control register (BCCR) and Basic Interval Timer register (BITR). If BCLR bit set to '1', BITR becomes '0' and then counts up. After 1 machine cycle, BCLR bit is cleared as '0' automatically.

## 11.2.5 Register description for Bit Interval Timer

## BCCR (BIT Clock Control Register) : 8BH

7	6	5		4	3	2	1	0
BITF	-	-		-	BCLR	BCK2	BCK1	BCK0
RW	R	R		R	RW	RW	RW	RW
								Initial value : 05
		BITF	When E to this b		rupt occurs	, this bit become	es '1'. For cle	aring bit, write '
			0	no gene	eration			
			1	generat	tion			
		BCLR	If BCLF	Bit is w	ritten to '1	, BIT Counter is	cleared as '0	,
			0	Free Ru	unning			
			1	Clear C	ounter			
	В	CK[2:0]	Select I	BIT over	flow period	(BIT Clock ≒3.9	9 KHz)	
			BCK2	BCK1	BCK0			
			0	0	0	0.512msec (BI	Γ Clock * 2)	
			0	0	1	1.024msec		
			0	1	0	2.048msec		
			0	1	1	4.096msec		
			1	0	0	8.192msec		
			1	0	1	16.384msec (d	efault)	
			1	1	0	32.768msec		
			1	1	1	65.536msec		
ITR (Basic	c Interval Tir	ner Regis	ster) : 80	н				
7	6	5		4	3	2	1	0

 7	6	5	4	3	2	1	0
BIT7	BIT6	BIT5	BIT4	BIT3	BIT2	BIT1	BITO
R	R	R	R	R	R	R	R
						I	nitial value : 0

BIT[7:0] BIT Counter
# 11.3 WDT

### 11.3.1 Overview

The watchdog timer rapidly detects the CPU malfunction such as endless looping caused by noise or the like, and resumes the CPU to the normal state. The watchdog timer signal for detecting malfunction can be selected either a reset CPU or an interrupt request. When the watchdog timer is not being used for malfunction detection, it can be used as a timer to generate an interrupt at fixed intervals. It is possible to use free running 8-bit timer mode (WDTRSON='0') or watch dog timer mode (WDTRSON='1') as setting WDTMR[6] bit. If writing WDTMR[5] to '1', WDT counter value is cleared and counts up. After 1 machine cycle, this bit has '0' automatically. The watchdog timer consists of 8-bit binary counter and the watchdog timer data register. When the value of 8-bit binary counter is equal to the 8 bits of WDTR, the interrupt request flag is generated. This can be used as Watchdog timer interrupt or reset the CPU in accordance with the bit WDTRSON.

The clock source of Watch Dog Timer is BIT overflow output. The interval of watchdog timer interrupt is decided by BIT overflow period and WDTR set value. The equation is as below

WDT Interrupt Interval = (BIT Interrupt Interval) X (WDTR Value+1)



### 11.3.2 Block Diagram

Figure 11-3 WDT Block Diagram

### 11.3.3 Register Map

Table 11-3 Register Map

Name	Address	Dir	Default	Description
WDTR	8EH	W	FFH	Watch Dog Timer Register
WDTCR	8EH	R	00H	Watch Dog Timer Counter Register
WDTMR	8DH	R/W	00H	Watch Dog Timer Mode Register

### 11.3.4 Watch Dog Timer Register description

The Watch dog timer (WDT) Register consists of Watch Dog Timer Register (WDTR), Watch Dog Timer Counter Register (WDTCR) and Watch Dog Timer Mode Register (WDTMR).

### 11.3.5 Register description for Watch Dog Timer

### WDTR (Watch Dog Timer Register: Write Case) : 8EH

7	6	5	4	3	2	1	0
WDTR7	WDTR6	WDTR5	WDTR4	WDTR3	WDTR2	WDTR1	WDTR0
W	W	W	W	W	W	W	W
						li li	nitial value : FFI

WDTR[7:0] Set a period

WDT Interrupt Interval=(BIT Interrupt Interval) x(WDTR Value+1)

Note) To guarantee proper operation, the data should be greater than 01H.

### WDTCR (Watch Dog Timer Counter Register: Read Case) : 8EH

7	6	5	4	3	2	1	0
WDTCR7	WDTCR6	WDTCR5	WDTCR4	WDTCR3	WDTCR2	WDTCR1	WDTCR0
R	R	R	R	R	R	R	R

Initial value : 00H

WDTCR[7:0] WDT Counter

### WDTMR (Watch Dog Timer Mode Register) : 8DH

7	6	5		4	3	2	1	0
WDTEN	WDTRSON	WDTCL		-	-	-	-	WDTIFR
RW	RW	RW		-	-	-	-	RW
							I	nitial value : 00H
	v	VDTEN	Control	WDT o	peration			
			0	disable	e			
			1	enable	)			
	W	OTRSON	Control	WDT F	leset operation	า		
			0	Free R	lunning 8-bit ti	mer		
			1	Watch	Dog Timer Re	eset ON		
	V	VDTCL	Clear W	DT Co	unter			
			0	Free R	lun			
			1	Clear	NDT Counter	(auto clear aft	er 1 Cycle)	
	v	VDTIFR			terrupt occurs auto clear by			clearing bit, write
			0	WDT I	nterrupt no ge	neration		
			1	WDT I	nterrupt gener	ation		







# 11.4 WT

### 11.4.1 Overview

The watch timer has the function for RTC (Real Time Clock) operation. It is generally used for RTC design. The internal structure of the watch timer consists of the clock source select circuit, timer counter circuit, output select circuit and watch timer mode register. To operate the watch timer, determine the input clock source, output interval and set WTEN to '1' in watch timer mode register (WTMR). It is able to execute simultaneously or individually. To stop or reset WT, clear the WTEN bit in WTMR register. Even if CPU is STOP mode, sub clock is able to be alive so WT can continue the operation. The watch timer counter circuits may be composed of 21-bit counter which is low 14-bit with binary counter and high 7-bit with auto reload counter in order to raise resolution. In WTR, it can control WT clear and set Interval value at write time, and it can read 7-bit WT counter value at read time.



### 11.4.2 Block Diagram

Figure 11-5 Watch Timer Block Diagram

### 11.4.3 Register Map

Table 11-4 Register Map

Name	Address	Dir	Default	Description
WTMR	9DH	R/W	00H	Watch Timer Mode Register
WTR	9EH	W	7FH	Watch Timer Register
WTCR	9EH	R	00H	Watch Timer Counter Register

### 11.4.4 Watch Timer Register description

The watch timer register (WT) consists of Watch Timer Mode Register (WTMR), Watch Timer Counter Register (WTCR) and Watch Timer Register (WTR). As WTMR is 6-bit writable/readable register, WTMR can control the clock source (WTCK), interrupt interval (WTIN) and function enable/disable (WTEN). Also there is WT interrupt flag bit (WTIFR).

### 11.4.5 Register description for Watch Timer

7	6	5	4		3	2	1	0
WTEN	-	-	WTIF	R V	VTIN1	WTINO	WTCK1	WTCK0
RW	-	-	RW	1	RW	RW	RW	RW
							I	nitial value : 00
		WTEN	Control W	atch Time	r			
			0 d	isable				
			1 e	nable				
		WTIFR				his bit become ACK signal.		aring bit, write
			0 V	VT Interrup	t no gene	eration		
			1 V	VT Interrup	t genera	tion		
	v	VTIN[1:0]	Determin	e interrupt	interval			
			WTIN1	WTIN0	descrip	otion		
			0	0	fwck/2	048		
			0	1	fwck/8	192		
			1	0	fwck/1	6384		
			1	1	fwck/1	6384 x (7bit W	/T Value)	
	N	/TCK[1:0]	Determin	e Source (	Clock			
			WTCK1	WTCK0	descrip	otion		
			0	0	fsub			
			0	1	fx/256			
			1	0	fx/128			
			1	1	fx/64			
	Remark:	fx- Main sy	stem clock	oscillation	frequenc	;y		
		fsub- Sub o	clock oscilla	tion freque	ncy			
		fwck- selec	ted Watch	Timer clock	¢			

# WTMR (Watch Timer Mode Register) : 9DH

# WTR (Watch Timer Register: Write Case) : 9EH

7	6	5	4	3	2	1	0
WTCL	WTR6	WTR5	WTR4	WTR3	WTR2	WTR1	WTR0
W	W	W	W	W	W	W	W
						I	nitial value : 7FH
		-	Clear WT Cour				
		(	) Free R				
			1 Clear	NT Counter (a	auto clear after	r 1 Cycle)	
	v	VTR[6:0]	Set WT period WT Interrupt I		2^14) x(7bit W	/T Value+1)	
		1	Note) To guara	ntee proper op	peration, it is g	greater than 0 <sup>-</sup>	1H to write WTR.
WTCR (Wat	ch Timer Co	ounter Regi	ster: Read C	ase) : 9EH			

7	6	5	4	3	2	1	0
	WTCR6	WTCR5	WTCR4	WTCR3	WTCR2	WTCR1	WTCR0
-	R	R	R	R	R	R	R
							nitial value : 0

WTCR[6:0] WT Counter

### 11.5 Timer/PWM

### 11.5.1 8-bit Timer/Event Counter 0, 1

### 11.5.1.1 Overview

Timer 0 and timer 1 can be used either two 8-bit timer/counter or one 16-bit timer/counter with combine them. Each 8-bit timer/event counter module has multiplexer, 8-bit timer data register, 8-bit counter register, mode register, input capture register, comparator. For PWM, it has PWM register (T1PPR, T1ADR, T1BDR, T1CDR, T1PCR, T1PCR2, T1PCR3, T1PHR, T1DLYA, T1DLYB, T1DLYC, T1ISR, T1IMSK).

It has seven operating modes:

- 8-bit timer/counter mode
- 8-bit capture mode
- 8-bit compare output mode
- 16-bit timer/counter mode
- 16-bit capture mode
- 16-bit compare output mode
- PWM mode

Note> TxDR must be set to higher than 0x03 for guaranteeing operation.

The timer/counter can be clocked by an internal or an external clock source (external EC0). The clock source is selected by clock select logic which is controlled by the clock select (T0CK[2:0], T1CK[3:0]). Also the timer/PWM/event counter 1 can use more clock sources than timer/event counter 0.

- TIMER0 clock source: fX/2, 4, 8, 32, 128, 512, 2048, EC0

- TIMER1 clock source: fX/1, 2, 4, 8, 16, 32, 64, 128, 256, 512, 1024, 2048, 4096, 8192, 16384, T0CK

In the capture mode, by INT0, INT1, the data is captured into Input Capture Register. The timer 0 outputs the compare result to T0 port in 8/16-bit mode. Also the timer 1 outputs the result to T1 port in the timer mode and the PWM wave form to PWMA, PWMAB(bar), PWMB, PWMBB, PWMC, PWMCB Port(6-channel) in the PWM mode.

16 Bit	CAP0	CAP1	PWM1E	T0CK[2:0]	T1CK[3:0]	T0/1_PE	Timer 0	Timer 1
0	0	0	0	XXX	XXXX	00	8 Bit Timer	8 Bit Timer
0	0	1	0	111	XXXX	00	8 Bit Event Counter	8 Bit Capture
0	1	0	0	XXX	XXXX	01	8 Bit Capture	8 Bit Compare Output
0	0	0	1	XXX	XXXX	11	8 Bit Timer/Counter	10 Bit PWM
1	0	0	0	XXX	1111	00	16 B	it Timer
1	0	0	0	111	1111	00	16 Bit Ev	ent Counter
1	1	1	0	XXX	1111	00	16 Bit	Capture
1	0	0	0	XXX	1111	01	16 Bit Con	npare Output

Table 11-5 Timer 0,1 operating modes

### 11.5.1.2 8 Bit Timer/Counter Mode



The 8-bit Timer/Counter Mode is selected by control registers as shown in Figure 11-6.

Figure 11-6 8 Bit Timer/Event Counter 0, 1 Block Diagram

The two 8-bit timers have each counter and data register. The counter register is increased by internal or external clock input. The timer 0 can use the input clock with one of 2, 4, 8, 32, 128, 512, 2048 prescaler division rates (T0CK[2:0]). The timer 1 can use the input clock with one of 1, 2, 8 ~ 16384 and timer 0 overflow clock (T1CK[3:0]). When the value of T0,1 value and the value of T0DR, T1DR are respectively identical in Timer 0, 1, the interrupt of TimerP 0, 1 occurs. The external clock (EC0) counts up the timer at the rising edge. If EC0 is selected from T0CK[2:0], EC0 port becomes input port. The timer 1 can't use the external EC0 clock.



Figure 11-7 Timer/Event Counter 0, 1 Example



Figure 11-8 Timer/Event Counter0, 1 Count Operation

### 11.5.1.3 16 Bit Timer/Counter Mode

The timer register is being run with all 16 bits. A 16-bit timer/counter register T0, T1 are incremented from 0003H to FFFFH until it matches T0DR, T1DR and then resets to 0000H. The match output

generates the Timer 0 interrupt (No timer 1 interrupt). The clock source is selected from T0CK[2:0] and T1CK[3:0] must set 1111b and 16BIT bit must set to '1'. The timer 0 is LSB 8-bit, the timer 1 is MSB 8-bit. T0DR must not be 0x00 ( 0x01~0xFF). The 16-bit mode selection is shown as Figure 11-9.



Figure 11-9 16-bit Timer/Counter for Time 0, 1

#### 11.5.1.4 8-Bit Capture Mode

The timer 0, 1 capture mode is set by CAP0, CAP1 as '1'. The clock source can use the internal/external clock. Basically, it has the same function of the 8-bit timer/counter mode and the interrupt occurs at T0, T1 and T0DR, T1DR matching time, respectively. The capture result is loaded into CDR0, CDR1. The T0, T1 value is automatically cleared by hardware and restarts counter.

This timer interrupt in capture mode is very useful when the pulse width of captured signal is wider than the maximum period of timer.

As the EIEDGE and EIPOLA register setting, the external interrupt INT1, INT0 function is chosen.

The CDR0, T0 and T0DR are in same address. In the capture mode, reading operation reads the CDR0, not T0DR because path is opened to the CDR0. The CDR1 has the same function.

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- WWW



Figure 11-10 8-bit Capture Mode for Timer 0, 1

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Figure 11-11 Input Capture Mode Operation of Timer 0, 1



Figure 11-12 Express Timer Overflow in Capture Mode

#### 11.5.1.5 16 Bit Capture Mode

The 16-bit capture mode is the same operation as 8-bit capture mode, except that the timer register uses 16 bits.

The clock source is selected from T0CK[2:0] and T1CK[3:0] must set 1111b and 16BIT bit must set to '1'. The 16-bit mode setting is shown as Figure 11-13.



Figure 11-13 16-bit Capture Mode of Timer 0, 1

### 11.5.1.6 PWM Mode

The timer 1 has a high speed PWM (pulse Width Modulation) function. In PWM mode, the 6-channel pins output up to 10-bit resolution PWM output. This pin should be configured as a PWM output by set PWM1E to '1'. The period of the PWM output is determined by the T1PPR (PWM period register) + T1PHR[1:0], T1xDR (each channel PWM duty register) + T1PHR[7:2].

PWM Period = [ T1PHR[1:0]T1PPR ] X Source Clock PWM Duty(A-ch) = [ T1PHR[7:6] T1ADR ] X Source Clock

Note> T1PPR must be set to higher than T1PDR for guaranteeing operation.

Devel for		Frequency	
Resolution	T1CK[3:0]=0001 (250ns)	T1CK[3:0]=0010 (500ns)	T1CK[3:0]=0100 (2us)
10 Bit	3.9KHz	1.95KHz	0.49KHz
9 Bit	7.8KHz	3.9KHz	0.98KHz
8 Bit	15.6KHz	7.8KHz	1.95KHz
7 Bit	31.2KHz	15.6KHz	3.91KHz

Table 11-6 PWM Frequency vs. Resolution at 8 Mhz

The POLx bit of T1PCR3 register decides the polarity of duty cycle. If the duty value is set same to the period value, the PWM output is determined by the bit POLx (1: High, 0: Low). And if the duty value is set to "00H", the PWM output is determined by the bit POLx (1: Low, 0: High).

- WW



Figure 11-14 PWM Mode (Force 6-ch)

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Figure 11-15 PWM Mode (Force All-Ch)



Figure 11-16 Example of PWM at 4MHz





### Update period & duty register value at once

The period and duty of PWM comes to move from temporary registers to PPR (PWM Period Register) and PDR (PWM Duty Register) when always period match occurs. If you want that the period and duty is immediately changed, the UPDT bit in the T1PCR register must set to '1'. It should be noted that it needs the 3 cycle of timer clock for data transfer in the internal clock synchronization circuit. So the update data is written before 3 cycle of timer clock to get the right output waveform.

### Phase correction & Frequency correction

On operating PWM, it is possible that it is changed the phase and the frequency by using BMOD bit (back-to-back mode) in T1PCR register. (Figure 11-18, Figure 11-19, Figure 11-20 referred)

In the back-to-back mode, the counter of PWM repeats up/down count. In fact, the effective duty and period becomes twofold of the register set values. (Figure 11-18, Figure 11-19 referred)



Figure 11-18 Example of PWM Output Waveform



Figure 11-19 Example of PWM waveform in Back-to-Back mode at 4Mhz

- WWW



Figure 11-20 Example of Phase Correction and Frequency correction of PWM

### External Sync

If using ESYNC bit of T1PCR register, it is possible to synchronize the output of PWM from external signal.

If ESYNC bit sets to '1', the external signal moves to PWM module through the BLNKB pin (P16). If BLNKB signal is low, immediately PWM output becomes a reset value, and internal counter becomes reset. If BLNKB signal returns to '1', the counter is started again and PWM output is normally generated. (Figure 11-21 referred)

### **PWM Halt**

If using PHLT bit of T1PCR register, it is possible to stop PWM operation by the software. During PHLT bit being '1', PWM output becomes a reset value, and internal counter becomes reset as 0. Without changing PWM setting, temporarily it is able to stop PWM. In case of T1CN, when stopping counter, PWM output pin remains before states. But if PHLT bit sets to '1', PWM output pin has reset value.



Figure 11-21 Example of PWM External Synchronization with BLNKB Input

### FORCE Drive ALL ch with A-ch mode

If FORCA bit sets to '1', it is possible to enable or disable all PWM output pins through PWM outputs which occur from A-ch duty counter. It is noted that the inversion outputs of A, B, C channel have the same A-ch output waveform. According to POLA/B/C, it is able to control the inversion of outputs.





### FORCE 6-Ch Drive

If FORC6 bit sets to '1', it is possible to enable or disable PWM output pin and inversion output pin generated through the duty counter of each channel. The inversion output is the reverse phase of the PWM output. A A/AB output of the A-channel duty register, a B/BB output of the B-channel duty register, a C/CB output of the C-channel duty register are controlled respectively. If the UALL bit is set to '1', it is updated B/C channel duty at the same time, when it is written by a A-channel duty register.



Figure 11-23 Example of Force Drive 6-ch Mode

- WWW

0000_0000 <sub>E</sub>
)000_0000
0000 0000
0000 00-0 <sub>8</sub>
stop*
<b>stop*</b> '0"
stop*

The port is fixed as the output at the Timer mode. But at the PWM mode, using POCON bit, when PWM output port does not work, it can change the state of output port to input (high-Z) state. It is determined from the port control register to select in/out signal. If using FORCE mode, it can only change the wanted channel direction of the 6-channel outputs. In the FORCE mode, the channel direction of the disabled output is determined by each port control register bit, regardless of the PWM stop.

Figure 11-24 PWM Port control

#### **PWM output Delay**

If using the PDLYA, PDLYB, PDLYC register, it can delay PWM output based on the rising edge. At that time, it does not change the falling edge, so the duty is reduced as the time delay. In POLA/B/C setting to '0', the delay is applied to the falling edge. In POLA/B/C setting to '1', the delay is applied to the rising edge. It can produce a pair of Non-overlapping clock. The each channel is able to have 4-bit delay. As it can select the clock up to 1/8 divided clock using NOPS1, NOPS0, the delay of its maximum 128 timer clock cycle is produced.

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Figure 11-25 Example of PWM Delay

#### 11.5.1.7 8-Bit (16 Bit) Compare Output Mode

If the T1 (T0+T1) value and the T1DR (T0DR+T1DR) value are matched, T1/PWM1A port outputs. The output is 50:50 of duty square wave, the frequency is following

 $f_{COMP} = \frac{\text{Oscillator Frequency}}{2 \times \text{Prescaler Value} \times (TDR + 1)}$ 

To export the compare output as T1/PWM1A, the T1\_PE bit in the T1PCR3 register must set to '1'.

#### 11.5.1.8 Register Map

Name	Address	Dir	Default	Description
T0CR	B2H	R/W	00H	Timer 0 Mode Control Register
Т0	ВЗН	R	00H	Timer 0 Register
T0DR	ВЗН	W	FFH	Timer 0 Data Register
CDR0	ВЗН	R	00H	Capture 0 Data Register
T1CR	B4H	R/W	00H	Timer 1 Mode Control Register
T1DR	B5H	W	FFH	Timer 1 Data Register
T1PPR	B5H	W	FFH	Timer 1 PWM Period Register
T1	B6H	R	00H	Timer 1 Register
T1ADR	B6H	R/W	7FH	Timer 1 PWM 1A Duty Register
CDR1	B6H	R	00H	Capture 1 Data Register
T1PCR	B7H	R/W	00H	Timer 1 PWM Control Register
T1BDR	BAH	R/W	7FH	Timer 1 PWM 1B Duty Register
T1CDR	BBH	R/W	7FH	Timer 1 PWM 1C Duty Register
T1PHR	всн	R/W	00H	Timer 1 PWM High Register
T1PCR2	BDH	R/W	00H	Timer 1 PWM Control Register 2
T1PCR3	BEH	R/W	00H	Timer 1 PWM Control Register 3
T1DLYA	BFH	R/W	00H	PWM1 Non-Overlap Delay Register ch. A/AB
T1DLYB	C2H	R/W	00H	PWM1 Non-Overlap Delay Register ch. B/BB
T1DLYC	СЗН	R/W	00H	PWM1 Non-Overlap Delay Register ch. C/CB
T1ISR	C4H	R/W	00H	Timer 1 Interrupt Status Register
T1IMSK	C5H	R/W	00H	Timer 1 Interrupt Mask Register
PLLCR	85H	R/W	42H	Timer1 PLL Control Register

### 11.5.1.9 Timer/Counter 0, 1 Register description

The Timer/Counter 0,1 register consists of Timer 0 Mode Control Register (T0CR), Timer 0 Register (T0), Timer 0 Data Register (T0DR), Capture 0 Data Register (CDR0), Timer 1 Mode Control Register (T1CR), Timer 1 Data Register (T1DR), Timer 1 PWM Period Register (T1PPR), Timer 1 Register (T1), Timer 1 PWM 1A Duty Register (T1ADR), Capture 1 Data Register (CDR1), Timer 1 PWM Control Register (T1PCR), Timer 1 PWM 1B Duty Register (T1BDR), Timer 1 PWM 1C Duty Register

(T1CDR), Timer 1 PWM High Register (T1PHR), Timer 1 PWM Control Register 2 (T1PCR2), Timer 1 PWM Control Register 3 (T1PCR3), PWM1 Non-Overlap Delay Register ch. A/AB (T1DLYA), PWM1 Non-Overlap Delay Register ch. B/BB (T1DLYB), PWM1 Non-Overlap Delay Register ch. C/CB (T1DLYC), Timer 1 Interrupt Status Register (T1ISR), Timer 1 Interrupt Mask Register (T1IMSK) and PLL Control Register (PLLCR)

### 11.5.1.10 Register description for Timer/Counter 0, 1

7	6	5		4	3	2	1	0
TOEN	T0_PE	CAP0	TO	CK2	T0CK1	TOCK0	TOCN	TOST
RW	RW	RW	R	Ŵ	RW	RW	RW	RW
							I	nitial value : 00H
		T0EN	Control	Timer C	)			
			0	Timer (	) disable			
			1	Timer (	) enable			
		T0_PE	Control	Timer C	Output port			
			0	Timer (	0 Output disat	ble		
			1	Timer (	0 Output enab	le		
		CAP0	Control	Timer C	operation mo	ode		
			0	Timer/	Counter mode	1		
			1	Captur	e mode			
	Т	0CK[2:0]	Select T	imer 0	clock source.	Fx is main sys	stem clock fre	quency
			T0CK2	T0CK	(1 T0CK0	description		
			0	0	0	fx/2		
			0	0	1	fx/4		
			0	1	0	fx/8		
			0	1	1	fx/32		
			1	0	0	fx/128		
			1	0	1	fx/512		
			1	1	0	fx/2048		
			1	1	1	External Clo	ock (EC0)	
		T0CN	Control	Timer C	Count pause	/continue		
			0	Tempo	rary count sto	р		
			1		ue count			
		TOST	Control		) start/stop			
			0	Counte	er stop			
			1	Clear o	counter and st	art		

#### T0CR (Timer 0 Mode Control Register) : B2H

#### T0 (Timer 0 Register: Read Case) : B3H

7	6	5	4	3	2	1	0
T07	T06	T05	T04	T03	T02	T01	T00
R	R	R	R	R	R	R	R
						I	nitial value : 00H

T0 Counter

T0[7:0]

### T0DR (Timer 0 Data Register: Write Case) : B3H

7	6	5	4	3	2	1	0
T0D7	T0D6	T0D5	T0D4	T0D3	T0D2	T0D1	TODO
W	W	W	W	W	W	W	W
						h	nitial value : FF

T0D[7:0] T0 Compare

# CDR0 (Capture 0 Data Register: Read Case, Capture mode only) : B3H

7	6	5	4	3	2	1	0
CDR07	CDR06	CDR05	CDR04	CDR03	CDR02	CDR01	CDR00
R	R	R	R	R	R	R	R
						I	nitial value : 00

CDR0[7:0] T0 Capture

# T1CR (Timer 1 Mode Count Register) : B4H

7	6	5	4	3	2	1	0
16BIT	CAP1	T1CN	T1ST	T1CK3	T1CK2	T1CK1	T1CK0
RW	RW	RW	RW	RW	RW	RW	RW
		CAP1	1 Captu	1 operation mo Counter mode re mode			Initial value : 0
				1 Count pause prary count sto ue count			
				1 start/stop er stop counter and st	art		
	Τ	1CK[3:0]	Select Timer 1	clock source.	Fx is main sys	stem clock fre	equency
				CK2 T1CK1		descriptior	า
			0 0	0	0	fx	
			0 0	0	1	fx/2	
			0 0	1	0	fx/4	
			0 0	1	1	fx/8	
			0 1	0	0	fx/16	
			0 1	0	1	fx/32	
			0 1	1	0	fx/64	
			0 1	1	1	fx/128	
			1 0	0	0	fx/256	
			1 0	0	1	fx/512	
			1 0	1	0	fx/1024	
			1 0	1	1	fx/2048	
			1 1	0	0	fx/4096	

						Product	Z51F0811 Specification
			1 1	0	1	fx/8192	
		1	1 1	1	0	fx/16384	
		1	I 1	1	1	Using Time	er 0 Clock
		Note	) If you want to	use "Using Ti	imer 0 Clock",	you can set T	OEN bit in TOCR
T1DR (Time	r 1 Data Reg	gister: Write	e Case) : B5I	1			
7	6	5	4	3	2	1	0
T1D7	T1D6	T1D5	T1D4	T1D3	T1D2	T1D1	T1D0
W	W	W	W	W	W	W	W
						li	nitial value : FFH
	T	[1D[7:0]	T1 Compare				
T1PPR (Tim	er 1 PWM P	eriod Regis	ster: Write Ca	ase PWM m	ode only) : I	35H	
7	6	5	4	3	2	1	0
T1PP7	T1PP6	T1PP5	T1PP4	T1PP3	T1PP2	T1PP1	T1PP0
W	W	W	W	W	W	W	W
						lı	nitial value : FFH
	-	40017-01		-1			

T1PP[7:0] T1 PWM period

#### T1 (Timer 1 Register: Read Case) : B6H

7	6	5	4	3	2	1	0
T17	T16	T15	T14	T13	T12	T11	T10
R	R	R	R	R	R	R	R

Initial value : 00H

T1[7:0] T1 Counter period

### T1ADR (Timer 1 PWM 1A Duty Register PWM mode only) : B6H

7	6	5	4	3	2	1	0
PAD7	PAD6	PAD5	PAD4	PAD3	PAD2	PAD1	PAD0
RW							
						I	nitial value : 7F

T1ADR[7:0] T1 PWM Duty

Note) only write, when PWM1E '1'

### CDR1 (Capture 1 Data Register: Read Case, Capture mode only) : B6H

7	6	5	4	3	2	1	0	
CDR17	CDR16	CDR15	CDR14	CDR13	CDR12	CDR11	CDR10	
R	R	R	R	R	R	R	R	
						I	nitial value : 0	0H

CDR1[7:0] T1 Capture

### T1PCR (Timer 1 PWM Control Register) : B7H

7	6	5	4	3	2	1	0
PWM1E	ESYNC	BMOD	PHLT	UPDT	UALL	NOPS1	NOPS0

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		-	al fai	100	1
4	h	-	100	_	

RW	RW	RW		RW	RW	RW	RW	RW			
								Initial value : 00			
		PWM1E	Control PWM								
			0	0 PWM disable							
			1	PWM	enable						
		ESYNC	Select the operation of External Sync Mode								
			0	Exterr	nal Sync Mo	de disable					
			1	1 External Sync Mode enable (using with BLNKB(P16))							
		BMOD	Contr	Control Back-To-Back Mode operation							
			0	0 BtB mode disable (only up count)							
			1	1 BtB mode enable (Up/Down count)							
		PHLT	Contr	ol PWM							
			0 PWM running								
			1								
		UPDT	Determine the update time of PPR, PDR								
			0 Update at period match								
			1	Updat	te at any tim	ne (after 3 timer o	clock, updat	e)			
		UALL	Contr	ol update	e all duty reg	gister					
			0	Write	duty registe	r separately					
			1	Write	all duty regi	sters (via A duty	)				
	N	IOPS1[1:0]			erlap presca						
					-	tion clock freque	ncy				
			NOF	• •		description					
			0	C		fpwm					
			0	1		fpwm/2					
			1	C		fpwm/4					
			1	1		fpwm/8					

history and the second

### T1BDR (Timer 1 PWM 1B Duty Register) : BAH

7	6	5	4	3	2	1	0
PBD7	PBD6	PBD5	PBD4	PBD3	PBD2	PBD1	PBD0
RW							

Initial value : 7FH

T1BDR[7:0] PWM 1B ch Duty Note) only write, when PWM1E '1'

T1CDR (Timer 1 PWM 1C Duty Register) : BBH

7	6	5	4	3	2	1	0
PCD7	PCD6	PCD5	PCD4	PCD3	PCD2	PCD1	PCD0
RW							
							nitial value : 7FI

T1CDR[7:0] PWM 1C ch Duty

Note) only write, when PWM1E '1'

## T1PHR (Timer 1 PWM High Register) : BCH

7	6	5	4	3	2	1	0
ADR9	ADR8	BDR9	BDR8	CDR9	CDR8	PPR9	PPR8
RW							

Initial value : 00H

:01

:0]

:0]

:0]

ADR[9:8]	F	PWM 1A High (Bit [9:8])									
BDR[9:8]	F	PWM 1B High (Bit [9:8])									
CDR[9:8]	F	PWM 1C High (Bit [9:8])									
PPR[9:8]	F	PERIOD High (Bit [9:8])									
PERIOD:		PPR9	PPR8		T1PPR[7:						
DUTY A:		ADR9	ADR8		P1ADR[7:						
DUTY B:		BDR9	BDR8		P1BDR[7:						
DUTY C:		CDR9	CDR8		P1CDR[7]						

### T1PCR2 (Timer 1 PWM Control Register 2) : BDH

7	6	5	4	3	2	1	0			
FORCA	FORC6	PAOE	PABOE	PBOE	PBBOE	PCOE	PCBOE			
RW	RW	RW	RW	RW	RW	RW	RW			
						I	nitial value : 00			

FORCA	Contro	Control Force Drive A Channel mode						
	0	Force Drive A Channel mode disable						
	1	Force Drive A Channel mode enable						
FORC6		ol Force 6 Channel mode PAOE~PCBOE is effective when FORC6 sets to '1'						
	0	Force 6 Channel mode disable						
	1	Force 6 Channel mode enable						
PAOE/	Select	t Channel A/AB operation						
PABOE	0	P1A (or P1AB) output disable						
	1	P1A (or P1AB) output enable						
PBOE/	Select	t Channel B/BB operation						
PBBOE	0	P1B (or P1BB) output disable						
	1	P1B (or P1BB) output enable						

PRELIMINARY

PCOE/	Select Channel C/CB operation							
PCBOE	0	P1C (or P1CB) output disable						
	1	P1C (or P1CB) output enable						

### T1PCR3 (Timer 1 PWM Control Register 3) : BEH

6 5	4	3	2	1	0				
E POLA POLB	POLC	POCON	HCKE	-	PLLPDB				
RW RW	RW	RW	RW	-	RW				
					nitial value : 00				
T1_PE	Control Timer1	/PWM1 Outpu	it port						
	0 T1, P\	VM1 Output o	peration disab	le					
	1 T1, P\	VM1 Output o	peration enabl	е					
POLA	Configure PWI	M A-ch polarity	/						
	0 Negat	ive (Duty Mate	h time, Clear)						
	1 Positiv	ve (Duty Match	i time, Set)						
POLB	POLB Configure PWM B-ch polarity								
	0 Negat	ive (Duty Mate	h time, Clear)						
	1 Positiv	ve (Duty Match	i time, Set)						
POLC	Configure PWI	M C-ch polarity	/						
	0 Negat	ive (Duty Mate	h time, Clear)						
	1 Positiv	ve (Duty Match	i time, Set)						
POCON	Control PWM	output operatio	n						
	0 PWM	output control	disable						
	1 PWM	output control	enable						
HCKE	Select High fre	quency							
	Note) fCK is sy	stem frequent	cy, Fout is PLL	output freque	ency				
	0 High f	requency disal	ble						
	1 High f	requency enat	ole (Fout > 3* f	CK)					
PLLPDB	Control PLL po	ower Down mo	de						
	0 PLL d	isable (power	down mode)						
	1 PLL e	nable (for stat	ole, needs 1ms	s wait)					

# T1DLYA (PWM1 Non-Overlap Delay Register for channel A/AB) : BFH

7	6	5	4	3	2	1	0		
DLYA3	DLYA2	DLYA1	DLYAO	DLYAB3	DLYAB2	DLYAB1	DLYAB0		
RW	RW	RW	RW	RW	RW	RW	RW		
Initial value : 00H									
			PWM A chann PWM AB char	•	, 0 0	.,			

# T1DLYB (PWM1 Non-Overlap Delay Register for channel B/BB) : C2H

7	6	5	4	3	2	1	0	
DLYB3	DLYB2	DLYB1	DLYB0	DLYBB3	DLYBB2	DLYBB1	DLYBB0	
RW	RW	RW	RW	RW	RW	RW	RW	

Initial value : 00H

**DLYB[3:0]**PWM B channel Output Delay (Rising edge only)**DLYBB[3:0]**PWM BB channel Output Delay (Rising edge only)

### T1DLYC (PWM1 Non-Overlap Delay Register for channel C/CB) : C3H

7	6	5	4	3	2	1	0
DLYC3	DLYC2	DLYC1	DLYC0	DLYCB3	DLYCB2	DLYCB1	DLYCB0
RW	RW	RW	RW	RW	RW	RW	RW
							1011 1 1 001

Initial value : 00H

**DLYC[3:0]**PWM C channel Output Delay (Rising edge only)**DLYCB[3:0]**PWM CB channel Output Delay (Rising edge only)

#### T1ISR (Timer 1 Interrupt Status Register) : C4H

7	6	5	4	3	2	1	0			
<b>IOV</b> R	IBTM	ICMA	ICMB	ICMC	ICAP	-	-			
RW	RW	RW	RW	RW	RW	RW	RW			
							Initial value : 00			
		IOVR	Overflow (mate interrupt status	de or T1PPR	in PWM mod					
			Note) for clear,	write '1' to thi	s bit					
			0 Overfle	ow no occurre	nce					
			1 Overfle							
		IBTM Timer Bottom (goto zero) interrupt status in PWM Back-to-E Note) for clear, write '1' to this bit								
			0 Timer	Bottom no oco	currence					
			1 Timer Bottom occurrence							
		ICMA	PWM A-ch Duty Match interrupt status							
			Note) for clear,	write '1' to thi	s bit					
			0 PWM A-ch Duty Match no occurrence							
			1 PWM A-ch Duty Match occurrence							
		ICMB	PWM B-ch Duty Match interrupt status							
			Note) for clear,	write '1' to thi	s bit					
			0 PWM	B-ch Duty Mat	ch no occurre	nce				
			1 PWM	PWM B-ch Duty Match occurrence						
		ICMC								
			Note) for clear, write '1' to this bit							
			0 PWM	C-ch Duty Mat	ch no occurre	nce				
			1 PWM C-ch Duty Match occurrence							
		ICAP	Timer Capture event interrupt status							
			Note) for clear,	write '1' to thi	s bit					
			0 Timer	Capture event	no occurrenc	е				
			1 Timer	Capture event	occurrence					
1MSK (Time	er 1 Interro	upt Mask F	Register) : C5H	4						
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7	6	5	4	3	2	1	0			
OVRMSK	BTMMSK	CMAMSK	CMBMSK	CMCMSK	CAPMSK	-	-			
RW	RW RW RW		RW	RW	RW	RW	RW			
						I	nitial value : 00			
	0	VRMSK	Control Overflow interrupt							
			0 Overflow interrupt disable							
			1 Overflow interrupt enable							
	В	TMMSK	Control Timer Bottom interrupt							
			0 Timer	Timer Bottom interrupt disable						
			1 Timer Bottom interrupt enable							
	C	MAMSK	Control Timer Compare Match (or PWM A-ch Match) interrupt							
			0 Timer Compare Match (or PWM A-ch Match) interrupt disable							
			1 Timer Compare Match (or PWM A-ch Match) interrupt enable							
	C	MBMSK	Control PWM B-ch Match interrupt							
			0 PWM B-ch Match interrupt disable							
			1 PWM I	B-ch Match in	terrupt enable					
	C	MCMSK	Control PWM C-ch Match interrupt							
			0 PWM C-ch Match interrupt disable							
			1 PWM C-ch Match interrupt enable							
	С	APMSK	Control Timer Capture event interrupt							
			0 Timer Capture event interrupt disable							
			1 Timer	Capture event	t interrupt enal	ole				

A TELL

7	6	5	4		3	2	1	0
-	PLLCT6	PLLCT5	PLLC	T4 F	PLLCT3	PLLCT2	PLLCT1	PLLCT0
-	RW	RW	RV	V	RW	RW	RW	RW
								Initial value : 42H
	P	LCT[6:5]	Pre Sca	aler (divide	r) Control			
			PLLCT	6 PLLC1	5 des	cription		
			0	0	Div	1		
			0	1	Div	2		
			1	0	Div	4		
			1	1	Div	8		
	P	LLCT[4]	Feedbacl	k Control				
			0 [	Div 64				
			1 C	Div 50				
	PL	LCT[3:1]	Post Sca	ler (divider)	Control (	FOUT=fvco/N	Л)	
			PLLCT3	PLLCT2	PLLC	T1 descri	ption	
			0	0	0	M=1		
			0	0	1	M=2		
			0	1	0	M=4		
			0	1	1	M=5		
			1	0	0	M=6		
			1	0	1	M=8		
			1	1	0	M=10		
			1	1	1	M=16		
	P	LLCT[0]	PLL enab	ole (this bit	should en	able before 1	ms for using	g PLL)
			0 F	PLL disable	•			
			1 F	PLL enable				

PLLCR (PLL Control Register) : 85H

Note) FVCOIN value must be 2 MHz for desire FOUT.





FVCOIN = 2MHz (to be fixed) = FXIN / Pre-Divide

FVCO = FVCOIN \* Feedback-Divider = 100 MHz or 128 MHz FOUT = FVCO / Post-Divider

#### 11.5.2 8-bit Timer/Event Counter 2, 3

#### 11.5.2.1 Overview

Timer 2 and timer 3 can be used either two 8-bit timer/counter or one 16-bit timer/counter with combine them. Each 8-bit timer/event counter module has multiplexer, 8-bit timer data register, 8-bit counter register, mode register, input capture register, comparator. For PWM, it has PWM register (T3PPR, T3PDR, T3PWHR).

It has seven operating modes:

- 8 Bit Timer/Counter Mode
- 8 Bit Capture Mode
- 8 Bit Compare Output Mode
- 16 Bit Timer/Counter Mode
- 16 Bit Capture Mode
- 16 Bit Compare Output Mode
- PWM Mode

The timer/counter can be clocked by an internal or external clock source (external EC2). The clock source is selected by clock select logic which is controlled by the clock select (T2CK[2:0], T3CK[1:0]).

- TIMER2 clock source : fX/1, 2, 4, 64, 256, 1024, 4096, EC2
- TIMER3 clock source : fX/1, 2, 16, T2CK

In the capture mode, by INT2, INT3, the data is captured into Input Capture Register. The Timer 2 outputs the compare result to T2 port in 8/16-bit mode. Also the timer 3 outputs the result T3 port in the timer mode and the PWM waveform to PWM3 in the PWM mode.

16 Bit	CAP2	CAP3	PWM3E	T2CK[2:0]	T3CK[1:0]	T2/3_PE	Timer 2	Timer 3	
0	0	0	0	XXX	XX	00	8 Bit Timer	8 Bit Timer	
0	0	1	0	111	XX	00	8 Bit Event Counter	8 Bit Capture	
0	1	0	0	XXX	XX	01	8 Bit Capture	8 Bit Compare Output	
0	0	0	1	XXX	XX	11	8 Bit Timer/Counter	10 Bit PWM	
1	0	0	0	XXX	11	00	16 Bit Timer		
1	0	0	0	111	11	00	16 Bit Event Counter		
1	1	1	0	XXX	11	00	16 Bit Capture		
1	0	0	0	XXX	11	01	16 Bit Compare Output		

Table 11-8 Operating Modes of Timer

### 11.5.2.2 8-Bit Timer/Counter Mode



The 8-bit Timer/Counter Mode is selected by control registers as shown in Figure 11-26.

Figure 11-26 8 Bit Timer/Event Counter2, 3 Block Diagram

The two 8-bit timers have each counter and data register. The counter register is increased by internal or external clock input. The timer 2 can use the input clock with 2, 4, 16, 64, 256, 1024, 4096 prescaler division rates (T2CK[2:0]). The timer 3 can use the input clock with 1, 2, 16 and timer 2 overflow clock (T3CK[1:0]). When the value of T2, 3value and the value of T2DR, T3DR are respectively identical in Timer 2, 3, the interrupt of timer T2, 3 occurs. The external clock (EC2) counts up the timer at the rising edge. If EC2 is selected from T2CK[2:0], EC2 port becomes input port. The timer 3 can't use the external EC clock.

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Figure 11-27 Timer/Event Counter2, 3 Example



Figure 11-28 Operation Example of Timer/Event Counter2, 3

# 11.5.2.3 16-Bit Timer/Counter Mode

The timer register is being run with all 16bits. A 16-bit timer/counter register T2, T3 are incremented from 0003H to FFFFH until it matches T2DR, T3DR and then resets to 0000H. the match output
generates the Timer 2 interrupt (no timer 3 interrupt). The clock source is selected from T2CK[2:0] and T3CK[1:0] must set 11b and 16BIT bit must set to '1'. The timer 2 is LSB 8-bit, the timer 3 is MSB 8-bit. T2DR must not be 0x00(0x01~0xFF). The 16-bit mode setting is shown as Figure 11-29.



Figure 11-29 16 Bit Timer/Event Counter2, 3 Block Diagram

#### 11.5.2.4 8-Bit Capture Mode

The timer 2, 3 capture mode is set by CAP2, CAP3 as '1'. The clock source can use the internal/external clock. Basically, it has the same function of the 8-bit timer/counter mode and the interrupt occurs at T 2, 3 and T2DR, T3DR matching time, respectively. The capture result is loaded into CDR2, CDR3. The T2, T3 value is automatically cleared by hardware and restarts counter.

This timer interrupt in capture mode is very useful when the pulse width of captured signal is wider than the maximum period of timer.

As the EIEDGE and EIPOLA register setting, the external interrupt INT2, INT3 function is chosen.

The CDR2, T2 and T2DR are in same address. In the capture mode, reading operation is read the CDR2, not T2DR because path is opened to the CDR2. The CDR3 has the same function.

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Figure 11-30 8-bit Capture Mode for Timer2, 3

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Figure 11-31 Input Capture Mode Operation of Timer 2, 3



Figure 11-32 Express Timer Overflow in Capture Mode

# 11.5.2.5 16-Bit Capture Mode

The 16-bit capture mode is the same operation as 8-bit capture mode, except that the timer register uses 16 bits.



The clock source is selected from T2CK[2:0] and T3CK[1:0] must set 11b and 16BIT2 bit must set to '1'. The 16-bit mode setting is shown as Figure 11-33.

Figure 11-33 16-bit Capture Mode of Timer 2, 3

#### 11.5.2.6 PWM Mode

The timer 3 has a PWM (pulse Width Modulation) function. In PWM mode, the T3/PWM3 output pin outputs up to 10-bit resolution PWM output. This pin should be configured as a PWM output by set T3\_PE to '1'. The period of the PWM output is determined by the T3PPR (PWM period register) + T3PWHR[3:2] + T3PWHR[1:0]

PWM Period = [ T3PWHR[3:2]T3PPR ] X Source Clock PWM Duty = [ T3PWHR[1:0] T3PDR ] X Source Clock

Desit for		Frequency	
Resolution	T3CK[1:0]=00 (125ns)	T3CK[1:0]=01 (250ns)	T3CK[1:0]=10 (2us)
10 Bit	7.8KHz	3.9KHz	0.49KHz
9 Bit	15.6KHz	7.8KHz	0.98KHz
8 Bit	31.2KHz	15.6KHz	1.95KHz

Table 11-9 PWM Frequency vs. Resolution at 8 Mhz

- WWW

	7 Bit	62 4KHz	31 2KHz	3 91KHz
		02.41112	51.21(1)2	5.31112

The POL bit of T3CR register decides the polarity of duty cycle. If the duty value is set same to the period value, the PWM output is determined by the bit POL (1: High, 0: Low). And if the duty value is set to "00H", the PWM output is determined by the bit POL (1: Low, 0: High). If duty value and period value are equal, PWM output is not retain high or low but toggle.



Figure 11-34 PWM Mode



Figure 11-35 Example of PWM at 4MHz



Figure 11-36 Example of Changing the Period in Absolute Duty Cycle at 4Mhz

# 11.5.2.7 8-Bit (16-Bit) Compare Output Mode

If the T3 (T2+T3) value and the T3DR (T2DR+T3DR) value are matched, T3/PWM3 port outputs. The output is 50:50 of duty square wave, the frequency is following

$$f_{COMP} = \frac{\text{Oscillator Frequency}}{2 \times \text{Prescaler Value} \times (TDR + 1)}$$

To export the compare output as T3/PWM3, the T3\_PE bit in the T3PWHR register must set to '1'.

#### 11.5.2.8 Register Map

Name	Address	Dir	Default	Description
T2CR	C6	R/W	00H	Timer 2 Mode Control Register
T2	C7	R	00H	Timer 2 Register
T2DR	C7	W	FFH	Timer 2 Data Register
CDR2	C7	R	00H	Capture 2 Data Register
T3CR	CA	R/W	00H	Timer 3 Mode Control Register
T3DR	СВ	W	FFH	Timer 3 Data Register
T3PPR	СВ	W	FFH	Timer 3 PWM Period Register
Т3	СС	R	00H	Timer 3 Register
T3PDR	CC	R/W	00H	Timer 3 PWM Duty Register
CDR3	CC	R	00H	Capture 3 Data Register
T3PWHR	CD	W	00H	Timer 3 PWM High Register

Table 11-10 Register Map

# 11.5.2.9 Timer/Counter 2, 3 Register description

The Timer/Counter 2, 3 Register consists of Timer 2 Mode Control Register (T2CR), Timer 2 Register (T2), Timer 2 Data Register (T2DR), Capture 2 Data Register (CDR2), Timer 3 Mode Control Register (T3CR), Timer 3 Data Register (T3DR), Timer 3 PWM Period Register (T3PPR), Timer 3 Register (T3), Timer 3 PWM Duty Register (T3PPR), Capture 3 Data Register (CDR3) and Timer 3 PWM High Register (T3PWHR).

# 11.5.2.10 Register description for Timer/Counter 2, 3

7	6	5		4	3	2	1	0	
T2EN	T2_PE	CAP2	T2	CK2	T2CK1	T2CK0	T2CN	T2ST	
RW	RW	RW	R	WV.	RW	RW	RW	RW	
							I	nitial value : 00H	
		T2EN	Control	Timer 2					
			0	Timer 2 d	isable				
			1	Timer 2 e	nable				
		T2_PE	Control	Timer 2 O	utput port				
			0	Timer 2 O	output disab	ble			
			1	Timer 2 O	output enab	le			
		CAP2	Control	-	peration mo				
			0	Timer/Cou	unter mode				
			1	Capture n	node				
	T2	2CK[2:0]	Select T	Timer 2 clo	ck source.	Fx is main sys	stem clock fre	quency	
				T2CK1	T2CK0	Description			
				0	0	fx/2			
			0	0	1	fx/4			
			0	1	0	fx/16			
			0	1	1	fx/64			
			1	0	0	fx/256			
			1	0	1	fx/1024			
			1	1	0	fx/4096			
			1	1	1	External Clo	ock (EC2)		
		T2CN			ount pause				
			0 Temporary count stop						
			1 Continue count						
		T2ST		Timer 2 st	-				
			0	Counter s	-				
			1	Clear cou	nter and sta	art			

# T2CR (Timer 2 Mode Control Register) : C6H

# T2 (Timer 2 Register: Read Case) : C7H

7	6	5	4	3	2	1	0	
T27	T26	T25	T24	T23	T22	T21	T20	
R	R	R	R	R	R	R	R	
						I	nitial value : 0	0H

T2[7:0] T2 Counter data

Industrial to the

# T2DR (Timer 2 Data Register: Write Case) : C7H

7	6	5	4	3	2	1	0	
T2D7	T2D6	T2D5	T2D4	T2D3	T2D2	T2D1	T2D0	
W	W	W	W	W	W	W	W	
						li	nitial value : FF	۶H

T2D[7:0] T2 Compare data

# CDR2 (Capture 2 Data Register: Read Case) : C7H

7	6	5	4	3	2	1	0	
CDR27	CDR26	CDR25	CDR24	CDR23	CDR22	CDR21	CDR20	
R	R	R	R	R	R	R	R	
						I	nitial value : 0	)0H

CDR2[7:0] T2 Capture data

# T3CR (Timer 3 Mode Count Register) : CAH

7	6	5	4	3	2	1	0	
POL	16BIT	PWM3E	CAPS	3 T3CK1	T3CK0	T3CN	T3ST	
RW	RW	RW	RW	RW	RW	RW	RW	
							Initial value : 0	
		POL	Configure	PWM polarity				
			0 N	egative (Duty Ma	atch: Clear)			
			1 P	ositive (Duty Mat	ch: Set)			
		16BIT	Select Tim	ner 1 8/16Bit				
			0 8	Bit				
			1 10	6 Bit				
	I	PWM3E	Control P	VM enable				
			0 P	WM disable				
			1 P	WM enable				
	CAP3 Control Timer 3 mode							
0 Timer/Counter mode								
				apture mode				
	Т	'3CK[1:0]		ock source of Tin		requency of m	ain system.	
			T3CK1		Description			
			0	-	x			
			0		x/2			
			1	-	x/16			
	1 1 Use Timer 2 Clock							
	Note		you want to use "Using Timer 2 Clock", you can set T2EN bit in T2CR					
	T3CN Control Timer 3 Count pause/continue							
0 Temporary count stop								
		1 Continue count						
		T3ST		mer 3 start/stop				
				ounter stop				
			1 C	lear counter and	start			

# T3DR (Timer 3 Data Register: Write Case) : CBH

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7	c	E	4	2	2	4	0
/	0	3	4	ა 	2	1	U
T3D7	T3D6	T3D5	T3D4	T3D3	T3D2	T3D1	T3D0
W	W	W	W	W	W	W	W
						I	nitial value : FF

T3D[7:0] T3 Compare data

#### T3PPR (Timer 3 PWM Period Register: Write Case PWM mode only) : CBH

7	6	5	4	3	2	1	0	
T3PP7	T3PP6	T3PP5	T3PP4	T3PP3	T3PP2	T3PP1	T3PP0	
W	W	W	W	W	W	W	W	

Initial value : FFH

T3PP[7:0] T3 PWM Period data

# T3 (Timer 3 Register: Read Case) : CCH

_	7	6	5	4	3	2	1	0	_
	T37	T36	T35	T34	T33	T32	T31	T30	
	R	R	R	R	R	R	R	R	
							I	nitial value : 0	0H

T3[7:0] T3 Counter Period data

### T3PDR (Timer 3 PWM Duty Register) : CCH

7	6	5	4	3	2	1	0
T3PD7	T3PD6	T3PD5	T3PD4	T3PD3	T3PD2	T3PD1	T3PD0
RW							

Initial value : 00H

T3PD[7:0]

 T3 PWM Duty data Note) only write, when PWM3E '1'

### CDR3 (Capture 3 Data Register: Read Case) : CCH

7	6	5	4	3	2	1	0
CDR37	CDR36	CDR35	CDR34	CDR33	CDR32	CDR31	CDR30
R	R	R	R	R	R	R	R
						I	nitial value : 00H

CDR3[7:0] T3 Capture data

#### T3PWHR (Timer 3 PWM High Register) : CDH

7	6	5	4	3	2	1	0	
T3_PE	-	-	-	PW3H3	PW3H2	PW3H1	PW3H0	
W	-	-	-	W	W	W	W	
						I	nitial value : 00	
T3_PE Control Timer 3 Output port operation   Note) only writable Bit. Be careful   0 Timer 3 Output disable   1 Timer 3 Output enable								
	Р	W3H[3:2]	PWM period F	ligh value (Bit	[9:8])			
PS029602-02	212		PRELI	MINARY			11	

**PW3H[1:0]** PWM duty High value (Bit [9:8])

PERIOD:	PW3H3	PW3H2	T3PPR[7:0]
DUTY:	PW3H1	PW3H0	T3PDR[7:0]

# 11.5.3 16-Bit Timer 4

# 11.5.3.1 Overview

The 16-bit timer 4 consists of Multiplexer, Timer Data Register High/Low, Timer Register High/Low, Timer Mode Control Register. It is able to use internal 16-bit timer/ counter without a port output function.

The 16-bit timer 4 is able to use the divided clock of the main clock selected from pre-scalar output.

# 11.5.3.2 16 Bit Timer/Counter Mode



Figure 11-37 Timer4 16-bit Mode Block Diagram

#### 11.5.3.3 Register Map

Name	Address	Dir	Default	Description
T4CR	0xCE	R/W	00H	Timer 4 Mode Control Register
T4L	0xCF	R	00H	Timer 4 Low Register
T4LDR	0xCF	W	FFH	Timer 4 Low Data Register
LCDR4	0xCF	R	00H	Low Capture 4 Data Register
T4H	0xD5	R	00H	Timer 4 High Register
T4HDR	0xD5	R/W	00H	Timer 4 High Data Register
HCDR4	0xD5	R	00H	High Capture 4 Data Register

Table 11-11 Register Map

# 11.5.3.4 Timer 4 Register description

The timer 4 register consists of Timer 4 Mode Control Register (T4CR), Timer 4 Low Register (T4L), Timer 4 Low Data Register (T4LDR), Low Capture 4 Data Register (LCDR4), Timer 4 High Register (T4H), Timer 4 High Data Register (T4HDR), High Capture 4 Data Register (HCDR4).

# 11.5.3.5 Register description for Timer 4

### T4CR (Timer 4 Mode Control Register) : CEH

7	6	5		4	3	2	1	0
T4EN	-	CAP4	Ţ	4CK2	T4CK1	T4CK0	T4CN	T4ST
RW	-	RW	1	RW	RW	RW	RW	RW
								Initial value : 00
		T4EN	Contro	Timer 4 op	eration			
			0	Timer 4 di	sable			
			1	Timer 4 e	nable			
		CAP4	Contro	Timer 4 m	ode			
			0	Timer/Cou	unter mode			
			1	Capture n	node			
		T4CK[2:0]	Select	Timer 4 clo	ck source.	fx is main syst	em clock fred	quency
				T4CK1	T4CK0	Description		
			0	0	0	fx/2		
			0	0	1	fx/4		
			0	1	0	fx/8		
			0	1	1	fx/16		
			1	0	0	fx/64		
			1	0	1	fx/256		
			1	1	0	fx/1024		
			1	1	1	fx/2048		
		T4CN	Contro	Timer 4 Co	ount pause	/continue		
			0	Temporar	y count sto	р		
			1	Continue	count			
		T4ST	Contro	Timer 4 sta	art/stop			
000000 0040	-							4.4



1 Clear Counter and start

#### T4L (Timer 4 Low Register: Read Case) : CFH



T4L[7:0] T4L Counter

#### T4LDR (Timer 4 Low Data Register: Write Case) : CFH

7	6	5	4	3	2	1	0
T4LD7	T4LD6	T4LD5	T4LD4	T4LD3	T4LD2	T4LD1	T4LD0
W	W	W	W	W	W	W	W
						l.	nitial value : FF

T4LD[7:0] T4L Compare

#### LCDR4 (Low Capture 4 Data Register: Read Case) : CFH

7	6	5	4	3	2	1	0	
LCDR47	LCDR46	LCDR45	LCDR44	LCDR43	LCDR42	LCDR41	LCDR40	
R	R	R	R	R	R	R	R	
						I	nitial value : 0	OН

LCDR4[7:0] T4L Capture data

### T4H (Timer 4 High Register: Read Case) : D5H

7	6	5	4	3	2	1	0
T4H7	T4H6	T4H5	T4H4	T4H3	T4H2	T4H1	T4H0
R	R	R	R	R	R	R	R
						I	nitial value : 00H

T4H[7:0] T4H Counter Period

### T4HDR (Timer 4 High Data Register: Write Case) : D5H

7	6	5	4	3	2	1	0	
T4HD7	T4HD6	T4HD5	T4HD4	T4HD3	T4HD2	T4HD1	T4HD0	
W	W	W	W	W	W	W	W	
							nitial value : Fl	FH

T4HD[7:0] T4H Compare

#### HDR4 (High Capture 4 Data Register: Read Case) : D5H

7	6	5	4	3	2	1	0
HCDR47	HCDR46	HCDR45	HCDR44	HCDR43	HCDR42	HCDR41	HCDR40
R	R	R	R	R	R	R	R
							nitial value + 00L

Initial value : 00H

HCDR4[7:0] T4H Capture data

#### 11.5.4 Timer Interrupt Status Register (TMISR)

#### 11.5.4.1 Register description for TMISR

# TMISR (Timer Interrupt Status Register) : D5H

7	6	5	4	3	2	1	0
-	-	TMIF5	TMIF4	TMF3	TMIF2	TMIF1	TMIFO
-	-	R	R	R	R	R	R
						I	nitial value : 00
		TMIF5	Timer 5 Interru	pt Flag			
			0 No Tin	ner 5 interrupt			
			1 Timer	5 interrupt occ	urred, write "1	" to clear inter	rrupt flag
		TMIF4	Timer 4 Interru	pt Flag			
			0 No Tin	ner 4 interrupt			
			1 Timer	4 interrupt occ	urred, write "1	" to clear inter	rrupt flag
		TMIF3	Timer 3 Interru	pt Flag			
			0 No Tin	ner 3 interrupt			
			1 Timer	3 interrupt occ	curred, write "1	" to clear inter	rrupt flag
		TMIF2	Timer 2 Interru	pt Flag			
			0 No Tin	ner 2 interrupt			
			1 Timer	2 interrupt occ	urred, write "1	" to clear inter	rrupt flag
		TMIF1	Timer 1 Interru	pt Flag			
			0 No Tin	ner 1 interrupt			
			1 Timer	1 interrupt occ	curred, write "1	" to clear inter	rrupt flag
		TMIF0	Timer 0 Interru	pt Flag			
			0 No Tin	ner 0 interrupt			
			1 Timer	0 interrupt occ	urred, write "1	" to clear inter	rrupt flag

Note) The Timer Interrupt Status Register contains interrupt information of each timers. Even if user disabled timer interrupt at IE2, user could check timer interrupt condition from this register.

# 11.6 Buzzer Driver

### 11.6.1 Overview

The Buzzer consists of 8 Bit Counter and BUZDR (Buzzer Data Register), BUZCR (Buzzer Control Register). The Square Wave (61.035Hz~125 KHz, @8MHz) gets out of P12/BUZ pin. BUZDR (Buzzer Data Register) controls the Buzzer frequency (look at the following expression). In the BUZCR (Buzzer Control Register), BUCK[1:0] selects source clock divided from prescaler.

 $f_{BUZ}(Hz) = \frac{\text{Oscillator Frequency}}{2 \times \text{Prescaler Ratio} \times (BUZDR + 1)}$ 

	Buzzer Frequency (kHz)								
BUZDR[7:0] BUZCR[2:1]=00		BUZCR[2:1]=01	BUZCR[2:1]=10	BUZCR[2:1]=11					
0000_0000	125kHz	62.5kHz	31.25kHz	15.625kHz					
0000_0001	62.5kHz	31.25kHz	15.625kHz	7.812kHz					
1111_1101	492.126Hz	246.063Hz	123.031Hz	61.515Hz					
1111_1110	490.196Hz	245.098Hz	122.549Hz	61.274Hz					
1111_1111	488.281Hz	244.141Hz	122.07Hz	61.035Hz					

### 11.6.2 Block Diagram



Figure 11-38 Buzzer Driver Block Diagram

#### 11.6.3 Register Map

Table 11-13 Register Map

Name	Address	Dir	Default	Description	
BUZDR	8FH	R/W	FFH	Buzzer Data Register	
BUZCR	9FH	R/W	00H	Buzzer Control Register	

#### 11.6.4 Buzzer Driver Register description

Buzzer Driver consists of Buzzer Data Register (BUZDR), Buzzer Control Register (BUZCR).

#### 11.6.5 Register description for Buzzer Driver

#### BUZDR (Buzzer Data Register) : 8FH

7	6	5	4	3	2	1	0
BUZDR7	BUZDR6	BUZDR5	BUZDR4	BUZDR3	BUZDR2	BUZDR1	BUZDR0
RW							
							nitial value : FF

**BUZDR[7:0]** This bits control the Buzzer frequency Its resolution is 00H ~ FFH

#### **BUZCR (Buzzer Control Register) : 9FH**

7	6	5	4	3	2	1	0
-	-	-	-	-	BUCK1	BUCK0	BUZEN
-	-	-	-	-	RW	RW	RW
						I	nitial value : 00H
	В	UCK[1:0]	Buzzer Drive	Source Clock	Selection		
			BUCK1 BL	ICK0 Source	e Clock		
			0 0	fx/32			
			0 1	fx/64			
			1 0	fx/128			
			1 1	fx/256			
		BUZEN	Buzzer Driver	Operation Co	ntrol		
			0 Buzzer Driver disable				
			1 Bu	zzer Driver en	able		
			Note) fx: Mai	n system clock	oscillation fre	quency	

# 11.7 USART

#### 11.7.1 Overview

The Universal Synchronous and Asynchronous serial Receiver and Transmitter (USART) is a highly flexible serial communication device. The main features are listed below.

- Full Duplex Operation (Independent Serial Receive and Transmit Registers)
- Asynchronous or Synchronous Operation
- Master or Slave Clocked Synchronous and SPI Operation
- Supports all four SPI Modes of Operation (Mode 0, 1, 2, 3)
- LSB First or MSB First Data Transfer @SPI mode
- High Resolution Baud Rate Generator
- Supports Serial Frames with 5,6,7,8, or 9 Data Bits and 1 or 2 Stop Bits
- Odd or Even Parity Generation and Parity Check Supported by Hardware
- Data OverRun Detection
- Framing Error Detection
- Digital Low Pass Filter
- Three Separate Interrupts on TX Complete, TX Data Register Empty and RX Complete
- Double Speed Asynchronous Communication Mode

USART has three main parts of Clock Generator, Transmitter and Receiver. The Clock Generation logic consists of synchronization logic for external clock input used by synchronous or SPI slave operation, and the baud rate generator for asynchronous or master (synchronous or SPI) operation. The Transmitter consists of a single write buffer, a serial shift register, parity generator and control logic for handling different serial frame formats. The write buffer allows a continuous transfer of data without any delay between frames. The receiver is the most complex part of the USART module due to its clock and data recovery units. The recovery unit is used for asynchronous data reception. In addition to the recovery unit, the Receiver includes a parity checker, a shift register, a two level receive FIFO (UDATAx) and control logic. The Receiver supports the same frame formats as the Transmitter and can detect Frame Error, Data OverRun and Parity Errors.

### 11.7.2 Block Diagram



Figure 11-39 USART Block Diagram

### 11.7.3 Clock Generation



Figure 11-40 Clock Generation Block Diagram

The Clock generation logic generates the base clock for the Transmitter and Receiver. The USART supports four modes of clock operation and those are Normal Asynchronous, Double Speed Asynchronous, Master Synchronous and Slave Synchronous. The clock generation scheme for Master SPI and Slave SPI mode is the same as Master Synchronous and Slave Synchronous operation mode. The UMSELn bit in UCTRLx1 register selects between asynchronous and synchronous operation. Asynchronous Double Speed mode is controlled by the U2X bit in the UCTRLx2 register. The MASTER bit in UCTRLx2 register controls whether the clock source is internal (Master mode, output port) or external (Slave mode, input port). The XCK pin is only active when the USART operates in Synchronous or SPI mode.

Table below contains equations for calculating the baud rate (in bps).

Operating Mode	Equation for Calculating Baud Rate
Asynchronous Normal Mode (U2X=0)	Baud Rate = $\frac{\text{fSCLK}}{16(\text{UBAUDx} + 1)}$
Asynchronous Double Speed Mode (U2X=1)	Baud Rate = $\frac{\text{fSCLK}}{8(\text{UBAUDx} + 1)}$
Synchronous or SPI Master Mode	Baud Rate = $\frac{\text{fSCLK}}{2(\text{UBAUDx} + 1)}$

		<u> </u>		
1 able 11-14 Ec	auations for	Calculating	Baud Rate	Register Setting
	Juanonio ioi	Calculating	Dada Halo	riogiotor ootanig

### 11.7.4 External Clock (XCK)

External clocking is used by the synchronous or spi slave modes of operation.

External clock input from the XCK pin is sampled by a synchronization logic to remove meta-stability. The output from the synchronization logic must then pass through an edge detector before it can be used by the Transmitter and Receiver. This process introduces a two CPU clock period delay and therefore the maximum frequency of the external XCK pin is limited by the following equation.

$$fXCK = \frac{fSCLK}{4}$$

where fXCK is the frequency of XCK and fSCLK is the frequency of main system clock (SCLK).

# 11.7.5 Synchronous mode Operation

When synchronous or spi mode is used, the XCK pin will be used as either clock input (slave) or clock output (master). The dependency between the clock edges and data sampling or data change is the same. The basic principle is that data input on RXD (MISO in spi mode) pin is sampled at the opposite XCK clock edge of the edge in the data output on TXD (MOSI in spi mode) pin is changed.

The UCPOL bit in UCTRLx1 register selects which XCK clock edge is used for data sampling and which is used for data change. As shown in the figure below, when UCPOL is zero the data will be changed at rising XCK edge and sampled at falling XCK edge.



Figure 11-41 Synchronous Mode XCKn Timing

#### 11.7.6 Data format

A serial frame is defined to be one character of data bits with synchronization bits (start and stop bits), and optionally a parity bit for error checking.

The USART supports all 30 combinations of the following as valid frame formats.

- 1 start bit
- 5, 6, 7, 8 or 9 data bits
- no, even or odd parity bit
- 1 or 2 stop bits

A frame starts with the start bit followed by the least significant data bit (LSB). Then the next data bits, up to a total of nine, are succeeding, ending with the most significant bit (MSB). If enabled the parity bit is inserted after the data bits, before the stop bits. A high to low transition on data pin is considered as start bit. When a complete frame is transmitted, it can be directly followed by a new frame, or the communication line can be set to an idle state. The idle means high state of data pin. The next figure shows the possible combinations of the frame formats. Bits inside brackets are optional.



Figure 11-42 frame format

1 data frame consists of the following bits

- Idle No communication on communication line (TxD/RxD)
- St Start bit (Low)
- Dn Data bits (0~8)
- Parity bit ----- Even parity, Odd parity, No parity
- Stop bit(s) ----- 1 bit or 2 bits

The frame format used by the USART is set by the USIZE[2:0], UPM[1:0] and USBS bits in UCTRLx1 register. The Transmitter and Receiver use the same setting.

# 11.7.7 Parity bit

The parity bit is calculated by doing an exclusive-or of all the data bits. If odd parity is used, the result of the exclusive-or is inverted. The parity bit is located between the MSB and first stop bit of a serial frame.

 $P_{even} = D_{n-1}^{-1} \dots^{-1} D_3^{-1} D_2^{-1} D_1^{-1} D_0^{-1} O_0^{-1}$  $P_{odd} = D_{n-1}^{-1} \dots^{-1} D_3^{-1} D_2^{-1} D_1^{-1} D_0^{-1} O_0^{-1}$ 

Peven : Parity bit using even parity

 $P_{odd}$ : Parity bit using odd parity

D<sub>n</sub> : Data bit n of the character

# 11.7.8 USART Transmitter

The USART Transmitter is enabled by setting the TXE bit in UCTRLx1 register. When the Transmitter is enabled, the normal port operation of the TXD pin is overridden by the serial output pin of USART. The baud-rate, operation mode and frame format must be setup once before doing any transmissions. If synchronous or spi operation is used, the clock on the XCK pin will be overridden and used as transmission clock. If USART operates in spi mode, SS pin is used as SS input pin in slave mode or can be configured as SS output pin in master mode. This can be done by setting SPISS bit in UCTRLx3 register.

# 11.7.8.1 Sending Tx data

A data transmission is initiated by loading the transmit buffer (UDATAx register I/O location) with the data to be transmitted. The data written in transmit buffer is moved to the shift register when the shift register is ready to send a new frame. The shift register is loaded with the new data if it is in idle state or immediately after the last stop bit of the previous frame is transmitted. When the shift register is loaded with new data, it will transfer one complete frame at the settings of control registers. If the 9-bit characters are used in asynchronous or synchronous operation mode (USIZE[2:0]=7), the ninth bit must be written to the TX8 bit in UCTRLx3 register before loading transmit buffer (UDATA register).

# 11.7.8.2 Transmitter flag and interrupt

The USART Transmitter has 2 flags which indicate its state. One is USART Data Register Empty (UDRE) and the other is Transmit Complete (TXC). Both flags can be interrupt sources.

UDRE flag indicates whether the transmit buffer is ready to be loaded with new data. This bit is set when the transmit buffer is empty and cleared when the transmit buffer contains data to be transmitted that has not yet been moved into the shift register. And also this flag can be cleared by writing '0' to this bit position. Writing '1' to this bit position is prevented.

When the Data Register Empty Interrupt Enable (UDRIE) bit in UCTRL2 register is set and the Global Interrupt is enabled, USART Data Register Empty Interrupt is generated while UDRE flag is set.

The Transmit Complete (TXC) flag bit is set when the entire frame in the transmit shift register has been shifted out and there are no more data in the transmit buffer. The TXC flag is automatically cleared when the Transmit Complete Interrupt service routine is executed, or it can be cleared by writing '0' to TXC bit in USTAT register.

When the Transmit Complete Interrupt Enable (TXCIE) bit in UCTRL2 register is set and the Global Interrupt is enabled, USART Transmit Complete Interrupt is generated while TXC flag is set.

#### 11.7.8.3 Parity Generator

The Parity Generator calculates the parity bit for the sending serial frame data. When parity bit is enabled (UPM[1]=1), the transmitter control logic inserts the parity bit between the MSB and the first stop bit of the sending frame.

#### 11.7.8.4 Disabling Transmitter

Disabling the Transmitter by clearing the TXE bit will not become effective until ongoing transmission is completed. When the Transmitter is disabled, the TXD pin is used as normal General Purpose I/O (GPIO) or primary function pin.

#### 11.7.9 USART Receiver

The USART Receiver is enabled by setting the RXE bit in the UCTRLx1 register. When the Receiver is enabled, the normal pin operation of the RXD pin is overridden by the USART as the serial input pin of the Receiver. The baud-rate, mode of operation and frame format must be set before serial reception. If synchronous or spi operation is used, the clock on the XCK pin will be used as transfer clock. If USART operates in spi mode, SS pin is used as SS input pin in slave mode or can be configured as SS output pin in master mode. This can be done by setting SPISS bit in UCTRLx3 register.

#### 11.7.9.1 Receiving Rx data

When USART is in synchronous or asynchronous operation mode, the Receiver starts data reception when it detects a valid start bit (LOW) on RXD pin. Each bit after start bit is sampled at predefined baud-rate (asynchronous) or sampling edge of XCK (synchronous), and shifted into the receive shift register until the first stop bit of a frame is received. Even if there's 2<sup>nd</sup> stop bit in the frame, the 2<sup>nd</sup> stop bit is ignored by the Receiver. That is, receiving the first stop bit means that a complete serial frame is present in the receiver shift register and contents of the shift register are to be moved into the receive buffer. The receive buffer is read by reading the UDATAx register.

If 9-bit characters are used (USIZE[2:0] = 7) the ninth bit is stored in the RX8 bit position in the UCTRLx3 register. The 9<sup>th</sup> bit must be read from the RX8 bit before reading the low 8 bits from the UDATAx register. Likewise, the error flags FE, DOR, PE must be read before reading the data from UDATAx register. This is because the error flags are stored in the same FIFO position of the receive buffer.

#### 11.7.9.2 Receiver flag and interrupt

The USART Receiver has one flag that indicates the Receiver state.

The Receive Complete (RXC) flag indicates whether there are unread data present in the receive buffer. This flag is set when there are unread data in the receive buffer and cleared when the receive

buffer is empty. If the Receiver is disabled (RXE=0), the receiver buffer is flushed and the RXC flag is cleared.

When the Receive Complete Interrupt Enable (RXCIE) bit in the UCTRLx2 register is set and Global Interrupt is enabled, the USART Receiver Complete Interrupt is generated while RXC flag is set.

The USART Receiver has three error flags which are Frame Error (FE), Data OverRun (DOR) and Parity Error (PE). These error flags can be read from the USTATx register. As data received are stored in the 2-level receive buffer, these error flags are also stored in the same position of receive buffer. So, before reading received data from UDATAx register, read the USTATx register first which contains error flags.

The Frame Error (FE) flag indicates the state of the first stop bit. The FE flag is zero when the stop bit was correctly detected as one, and the FE flag is one when the stop bit was incorrect, ie detected as zero. This flag can be used for detecting out-of-sync conditions between data frames.

The Data OverRun (DOR) flag indicates data loss due to a receive buffer full condition. A DOR occurs when the receive buffer is full, and another new data is present in the receive shift register which are to be stored into the receive buffer. After the DOR flag is set, all the incoming data are lost. To prevent data loss or clear this flag, read the receive buffer.

The Parity Error (PE) flag indicates that the frame in the receive buffer had a Parity Error when received. If Parity Check function is not enabled (UPM[1]=0), the PE bit is always read zero.

Note) The error flags related to receive operation are not used when USART is in SPI mode.

#### 11.7.9.3 Parity Checker

If Parity Bit is enabled (UPM[1]=1), the Parity Checker calculates the parity of the data bits in incoming frame and compares the result with the parity bit from the received serial frame.

#### 11.7.9.4 Disabling Receiver

In contrast to Transmitter, disabling the Receiver by clearing RXE bit makes the Receiver inactive immediately. When the Receiver is disabled the Receiver flushes the receive buffer and the remaining data in the buffer is all reset. The RXD pin is not overridden the function of USART, so RXD pin becomes normal GPIO or primary function pin.

#### 11.7.9.5 Asynchronous Data Reception

To receive asynchronous data frame, the USART includes a clock and data recovery unit. The Clock Recovery logic is used for synchronizing the internally generated baud-rate clock to the incoming asynchronous serial frame on the RXD pin.

The Data recovery logic samples and low pass filters the incoming bits, and this removes the noise of RXD pin.

The next figure illustrates the sampling process of the start bit of an incoming frame. The sampling rate is 16 times the baud-rate for normal mode, and 8 times the baud rate for Double Speed mode (U2X=1). The horizontal arrows show the synchronization variation due to the asynchronous sampling process. Note that larger time variation is shown when using the Double Speed mode.



Figure 11-43 Start Bit Sampling

When the Receiver is enabled (RXE=1), the clock recovery logic tries to find a high to low transition on the RXD line, the start bit condition. After detecting high to low transition on RXD line, the clock recovery logic uses samples 8,9, and 10 for Normal mode, and samples 4, 5, and 6 for Double Speed mode to decide if a valid start bit is received. If more than 2 samples have logical low level, it is considered that a valid start bit is detected and the internally generated clock is synchronized to the incoming data frame. And the data recovery can begin. The synchronization process is repeated for each start bit.

As described above, when the Receiver clock is synchronized to the start bit, the data recovery can begin. Data recovery process is almost similar to the clock recovery process. The data recovery logic samples 16 times for each incoming bits for Normal mode and 8 times for Double Speed mode. And uses sample 8, 9, and 10 to decide data value for Normal mode, samples 4, 5, and 6 for Double Speed mode. If more than 2 samples have low levels, the received bit is considered to a logic 0 and more than 2 samples have high levels, the received bit is considered to a logic 1. The data recovery process is then repeated until a complete frame is received including the first stop bit. The decided bit value is stored in the receive shift register in order. Note that the Receiver only uses the first stop bit of a frame. Internally, after receiving the first stop bit, the Receiver is in idle state and waiting to find start bit.



Figure 11-44 Sampling of Data and Parity Bit

The process for detecting stop bit is like clock and data recovery process. That is, if 2 or more samples of 3 center values have high level, correct stop bit is detected, else a Frame Error flag is set. After deciding first stop bit whether a valid stop bit is received or not, the Receiver goes idle state and monitors the RXD line to check a valid high to low transition is detected (start bit detection).



### 11.7.10 SPI Mode

The USART can be set to operate in industrial standard SPI compliant mode. The SPI mode has the following features.

- Full duplex, three-wire synchronous data transfer
- Master or Slave operation
- Supports all four SPI modes of operation (mode0, 1, 2, and 3)
- Selectable LSB first or MSB first data transfer
- Double buffered transmit and receive
- Programmable transmit bit rate

When SPI mode is enabled (UMSEL[1:0]=3), the Slave Select (SS) pin becomes active low input in slave mode operation, or can be output in master mode operation if SPISS bit is set.

Note that during SPI mode of operation, the pin RXD is renamed as MISO and TXD is renamed as MOSI for compatibility to other SPI devices.

### 11.7.10.1 SPI Clock formats and timing

To accommodate a wide variety of synchronous serial peripherals from different manufacturers, the USART has a clock polarity bit (UCPOL) and a clock phase control bit (UCPHA) to select one of four clock formats for data transfers. UCPOL selectively insert an inverter in series with the clock. UCPHA chooses between two different clock phase relationships between the clock and data. Note that UCPHA and UCPOL bits in UCTRLx1 register have different meanings according to the UMSEL[1:0] bits which decides the operating mode of USART.

Table below shows four combinations of UCPOL and UCPHA for SPI mode 0, 1, 2, and 3.

Table 11-15 CPOL Funtionality

SPI Mode	UCPOL	UCPHA	Leading Edge	Trailing Edge
0	0	0	Sample (Rising)	Setup (Falling)
1	0	1	Setup (Rising)	Sample (Falling)
2	1	0	Sample (Falling)	Setup (Rising)
3	1	1	Setup (Falling)	Sample (Rising)



Figure 11-46 SPI Clock Formats when UCPHA=0

When UCPHA=0, the slave begins to drive its MISO output with the first data bit value when SS goes to active low. The first XCK edge causes both the master and the slave to sample the data bit value on their MISO and MOSI inputs, respectively. At the second XCK edge, the USART shifts the second data bit value out to the MOSI and MISO outputs of the master and slave, respectively. Unlike the case of UCPHA=1, when UCPHA=0, the slave's SS input must go to its inactive high level between transfers. This is because the slave can prepare the first data bit when it detects falling edge of SS input.







When UCPHA=1, the slave begins to drive its MISO output when SS goes active low, but the data is not defined until the first XCK edge. The first XCK edge shifts the first bit of data from the shifter onto the MOSI output of the master and the MISO output of the slave. The next XCK edge causes both the master and slave to sample the data bit value on their MISO and MOSI inputs, respectively. At the third XCK edge, the USART shifts the second data bit value out to the MOSI and MISO output of the master and slave respectively. When UCPHA=1, the slave's SS input is not required to go to its inactive high level between transfers.

Because the SPI logic reuses the USART resources, SPI mode of operation is similar to that of synchronous or asynchronous operation. An SPI transfer is initiated by checking for the USART Data Register Empty flag (UDRE=1) and then writing a byte of data to the UDATA Register. In master mode of operation, even if transmission is not enabled (TXE=0), writing data to the UDATA register is necessary because the clock XCK is generated from transmitter block.

### 11.7.11 Register Map

Table 11-16 Register Map

Name	Address	Dir	Default	Description
UCTRL01	E2H	R/W	00H	USART Control 1 Register 0
UCTRL02	E3H	R/W	00H	USART Control 2 Register 0
UCTRL03	E4H	R/W	00H	USART Control 3 Register 0
USTAT0	E5H	R	80H	USART Status Register 0
UBAUD0	E6H	R/W	FFH	USART Baud Rate Generation Register 0
UDATA0	E7H	R/W	FFH	USART Data Register 0
UCTRL11	FAH	R/W	00H	USART Control 1 Register 1
UCTRL12	FBH	R/W	00H	USART Control 2 Register 1
UCTRL13	FCH	R/W	00H	USART Control 3 Register 1
USTAT1	FDH	R	80H	USART Status Register 1
UBAUD1	FEH	R/W	FFH	USART Baud Rate Generation Register 1
UDATA1	FFH	R/W	FFH	USART Data Register 2

# 11.7.12 USART Register description

USART module consists of USART Control 1 Register (UCTRLx1), USART Control 2 Register (UCTRLx2), USART Control 3 Register (UCTRLx3), USART Status Register (USTATx), USART Data Register (UDATAx), and USART Baud Rate Generation Register (UBAUDx).

# 11.7.13 Register description for USART

7	6	5	4		3	2	1	0
UMSEL1	UMSEL0	UPM1	UPIV	0	USIZE2	USIZE1 UDORD	USIZEO UCPHA	UCPOL
RW	RW	RW	RM	/	RW	RW	RW	RW
								Initial value : 00
	UN	/ISEL[1:0]	Selects of	operation r	node of L	ISART		
			UMSEL1	UMSE	•	erating Mode		
			0	0	-	nchronous M		
			0	1	-	nchronous Mo	de (Synchron	ous Uart)
			1	0		served		
			1	1	_	I Mode		
	L	JPM[1:0]		-		ind Check me	thods	
			UPM1	UPM0		y mode		
			0	0	No P			
			0 1	1 0	Rese	Parity		
			1	1		Parity		
	US	SIZE[2:0]		asynchror	nous or s	,	node of opera	ation, selects the
			USIZE2	USIZE1		0 Data ler	ath	
			0	0	0	5 bit	5	
			0	0	1	6 bit		
			0	1	0	7 bit		
			0	1	1	8 bit		
			1	0	0	Reserve	d	
			1	0	1	Reserve	ed	
			1	1	0	Reserve	d	
			1	1	1	9 bit		
	ι	JDORD	one the M LSB of the	ISB of the e data byte	e data by			node, when set to n set to zero the
			-	SB First ISB First				
		JCPOL			CK in evr	nchronous or s	ni mode	
	L L		-	-	-	g Edge, RXD	-	lling Edge
				-		ng Edge, RXD	-	
	U	ICPHA	This bit is with UCPC synchrono and trailing pulse. And	in the sar DL bit, sele us serial p g edge me I Sample	me bit pos ects one c peripheral eans 2 <sup>nd</sup> means d	sition with US of two clock fo s. Leading ed or last clock etecting of ind	IZE0. In SPI rmats for diffe ge means fir edge of XCK	mode, along erent kinds of st XCK edge in one XCK
			means pre				Troiling	Idao
				UCPH		ading Edge	Trailing E	-
			0 0	0 1		mple (Rising) up (Rising)	Setup (F Sample (	
			1	0		mple (Falling)	Sample ( Setup (R	
			1	1		up (Falling)	Setup (R Sample (	
			1	I	361	ωρ (i annig)	Sample	(tong)

# UCTRLx1 (USART Control 1 Register) E2H, FAH

# UCTRL2 (USART Control 2 Register) E3H, FBH

7	6	5	4	3	2	1	0
UDRIE	TXCIE	RXCIE	WAKEIE	TXE	RXE	USARTEN	U2X
RW	RW	RW	RW	RW	RW	RW	RW
							Initial value : 00
		UDRIE	Interrupt enable	e bit for USAR	T Data Regist	er Empty.	
			0 Interru	pt from UDRE	is inhibited (u	se polling)	
			1 When	UDRE is set,	request an inte	errupt	
		TXCIE	Interrupt enable	e bit for Trans	mit Complete.		
			0 Interru	pt from TXC is	s inhibited (use	e polling)	
			1 When	TXC is set, re	quest an inter	rupt	
		RXCIE	Interrupt enable	e bit for Recei	ve Complete		
			0 Interru	pt from RXC is	s inhibited (us	e polling)	
			1 When	RXC is set, re	quest an inter	rupt	
	v	VAKEIE	Interrupt enab device is in sto requested to wa	op mode, if R	XD goes to L		
			0 Interru	pt from Wake	is inhibited		
			1 When	WAKE is set,	request an inte	errupt	
		TXE	Enables the tra	nsmitter unit.			
			0 Transr	nitter is disabl	ed		
			1 Transr	nitter is enable	ed		
		RXE	Enables the red	ceiver unit.			
			0 Receiv	er is disabled			
			1 Receiv	er is enabled			
	U	SARTEN	Activate USAR	T module by s	upplying clock	κ.	
			0 USAR	T is disabled (	clock is halted	)	
			1 USAR	T is enabled			
		U2X	This bit only receiver sampli		the asynchr	onous operat	ion and selec
			0 Norma	l asynchronou	is operation		
			1 Double	Speed asynd	hronous oper	ation	

# UCTRL3 (USART Control 3 Register) E4H, FCH

7	6	5	4	3	2	1	0	
MASTER	LOOPS	DISXCK	SPISS	-	USBS	TX8	RX8	
RW	RW	RW	RW	-	RW	RW	RW	
							Initial value : 00	
	Μ	ASTER	Selects maste controls the dire			hronous mod	e operation and	
			0 Slave	mode operatio	n and XCK is	input pin.		
			1 Master	r mode operati	ion and XCK is	s output pin		
	L	OOPS	Controls the Lo	op Back mode	e of USART, fo	or test mode		
			0 Norma	l operation				
			1 Loop E	Back mode				
	D	ISXCK	In Synchronous mode of operation, selects the waveform of XCK output.					
				s free-running mode.	while USAR	T is enabled	in synchronou	

1	XCK is active while any frame is on transferring.	
---	---------------------------------------------------	--

- **SPISS** Controls the functionality of SS pin in master SPI mode.
  - 0 SS pin is normal GPIO or other primary function
  - 1 SS output to other slave device
- **USBS** Selects the length of stop bit in Asynchronous or Synchronous mode of operation.
  - 0 1 Stop Bit
  - 1 2 Stop Bit
- **TX8** The ninth bit of data frame in Asynchronous or Synchronous mode of operation. Write this bit first before loading the UDATA register.
  - 0 MSB (9<sup>th</sup> bit) to be transmitted is '0'
  - 1 MSB (9<sup>th</sup> bit) to be transmitted is '1'
- **RX8** The ninth bit of data frame in Asynchronous or Synchronous mode of operation. Read this bit first before reading the receive buffer.
  - 0 MSB (9<sup>th</sup> bit) received is '0'
  - 1 MSB (9<sup>th</sup> bit) received is '1'

#### USTAT (USART Status Register) E5H

7	6	5	4	3	2	1	0			
UDRE	TXC	RXC	WAKE	SOFTRST	DOR	FE	PE			
RW	RW	RW	RW	RW	R	R	R			
							Initial value : 80			
		UDRE	The UDRE flag indicates if the transmit buffer (UDATA) is ready to b loaded with new data. If UDRE is '1', it means the transmit buffer i empty and can hold one or two new data. This flag can generate a UDRE interrupt. Writing '0' to this bit position will clear UDRE flag.							
		0 Transmit buffer is not empty.								
			1 Transmit buffer is empty.							
		тхс	This flag is set when the entire frame in the transmit shift register has been shifted out and there is no new data currently present in the transmit buffer. This flag is automatically cleared when the interrup service routine of a TXC interrupt is executed. It is also cleared by writing '0' to this bit position. This flag can generate a TXC interrupt.							
			0 Transr	nission is ongo	oing.					
				nit buffer is e fted out comp		data in trans	mit shift registe			
		RXC	This flag is set when there are unread data in the receive buffer ar cleared when all the data in the receive buffer are read. The RXC fla can be used to generate a RXC interrupt.							
			0 There	is no data unre	ead in the rece	eive buffer				
			1 There	are more than	1 data in the	receive buffer				
		WAKE	This flag is set when the RX pin is detected low while the CPU is in stop mode. This flag can be used to generate a WAKE interrupt. This bit is se only when in asynchronous mode of operation.							
				KE interrupt is						
			1 WAKE	interrupt is ge	enerated.					
	S	OFTRST	This is an internal reset and only has effect on USART. Writing '1' to thi bit initializes the internal logic of USART and is auto cleared.							
			0 No ope	eration						
			1 Reset	USART						
		DOR					s bit is set, th he receive buffe			

#### is read.

- 0 No Data OverRun
- 1 Data OverRun detected
- **FE** This bit is set if the first stop bit of next character in the receive buffer is detected as '0'. This bit is valid until the receive buffer is read.
  - 0 No Frame Error
  - 1 Frame Error detected
- **PE** This bit is set if the next character in the receive buffer has a Parity Error when received while Parity Checking is enabled. This bit is valid until the receive buffer is read.
  - 0 No Parity Error
  - 1 Parity Error detected

<sup>NOTE</sup> When the WAKE function of USART is used as a release source from STOP mode, it is required to clear this bit in the RX interrupt service routine. Else the device will not wake-up from STOP mode again by the change of RX pin.

#### UBAUD (USART Baud-Rate Generation Register) E6H, FEH

7	6	5	4	3	2	1	0
UBAUD7	UBAUD6	UBAUD5	UBAUD4	UBAUD3	UBAUD2	UBAUD1	UBAUDO
RW							
							Initial value : FF

**UBAUD [7:0]** The value in this register is used to generate internal baud rate in asynchronous mode or to generate XCK clock in synchronous or spi mode. To prevent malfunction, do not write '0' in asynchronous mode, and do not write '0' or '1' in synchronous or spi mode.

#### UDATA (USART Data Register) E7H, FFH

7	6	5	4	3	2	1	0	
UDATA7	UDATA6	UDATA5	UDATA4	UDATA3	UDATA2	UDATA 1	UDATA0	
RW	RW							
							Initial value : F	ĒF <sub>H</sub>

**UDATA [7:0]** The USART Transmit Buffer and Receive Buffer share the same I/O address with this DATA register. The Transmit Data Buffer is the destination for data written to the UDATA register. Reading the UDATA register returns the contents of the Receive Buffer.

Write this register only when the UDRE flag is set. In spi or synchronous master mode, write this register even if TX is not enabled to generate clock, XCK.

# 11.7.14 Baud Rate setting (example)

	fOSC=1.00MHz			fOSC=1.8432MHz				fOSC=2.00MHz				
Baud	U2X=0 U2X=1			U2X=0 U2X=1			X=1	U2X=0 U2X=1				
Rate	UBAUD	ERROR	UBAUD	ERROR	UBAUD	ERROR	UBAUD	ERROR	UBAUD	ERROR	UBAUD	ERROR
2400	25	0.2%	51	0.2%	47	0.0%	95	0.0%	51	0.2%	103	0.2%
4800	12	0.2%	25	0.2%	23	0.0%	47	0.0%	25	0.2%	51	0.2%
9600	6	-7.0%	12	0.2%	11	0.0%	23	0.0%	12	0.2%	25	0.2%
14.4K	3	8.5%	8	-3.5%	7	0.0%	15	0.0%	8	-3.5%	16	2.1%
19.2K	2	8.5%	6	-7.0%	5	0.0%	11	0.0%	6	-7.0%	12	0.2%
28.8K	1	8.5%	3	8.5%	3	0.0%	7	0.0%	3	8.5%	8	-3.5%
38.4K	1	-18.6%	2	8.5%	2	0.0%	5	0.0%	2	8.5%	6	-7.0%
57.6K	-	-	1	8.5%	1	-25.0%	3	0.0%	1	8.5%	3	8.5%
76.8K	-	-	1	-18.6%	1	0.0%	2	0.0%	1	-18.6%	2	8.5%
115.2 K	-	-	-	-	-	-	1	0.0%	-	-	1	8.5%
230.4 K	-	-	-	-	-	-	-	-	-	-	-	-
		fOSC=3.0	6864MHz		fOSC=4.00MHz					fOSC=7.	3728MHz	
Baud	U2)	K=0	U2	K=1	U2	X=0		X=1	U2	X=0	U2	X=1
Rate	UBAUD	ERROR	UBAUD	ERROR	UBAUD	ERROR	UBAUD	ERROR	UBAUD	ERROR	UBAUD	ERROR
2400	95	0.0%	191	0.0%	103	0.2%	207	0.2%	191	0.0%	-	-
4800	47	0.0%	95	0.0%	51	0.2%	103	0.2%	95	0.0%	191	0.0%
9600	23	0.0%	47	0.0%	25	0.2%	51	0.2%	47	0.0%	95	0.0%
14.4K	15	0.0%	31	0.0%	16	2.1%	34	-0.8%	31	0.0%	63	0.0%
19.2K	11	0.0%	23	0.0%	12	0.2%	25	0.2%	23	0.0%	47	0.0%
28.8K	7	0.0%	15	0.0%	8	-3.5%	16	2.1%	15	0.0%	31	0.0%
38.4K	5	0.0%	11	0.0%	6	-7.0%	12	0.2%	11	0.0%	23	0.0%
57.6K	3	0.0%	7	0.0%	3	8.5%	8	-3.5%	7	0.0%	15	0.0%
76.8K	2	0.0%	5	0.0%	2	8.5%	6	-7.0%	5	0.0%	11	0.0%
115.2K	1	0.0%	3	0.0%	1	8.5%	3	8.5%	3	0.0%	7	0.0%
230.4K	-	-	1	0.0%	-	-	1	8.5%	1	0.0%	3	0.0%
250K	-	-	1	-7.8%	-	-	1	0.0%	1	-7.8%	3	-7.8%
0.5M	-	-	-	-	-	-	-	-	-	-	1	-7.8%
		fOSC=8	3.00MHz		fOSC=11.0592MHz			fOSC=14.7456MHz				
Baud	U2)	K=0	U2	X=1	U2X=0 U2X=1			X=1	U2X=0 U2X=1			X=1
Rate	UBAUD	ERROR	UBAUD	ERROR	UBAUD	ERROR	UBAUD	ERROR	UBAUD	ERROR	UBAUD	ERROR
2400	207	0.2%	-	-	-	-	-	-	-	-	-	-
4800	103	0.2%	207	0.2%	143	0.0%	-	-	191	0.0%	-	-
9600	51	0.2%	103	0.2%	71	0.0%	143	0.0%	95	0.0%	191	0.0%
14.4K	34	-0.8%	68	0.6%	47	0.0%	95	0.0%	63	0.0%	127	0.0%
19.2K	25	0.2%	51	0.2%	35	0.0%	71	0.0%	47	0.0%	95	0.0%
28.8K	16	2.1%	34	-0.8%	23	0.0%	47	0.0%	31	0.0%	63	0.0%
38.4K	12	0.2%	25	0.2%	17	0.0%	35	0.0%	23	0.0%	47	0.0%
57.6K	8	-3.5%	16	2.1%	11	0.0%	23	0.0%	15	0.0%	31	0.0%
76.8K	6	-7.0%	12	0.2%	8	0.0%	17	0.0%	11	0.0%	23	0.0%
115.2K	3	8.5%	8	-3.5%	5	0.0%	11	0.0%	7	0.0%	15	0.0%
230.4K	1	8.5%	3	8.5%	2	0.0%	5	0.0%	3	0.0%	7	0.0%
250K	1	0.0%	3	0.0%	2	-7.8%	5	-7.8%	3	-7.8%	6	5.3%
0.5M	-	-	1	0.0%	-	-	2	-7.8%	1	-7.8%	3	-7.8%
1M	-	-	-	-	-	-	-	-	-	-	1	-7.8%

#### Table 11-17 Examples of UBAUD Settings for Commonly Used Oscillator Frequencies
# 11.8 SPI

## 11.8.1 Overview

There is Serial Peripheral Interface (SPI) one channel in Z51F0811. The SPI allows synchronous serial data transfer between the external serial devices. It can do Full-duplex communication by 4-wire (MOSI, MISO, SCK, SS), support Master/Slave mode, can select serial clock (SCK) polarity, phase and whether LSB first data transfer or MSB first data transfer.

### 11.8.2 Block Diagram



Figure 11-48 SPI Block Diagram

#### 11.8.3 Data Transmit / Receive Operation

User can use SPI for serial data communication by following step

- 1. Select SPI operation mode(master/slave, polarity, phase) by control register SPICR.
- 2. When the SPI is configured as a Master, it selects a Slave by SS signal (active low).

When the SPI is configured as a Slave, it is selected by SS signal incoming from Master

- 3. When the user writes a byte to the data register SPIDR, SPI will start an operation.
- 4. In this time, if the SPI is configured as a Master, serial clock will come out of SCK pin. And Master shifts the eight bits into the Slave (transmit), Slave shifts the eight bits into the Master at the same time (receive). If the SPI is configured as a Slave, serial clock will come into SCK pin. And Slave shifts the eight bits into the Master (transmit), Master shifts the eight bits into the Slave at the same time (receive).
- 5. When transmit/receive is done, TCIR (Transmit Complete or Interrupt Request) bit will be set. If the SPI interrupt is enabled, an interrupt is requested. And TCIR bit is cleared by hardware when executing the corresponding interrupt. If SPI interrupt is disable, TCIR bit is cleared when user read the status register SPISR, and then access (read/write) the data register SPIDR.
- Note) If you want to use both transmit and receive, set the TXENA, RXENA bit of SPISR, and if user want to use only either transmit or receive, clear the TXENA or RXENA. In this case, user can use disabled pin by GPIO freely.

#### 11.8.4 SS pin function

- 1. When the SPI is configured as a Slave, the SS pin is always input. If LOW signal come into SS pin, the SPI logic is active. And if 'HIGH' signal come into SS pin, the SPI logic is stop. In this time, SPI logic will be reset, and invalidated any received data.
- 2. When the SPI is configured as a Master, the user can select the direction of the SS pin by port direction register (PxIO[x]). If the SS pin is configured as an output, user can use general GPIO output mode. If the SS pin is configured as an input, 'HIGH' signal must come into SS pin to guarantee Master operation. If 'LOW' signal come into SS pin, the SPI logic interprets this as another master selecting the SPI as a slave and starting to send data to it. To avoid bus contention, MS bit of SPICR will be cleared and the SPI becomes a Slave and then, TCIR bit of SPISR will be set, and if the SPI interrupt is enabled, an interrupt is requested.

Note)

- When the SS pin is configured as an output at Master mode, SS pin's output value is defined by user's software (PxDA[x]). Before SPICR setting, the direction of SS pin must be defined
- If you don't need to use SS pin, clear the SSENA bit of SPISR. So, you can use disabled pin by GPIO freely. In this case, SS signal is driven by 'HIGH' or 'LOW' internally. In other words, master is 'HIGH', salve is 'LOW'
- When SS pin is configured as input(master or slave), if 'HIGH' signal come into SS pin, this flag bit will be set at the SS rising time. And you can clear it by writing '0'.



Figure 11-49 SPI Transmit/Receive Timing Diagram at CPHA = 0



Figure 11-50 SPI Transmit/Receive Timing Diagram at CPHA = 1

#### 11.8.6 Register Map

	9 1			
Name	Address	Dir	Default	Description
SPICR	D2H	R/W	0H	SPI Control Register
SPIDR	D3H	R/W	0H	SPI Data Register
SPISR	D4H	R/W	0H	SPI Status Register

# Table 11-18 Register Map

### 11.8.7 SPI Register description

The SPI Register consists of SPI Control Register (SPICR), SPI Status Register (SPISR) and SPI Data Register (SPIDR)

### 11.8.8 Register description for SPI

### SPICR (SPI Control Register) : D2H

7	6	5	4		3	2	1	0				
SPIEN	FLSB	MS	CPC	L	CPHA	DSCR	SCR1	SCR0				
RW	RW	RW	RW	1	RW	RW	RW	RW				
								nitial value : 00H				
		SPIEN	This bit co	ontrols the	e SPI opera	ation						
			0 S	PI Disabl	е							
			1 S									
		FLSB	This bit se	elects the	data trans	mission sequ	ence					
			0 MSB First									
			1 L	SB First								
		MS	This bit se	elects who	ether Maste	er or Slave m	ode					
			0 S	lave mod	e							
			1 N	laster mo	de							
		CPOL				erial clock (S	,					
		СРНА			,		's value at idl					
				Clock Phase (CPHA) bit determine if data is sampled on the leading or trailing edge of SCK. Refer to Figure 11-49, Figure 11-50								
			CPOL	CPOL CPHA Leading Edge Trailing								
			0	0 Sample (Rising) Setup (Fa								
			0	1	Set	up (Rising)	Sample (	Falling)				
			1	0	Sar	nple (Falling)	Setup (R	ising)				
			1	1	Set	up (Falling)	Sample (	Rising)				
	s	DSCR SCR[2:0]						configured as a oubled in Master				
			fx– Main s	system cl	ock oscillat	ion frequency	/.					
			DSCR	SCR1	SCR0	SCK fre	quency					
			0	0	0	fx/4						
			0	0	1	fx/16						
			0	1	0	fx/64						
			0	1	1	fx/128						
			1	0	0	fx/2						
			1	0	1	fx/8						
			1	1	0	fx/32						
			1	1	1	fx/64						

#### SPIDR (SPI Data Register) : D3H

7	6	5	4	3	2	1	0	
SPIDR7	SPIDR6	SPIDR5	SPIDR4	SPIDR3	SPIDR2	SPIDR1	SPIDR0	
RW								
						I	nitial value : 00	ЭН

SPIDR [7:0]

SPI data register.

Although you only use reception, user must write any data in here to start the SPI operation.

#### SPISR (SPI Status Register) : D4H

7	6	5	4	3	2	1	0
TCIR	WCOL	SS_HIGH	-	TWOPIN	SSENA	TXENA	RXENA
R	R	RW	-	RW	RW	RW	RW

Initial value : 00H

TCIR	SPI ir cleare interru	a serial data transmission is complete, the TCIR bit is set. If the nterrupt is enabled, an interrupt is requested. And TCIR bit is d by hardware when executing the corresponding interrupt. If SPI upt is disable, TCIR bit is cleared when user read the status register R, and then access (read/write) the data register SPIDR.
	0	Interrupt cleared
	1	Transmission Complete and Interrupt Requested
WCOL	This b	it is set if the data register SPIDR is written during a data transfer. it is cleared when user read the status register SPISR, and then s (read/write) the data register SPIDR.
	0	No collision
	1	Write Collision
SS_HIGH	come can cl	SS pin is configured as input(master or slave), if 'HIGH' signal into SS pin, this flag bit will be set at the SS rising time. And you ear it by writing '0'. an write only zero.
	0	Flag is cleared
	1	Flag is set
TWOPIN	•	it controls the 2 pin operation.
		ster mode.
	0	Disable
	1	Enable
SSENA	This b	it controls the SS pin operation
	0	Disable
	1	Enable
TXENA	This b	it controls a data transfer operation
	0	Disable
	1	Enable
RXENA	This b	it controls a data reception operation
	0	Disable
	1	Enable

Note that if the MS is set to '0', when TWOPIN is set to '0', port 03 is set to MISO and if the MS is set to '0', when TWOPIN is set to '1', port 02 is set to MOSI. But if the MS is set to '1', when TWOPIN is set to '0', port 03 is set to MOSI and if the MS is set to '1', when TWOPIN is set to '1', port 02 is set to MISO.

# 11.9 I<sup>2</sup>C

## 11.9.1 Overview

The I<sup>2</sup>C is one of industrial standard serial communication protocols, and which uses 2 bus lines Serial Data Line (SDA) and Serial Clock Line (SCL) to exchange data. Because both SDA and SCL lines are open-drain output, each line needs pull-up resistor. The features are as shown below.

- Compatible with I<sup>2</sup>C bus standard
- Multi-master operation
- Up to 400 KHz data transfer speed
- 7 bit address
- Support two slave addresses
- Both master and slave operation
- Bus busy detection

### 11.9.2 Block Diagram



Figure 11-51 I<sup>2</sup>C Block Diagram

# 11.9.3 I<sup>2</sup>C Bit Transfer

The data on the SDA line must be stable during HIGH period of the clock, SCL. The HIGH or LOW state of the data line can only change when the clock signal on the SCL line is LOW. The exceptions are START(S), repeated START(Sr) and STOP(P) condition where data line changes when clock line is high.



Figure 11-52 Bit Transfer on the I<sup>2</sup>C-Bus

## 11.9.4 Start / Repeated Start / Stop

One master can issue a START (S) condition to notice other devices connected to the SCL, SDA lines that it will use the bus. A STOP (P) condition is generated by the master to release the bus lines so that other devices can use it.

A high to low transition on the SDA line while SCL is high defines a START (S) condition.

A low to high transition on the SDA line while SCL is high defines a STOP (P) condition.

START and STOP conditions are always generated by the master. The bus is considered to be busy after START condition. The bus is considered to be free again after STOP condition, ie, the bus is busy between START and STOP condition. If a repeated START condition (Sr) is generated instead of STOP condition, the bus stays busy. So, the START and repeated START conditions are functionally identical.



Figure 11-53 START and STOP Condition

## 11.9.5 Data Transfer

Every byte put on the SDA line must be 8-bits long. The number of bytes that can be transmitted per transfer is unlimited. Each byte has to be followed by an acknowledge bit. Data is transferred with

the most significant bit (MSB) first. If a slave can't receive or transmit another complete byte of data until it has performed some other function, it can hold the clock line SCL LOW to force the master into a wait state. Data transfer then continues when the slave is ready for another byte of data and releases clock line SCL.



Figure 11-54 STOP or Repeated START Condition

### 11.9.6 Acknowledge

The acknowledge related clock pulse is generated by the master. The transmitter releases the SDA line (HIGH) during the acknowledge clock pulse. The receiver must pull down the SDA line during the acknowledge clock pulse so that it remains stable LOW during the HIGH period of this clock pulse. When a slave is addressed by a master (Address Packet), and if it is unable to receive or transmit because it's performing some real time function, the data line must be left HIGH by the slave. And also, when a slave addressed by a master is unable to receive more data bits, the slave receiver must release the SDA line (Data Packet). The master can then generate either a STOP condition to abort the transfer, or a repeated START condition to start a new transfer.

If a master receiver is involved in a transfer, it must signal the end of data to the slave transmitter by not generating an acknowledge on the last byte that was clocked out of the slave. The slave transmitter must release the data line to allow the master to generate a STOP or repeated START condition.



Figure 11-55 Acknowledge on the I<sup>2</sup>C-Bus

#### 11.9.7 Synchronization / Arbitration

Clock synchronization is performed using the wired-AND connection of I<sup>2</sup>C interfaces to the SCL line. This means that a HIGH to LOW transition on the SCL line will cause the devices concerned to start counting off their LOW period and it will hold the SCL line in that state until the clock HIGH state is reached. However the LOW to HIGH transition of this clock may not change the state of the SCL line if another clock is still within its LOW period. In this way, a synchronized SCL clock is generated with its LOW period determined by the device with the longest clock LOW period, and its HIGH period determined by the one with the shortest clock HIGH period.

A master may start a transfer only if the bus is free. Two or more masters may generate a START condition. Arbitration takes place on the SDA line, while the SCL line is at the HIGH level, in such a way that the master which transmits a HIGH level, while another master is transmitting a LOW level will switch off its DATA output state because the level on the bus doesn't correspond to its own level. Arbitration continues for many bits until a winning master gets the ownership of I<sup>2</sup>C bus. Its first stage is comparison of the address bits.



Figure 11-56 Clock Synchronization during Arbitration Procedure



Figure 11-57 Arbitration Procedure of Two Masters

#### 11.9.8 Operation

The  $I^2C$  is byte-oriented and interrupt based. Interrupts are issued after all bus events except for a transmission of a START condition. Because the  $I^2C$  is interrupt based, the application software is free to carry on other operations during a  $I^2C$  byte transfer.

Note that when a I<sup>2</sup>C interrupt is generated, IIF flag in I2CMR register is set, it is cleared by writing an arbitrary value to I2CSR. When I<sup>2</sup>C interrupt occurs, the SCL line is hold LOW until writing any value to I2CSR. When the IIF flag is set, the I2CSR contains a value indicating the current state of the I<sup>2</sup>C bus. According to the value in I2CSR, software can decide what to do next.

l<sup>2</sup>C can operate in 4 modes by configuring master/slave, transmitter/receiver. The operating mode is configured by a winning master. A more detailed explanation follows below.

### 11.9.8.1 Master Transmitter

To operate I<sup>2</sup>C in master transmitter, follow the recommended steps below.

- 1. Enable I<sup>2</sup>C by setting IICEN bit in I2CMR. This provides main clock to the peripheral.
- Load SLA+W into the I2CDR where SLA is address of slave device and W is transfer direction from the viewpoint of the master. For master transmitter, W is '0'. Note that I2CDR is used for both address and data.
- 3. Configure baud rate by writing desired value to both I2CSCLLR and I2CSCLHR for the Low and High period of SCL line.
- Configure the I2CSDAHR to decide when SDA changes value from falling edge of SCL. If SDA should change in the middle of SCL LOW period, load half the value of I2CSCLLR to the I2CSDAHR.
- 5. Set the START bit in I2CMR. This transmits a START condition. And also configure how to handle interrupt and ACK signal. When the START bit is set, 8-bit data in I2CDR is transmitted out according to the baud-rate.
- 6. This is ACK signal processing stage for address packet transmitted by master. When 7-bit address and 1-bit transfer direction is transmitted to target slave device, the master can know whether the slave acknowledged or not in the 9<sup>th</sup> high period of SCL. If the master gains bus mastership, I<sup>2</sup>C generates GCALL interrupt regardless of the reception of ACK from the slave device. When I<sup>2</sup>C loses bus mastership during arbitration process, the MLOST bit in I2CSR is set, and I<sup>2</sup>C waits in idle state or can be operate as an addressed slave. To operate as a slave when the MLSOT bit in I2CSR is set, the ACKEN bit in I2CMR must be set and the received 7-bit address must equal to the SLA bits in I2CSAR. In this case I<sup>2</sup>C operates as a slave transmitter or a slave receiver (go to appropriate section). In this stage, I<sup>2</sup>C holds the SCL LOW. This is because to decide whether I<sup>2</sup>C continues serial transfer or stops communication. The following steps continue assuming that I<sup>2</sup>C does not lose mastership during first data transfer.

I<sup>2</sup>C (Master) can choose one of the following cases regardless of the reception of ACK signal from slave.

 Master receives ACK signal from slave, so continues data transfer because slave can receive more data from master. In this case, load data to transmit to I2CDR.
 Master stops data transfer even if it receives ACK signal from slave. In this case, set the STOP bit in I2CMR.

3) Master transmits repeated START condition with not checking ACK signal. In this case, load SLA+R/W into the I2CDR and set START bit in I2CMR.

After doing one of the actions above, write arbitrary value to I2CSR to release SCL line. In case of 1), move to step 7. In case of 2), move to step 9 to handle STOP interrupt. In case of 3), move to step 6 after transmitting the data in I2CDR and if transfer direction bit is '1' go to master receiver section.

- 7. 1-Byte of data is being transmitted. During data transfer, bus arbitration continues.
- This is ACK signal processing stage for data packet transmitted by master. I<sup>2</sup>C holds the SCL LOW. When I<sup>2</sup>C loses bus mastership while transmitting data arbitrating other masters, the MLOST bit in I2CSR is set. If then, I<sup>2</sup>C waits in idle state. When the data in I2CDR is transmitted completely, I<sup>2</sup>C generates TEND interrupt.

I<sup>2</sup>C can choose one of the following cases regardless of the reception of ACK signal from slave.

 Master receives ACK signal from slave, so continues data transfer because slave can receive more data from master. In this case, load data to transmit to I2CDR.
 Master stops data transfer even if it receives ACK signal from slave. In this case, set the STOP bit in I2CMR.

3) Master transmits repeated START condition with not checking ACK signal. In this case, load SLA+R/W into the I2CDR and set the START bit in I2CMR.

After doing one of the actions above, write arbitrary value to I2CSR to release SCL line. In case of 1), move to step 7. In case of 2), move to step 9 to handle STOP interrupt. In case of 3), move to step 6 after transmitting the data in I2CDR, and if transfer direction bit is '1' go to master receiver section.

 This is the final step for master transmitter function of I<sup>2</sup>C, handling STOP interrupt. The STOP bit indicates that data transfer between master and slave is over. To clear I2CSR, write arbitrary value to I2CSR. After this, I<sup>2</sup>C enters idle state.



The next figure depicts above process for master transmitter operation of I<sup>2</sup>C.

Figure 11-58 Formats and States in the Master Transmitter Mode

#### 11.9.8.2 Master Receiver

To operate I<sup>2</sup>C in master receiver, follow the recommended steps below.

- 1. Enable I<sup>2</sup>C by setting IICEN bit in I2CMR. This provides main clock to the peripheral.
- Load SLA+R into the I2CDR where SLA is address of slave device and R is transfer direction from the viewpoint of the master. For master receiver, R is '1'. Note that I2CDR is used for both address and data.
- 3. Configure baud rate by writing desired value to both I2CSCLLR and I2CSCLHR for the Low and High period of SCL line.
- Configure the I2CSDAHR to decide when SDA changes value from falling edge of SCL. If SDA should change in the middle of SCL LOW period, load half the value of I2CSCLLR to the I2CSDAHR.
- 5. Set the START bit in I2CMR. This transmits a START condition. And also configure how to handle interrupt and ACK signal. When the START bit is set, 8-bit data in I2CDR is transmitted out according to the baud-rate.
- 6. This is ACK signal processing stage for address packet transmitted by master. When 7-bit address and 1-bit transfer direction is transmitted to target slave device, the master can know whether the slave acknowledged or not in the 9<sup>th</sup> high period of SCL. If the master gains bus mastership, I<sup>2</sup>C generates GCALL interrupt regardless of the reception of ACK from the slave device. When I<sup>2</sup>C loses bus mastership during arbitration process, the MLOST bit in I2CSR is set, and I<sup>2</sup>C waits in idle state or can be operate as an addressed slave. To operate as a slave when the MLSOT bit in I2CSR is set, the ACKEN bit in I2CMR must be set and the received 7-bit address must equal to the SLA bits in I2CSAR. In this case I<sup>2</sup>C operates as a slave transmitter or a slave receiver (go to appropriate section). In this stage, I<sup>2</sup>C holds the SCL LOW. This is because to decide whether I<sup>2</sup>C continues serial transfer or stops communication. The following steps continue assuming that I<sup>2</sup>C does not lose mastership during first data transfer.

 $\rm I^2C$  (Master) can choose one of the following cases according to the reception of ACK signal from slave.

1) Master receives ACK signal from slave, so continues data transfer because slave can prepare and transmit more data to master. Configure ACKEN bit in I2CMR to decide whether  $I^2C$  ACKnowledges the next data to be received or not.

2) Master stops data transfer because it receives no ACK signal from slave. In this case, set the STOP bit in I2CMR.

3) Master transmits repeated START condition due to no ACK signal from slave. In this case, load SLA+R/W into the I2CDR and set START bit in I2CMR.

After doing one of the actions above, write arbitrary value to I2CSR to release SCL line. In case of 1), move to step 7. In case of 2), move to step 9 to handle STOP interrupt. In case of 3), move to step 6 after transmitting the data in I2CDR and if transfer direction bit is '0' go to master transmitter section.

- 7. 1-Byte of data is being received.
- 8. This is ACK signal processing stage for data packet transmitted by slave. I<sup>2</sup>C holds the SCL LOW. When 1-Byte of data is received completely, I<sup>2</sup>C generates TEND interrupt.

I<sup>2</sup>C can choose one of the following cases according to the RXACK flag in I2CSR.

1) Master continues receiving data from slave. To do this, set ACKEN bit in I2CMR to ACKnowledge the next data to be received.

2) Master wants to terminate data transfer when it receives next data by not generating ACK signal. This can be done by clearing ACKEN bit in I2CMR.

3) Because no ACK signal is detected, master terminates data transfer. In this case, set the STOP bit in I2CMR.

4) No ACK signal is detected, and master transmits repeated START condition. In this case,

load SLA+R/W into the I2CDR and set the START bit in I2CMR.

After doing one of the actions above, write arbitrary value to I2CSR to release SCL line. In case of 1) and 2), move to step 7. In case of 3), move to step 9 to handle STOP interrupt. In case of 4), move to step 6 after transmitting the data in I2CDR, and if transfer direction bit is '0' go to master transmitter section.

 This is the final step for master receiver function of I<sup>2</sup>C, handling STOP interrupt. The STOP bit indicates that data transfer between master and slave is over. To clear I2CSR, write arbitrary value to I2CSR. After this, I<sup>2</sup>C enters idle state.

The processes described above for master receiver operation of I<sup>2</sup>C can be depicted as the following figure.



Figure 11-59 Formats and States in the Master Receiver Mode

#### 11.9.8.3 Slave Transmitter

To operate I<sup>2</sup>C in slave transmitter, follow the recommended steps below.

- If the main operating clock (SCLK) of the system is slower than that of SCL, load value 0x00 into I2CSDAHR to make SDA change within one system clock period from the falling edge of SCL. Note that the hold time of SDA is calculated by SDAH x period of SCLK where SDAH is multiple of number of SCLK coming from I2CSDAHR. When the hold time of SDA is longer than the period of SCLK, I<sup>2</sup>C (slave) cannot transmit serial data properly.
- 2. Enable I<sup>2</sup>C by setting IICEN bit and INTEN bit in I2CMR. This provides main clock to the peripheral.
- When a START condition is detected, I<sup>2</sup>C receives one byte of data and compares it with SLA bits in I2CSAR. If the GCALLEN bit in I2CSAR is enabled, I<sup>2</sup>C compares the received data with value 0x00, the general call address.
- 4. If the received address does not equal to SLA bits in I2CSAR, I<sup>2</sup>C enters idle state ie, waits for another START condition. Else if the address equals to SLA bits and the ACKEN bit is enabled, I<sup>2</sup>C generates SSEL interrupt and the SCL line is held LOW. Note that even if the address equals to SLA bits, when the ACKEN bit is disabled, I<sup>2</sup>C enters idle state. When SSEL interrupt occurs, load transmit data to I2CDR and write arbitrary value to I2CSR to release SCL line.
- 5. 1-Byte of data is being transmitted.
- 6. In this step, I<sup>2</sup>C generates TEND interrupt and holds the SCL line LOW regardless of the reception of ACK signal from master. Slave can select one of the following cases.

No ACK signal is detected and I<sup>2</sup>C waits STOP or repeated START condition.
 ACK signal from master is detected. Load data to transmit into I2CDR.

After doing one of the actions above, write arbitrary value to I2CSR to release SCL line. In case of 1) move to step 7 to terminate communication. In case of 2) move to step 5. In either case, a repeated START condition can be detected. For that case, move step 4.

 This is the final step for slave transmitter function of I<sup>2</sup>C, handling STOP interrupt. The STOP bit indicates that data transfer between master and slave is over. To clear I2CSR, write arbitrary value to I2CSR. After this, I<sup>2</sup>C enters idle state.



The next figure shows flow chart for handling slave transmitter function of I<sup>2</sup>C.

Figure 11-60 Formats and States in the Slave Transmitter Mode

#### 11.9.8.4 Slave Receiver

To operate I<sup>2</sup>C in slave receiver, follow the recommended steps below.

- If the main operating clock (SCLK) of the system is slower than that of SCL, load value 0x00 into I2CSDAHR to make SDA change within one system clock period from the falling edge of SCL. Note that the hold time of SDA is calculated by SDAH x period of SCLK where SDAH is multiple of number of SCLK coming from I2CSDAHR. When the hold time of SDA is longer than the period of SCLK, I<sup>2</sup>C (slave) cannot transmit serial data properly.
- 2. Enable I<sup>2</sup>C by setting IICEN bit and INTEN bit in I2CMR. This provides main clock to the peripheral.
- When a START condition is detected, I<sup>2</sup>C receives one byte of data and compares it with SLA bits in I2CSAR. If the GCALLEN bit in I2CSAR is enabled, I<sup>2</sup>C compares the received data with value 0x00, the general call address.
- 4. If the received address does not equal to SLA bits in I2CSAR, I<sup>2</sup>C enters idle state ie, waits for another START condition. Else if the address equals to SLA bits and the ACKEN bit is enabled, I<sup>2</sup>C generates SSEL interrupt and the SCL line is held LOW. Note that even if the address equals to SLA bits, when the ACKEN bit is disabled, I<sup>2</sup>C enters idle state. When SSEL interrupt occurs and I<sup>2</sup>C is ready to receive data, write arbitrary value to I2CSR to release SCL line.
- 5. 1-Byte of data is being received.
- 6. In this step, I<sup>2</sup>C generates TEND interrupt and holds the SCL line LOW regardless of the reception of ACK signal from master. Slave can select one of the following cases.

1) No ACK signal is detected (ACKEN=0) and  $I^2C$  waits STOP or repeated START condition. 2) ACK signal is detected (ACKEN=1) and  $I^2C$  can continue to receive data from master.

After doing one of the actions above, write arbitrary value to I2CSR to release SCL line. In case of 1) move to step 7 to terminate communication. In case of 2) move to step 5. In either case, a repeated START condition can be detected. For that case, move step 4.

 This is the final step for slave receiver function of I<sup>2</sup>C, handling STOP interrupt. The STOP bit indicates that data transfer between master and slave is over. To clear I2CSR, write arbitrary value to I2CSR. After this, I<sup>2</sup>C enters idle state.

The process can be depicted as following figure when I<sup>2</sup>C operates in slave receiver mode.





Figure 11-61 Formats and States in the Slave Receiver Mode

# 11.9.9 Register Map

Name	Address	Dir	Default	Description
I2CMR	DAH	R/W	00H	I <sup>2</sup> C Mode Control Register
I2CSR	DBH	R	00H	I <sup>2</sup> C Status Register
I2CSCLLR	DCH	R/W	3FH	SCL Low Period Register
I2CSCLHR	DDH	R/W	3FH	SCL High Period Register
I2CSDAHR	DEH	R/W	01H	SDA Hold Time Register
I2CDR	DFH	R/W	FFH	I <sup>2</sup> C Data Register
I2CSAR	D7H	R/W	00H	I <sup>2</sup> C Slave Address Register
I2CSAR1	D6H	R/W	00H	I <sup>2</sup> C Slave Address Register 1

# 11.9.10 I<sup>2</sup>C Register description

I<sup>2</sup>C Registers are composed of I<sup>2</sup>C Mode Control Register (I2CMR), I<sup>2</sup>C Status Register (I2CSR), SCL Low Period Register (I2CSCLLR), SCL High Period Register (I2CSCLHR), SDA Hold Time Register (I2CSDAHR), I<sup>2</sup>C Data Register (I2CDR), and I<sup>2</sup>C Slave Address Register (I2CSAR).

### 11.9.11 Register description for I<sup>2</sup>C

# I2CMR (I<sup>2</sup>C Mode Control Register) : DAH

7	6	5	4	3	2	1	0				
IIF	ICEN	RESET	INTEN	ACKEN	MASTER	STOP	START				
RW	RW	RW	RW	RW	R	RW	RW				
							Initial value : 00				
		IIF	This is interrup	flag bit.							
			0 No inte	errupt is gener	ated or interru	pt is cleared					
			1 An inte	rrupt is gener	ated						
		IICEN	Enable I <sup>2</sup> C Fu	nction Block (b	by providing cl	ock)					
			0 l <sup>2</sup> C is i	nactive							
			1 l <sup>2</sup> C is a	active							
		RESET	Initialize internal registers of I <sup>2</sup> C.								
			0 No ope	eration							
			1 Initializ	e l <sup>2</sup> C, auto cle	eared						
		INTEN	Enable interrup	t generation o	of I <sup>2</sup> C.						
			0 Disable	e interrupt, op	erates in pollir	ng mode					
			1 Enable	interrupt							
		ACKEN	Controls ACK signal generation at ninth SCL period.								
			Note) ACK signal is output (SDA=0) for the following 3 cases.								
			When received address packet equals to SLA bits in I2CSAR								
			When received address packet equals to value 0x00 with GCALL enabled								
			When I <sup>2</sup> C operation	ates as a rece	iver (master o	r slave)					
			0 No AC	K signal is gei	nerated (SDA:	=1)					
			1 ACK s	gnal is genera	ated (SDA=0)						
	I	MASTER	Represent ope	rating mode of	f I <sup>2</sup> C						
			0 l <sup>2</sup> C is i	n slave mode							
			1 I <sup>2</sup> C is i	n master mod	e						
		STOP	When I <sup>2</sup> C is ma	ster, generate	es STOP conc	lition.					
			0 No ope	eration							
			1 STOP	condition is to	be generated						
		OTADT	When I <sup>2</sup> C is ma			dition					
		START		ister, generate	S START COL						
		START	0 No ope		START COL						

# I2CSR (I<sup>2</sup>C Status Register) : DBH

7	6	5	4	3	2	1	0			
GCALL	TEND	STOP	SSEL	MLOST	BUSY	TMODE	RXACK			
R	R	R	R	R	R	R	R			
						I	nitial value : 00			
		GCALL	This bit has di slave. Note 1)							
			When $I^2C$ is a (Address ACK)	from slave.						
			When I <sup>2</sup> C is a s				all.			
					d (Master mod	e)				
				is received (N	,					
					not general c	-	slave mode)			
					s is detected (S	,				
		TEND	This bit is set w	-		-	ely. Note 1)			
			<ul> <li>0 1 byte of data is not completely transferred</li> <li>1 byte of data is completely transferred</li> </ul>							
		STOP	This bit is set w			cted. Note 1)				
				OP condition i						
				condition is de						
		SSEL	This bit is set when $I^2C$ is addressed by other master. Note 1)							
				not selected as						
					other master a					
	I	MLOST	This bit represe			ion in master	mode. Note 1)			
				intains bus m						
					stership during	arbitration pro	ocess			
		BUSY	This bit reflects							
					y master can	issue a STAR	T condition			
				s is busy	0					
	1	TMODE	This bit is used		nether I <sup>2</sup> C is tr	ansmitter or re	eceiver.			
			0 l <sup>2</sup> C is a receiver							
			1 I <sup>2</sup> C is a transmitter							
	I	RXACK	This bit shows	the state of A	CK signal.					
			0 No AC	K is received						
			1 ACK is	agenerated at	ninth SCL per	riod				

Note 1) These bits can be source of interrupt.

When an I<sup>2</sup>C interrupt occurs except for STOP interrupt, the SCL line is hold LOW. To release SCL, write arbitrary value to I2CSR. When I2CSR is written, the TEND, STOP, SSEL, LOST, RXACK bits are cleared.

#### I2CSCLLR (SCL Low Period Register) : DCH

7	6	5	4	3	2	1	0	
SCLL7	SCLL6	SCLL5	SCLL4	SCLL3	SCLL2	SOLL1	SCLLO	
RW								

Initial value : 3FH

#### I2CSCLHR (SCL High Period Register) : DDH

7	6	5	4	3	2	1	0
SCLH7	SCLH6	SCLH5	SCLH4	SCLHB	SCLH2	SCLH1	SCLH0
RW							

Initial value : 3FH

So, the operating frequency of I<sup>2</sup>C in master mode (fI2C) is calculated by the following equation.

$$fI2C = \frac{1}{tSCLK \times (4 (SCLL + SCLH) + 4)}$$

#### I2CSDAHR (SDA Hold Time Register) : DEH

7	6	5	4	3	2	1	0
SDAH7	SDAH6	SDAH5	SDAH4	SDAH3	SDAH2	SDAH1	SDAH0
RW							
						I	nitial value : 01h

## I2CDR (I<sup>2</sup>C Data Register) : DFH

7	6	5	4	3	2	1	0
ICD7	ICD6	ICD5	ICD4	ICD3	ICD2	ICD1	ICD0
RW							

Initial value : FFH

**ICD[7:0]** When I<sup>2</sup>C is configured as a transmitter, load this register with data to be transmitted. When I<sup>2</sup>C is a receiver, the received data is stored into this register.

# I2CSAR (I<sup>2</sup>C Slave Address Register) : D7H

7	6	5	4	3	2	1	0
SLA7	SLA6	SLA5	SLA4	SLA3	SLA2	SLA1	GCALLEN
RW	RW	RW	RW	RW	RW	RW	RW
							nitial value : 00H
	S		These bits configure the slave address of this I <sup>2</sup> C module when I <sup>2</sup> operates in slave mode.				ule when I <sup>2</sup> C
	G		This bit decio when I <sup>2</sup> C oper			neral call add	dress or not
			0 Ignor	e general call	address		
			1 Allow	general call a	ddress		

# I2CSAR1 (I<sup>2</sup>C Slave Address Register 1) : D6H

7	6	5	4	3	2	1	0
SLA7	SLA6	SLA5	SLA4	SLA3	SLA2	SLA1	GCALLEN
RW	RW	RW	RW	RW	RW	RW	RW
						I	nitial value : 00
	S		These bits configure the slave address of this $I^2C \mod O$ operates in slave mode.				ule when I <sup>2</sup> C
	G	CALLEN	This bit decides whether I <sup>2</sup> C allows general call address or not when I <sup>2</sup> C operates in slave mode.				dress or not
			0 Ignor	e general call	address		
			1 Allow	general call a	address		

## 11.10 12-Bit A/D Converter

#### 11.10.1 Overview

The analog-to-digital converter (A/D) allows conversion of an analog input signal to a corresponding 12-bit digital value. The A/D module has tenth analog inputs. The output of the multiplex is the input into the converter, which generates the result via successive approximation. The A/D module has four registers which are the control register ADCM (A/D Converter Mode Register), ADCM2 (A/D Converter Mode Register 2) and A/D result register ADCHR (A/D Converter Result High Register) and ADCLR (A/D Converter Result Low Register). It is selected for the corresponding channel to be converted by setting ADSEL[3:0]. To executing A/D conversion, ADST bit sets to '1'. The register ADCHR and ADCLR contains the results of the A/D conversion. When the conversion is completed, the result is loaded into the ADCHR and ADCLR, the A/D conversion, AFLAG bit is read as '0'. If using STBY (power down) bit, the ADC is disabled. Also internal timer, external generating event, comparator, the trigger of timer1pwm and etc. can start ADC regardless of interrupt occurrence.

ADC Conversion Time = ADCLK \* 60 cycles

After STBY bit is reset (ADC power enable) and it is restarted, during some cycle, ADC conversion value may have an inaccurate value.



### 11.10.2 Block Diagram



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## 11.10.3 ADC Operation



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Figure 11-66 Converter Operation Flow

#### 11.10.4 Register Map

Name	Address	Dir	Default	Description
ADCM	9AH	R/W	8FH	A/D Converter Mode Register
ADCRH	9BH	R	-	A/D Converter Result High Register
ADCRL	9CH	R	-	A/D Converter Result Low Register
ADCM2	9BH	R/W	8FH	A/D Converter Mode 2 Register

#### 11.10.5 ADC Register description

The ADC Register consists of A/D Converter Mode Register (ADCM), A/D Converter Result High Register (ADCRH), A/D Converter Result Low Register (ADCRL), A/D Converter Mode 2 Register (ADCM2).

Note) when STBY bit is set to '1', ADCM2 can be read. If ADC enables, it is possible only to write ADCM2.When reading, ADCRH is read.

# 11.10.6 Register description for ADC

### ADCM (A/D Converter Mode Register) : 9AH

7	6	5	4		3	2	1	0
STBY	ADST	REFSEL	AFLAG	g ad	SEL3	ADSEL2	ADSEL1	ADSEL0
RW	RW	RW	R	1	RW	RW	RW	RW
							l	nitial value : 8FH
		STBY	Control op	eration of A	A/D standb	oy (power de	own)	
			0 A	DC module	enable			
			1 A	DC module	disable (p	ower down	)	
		ADST	Control A/	D Conversi	on stop/st	art.		
			0 A	DC Conver	sion Stop			
			1 A	DC Conver	sion Start			
	R	REFSEL	A/D Conve	erter refere	nce select	ion		
			0 In	ternal Refe	erence (VD	DD)		
			1 E	xternal Ref	erence(AV	/REF, AN0	disable)	
	I	AFLAG	A/D Conve	erter opera	tion state			
			0 D	uring A/D C	Conversior	ו		
				/D Convers		ed		
	AD	SEL[3:0]		erter input s				
			ADSEL3	ADSEL2	ADSEL1	ADSEL0	Description	
			0	0	0	0	Channel0(AN	,
			0	0	0	1	Channel1(AN	-
			0	0	1	0	Channel2(AN	,
			0	0	1	1	Channel3(AN	
			0	1	0	0	Channel4(AN	
			0	1	0	1	Channel5(AN	,
			0	1	1	0	Channel6(AN	
			0	1	1	1	Channel7(AN	,
			1	0	0	0	Channel8(AN	
			1	0	0	1	Channel9(AN	,
			1	0	1	0	Channel10(A	-
			1	0	1	1	Channel11(A	,
			1	1	0	0	Channel12(A	-
			1	1	0	1	Channel13(A	,
			1	1	1	0	Channel14(A	
			1	1	1	1	Channel15(V	18)

#### ADCRH (A/D Converter Result High Register) : 9BH

	7	6	5	4	3	2	1	0
AD	DM11	ADDM10	ADDM9	ADDM8	ADDM7	ADDM6	ADDM5	ADDM4
					ADDL11	ADDL10	ADDL9	ADDL8
	R	R	R	R	R	R	R	R

Initial value : xxH

ADDM[11:4]MSB align, A/D Converter High result (8-bit)ADDL[11:8]LSB align, A/D Converter High result (4-bit)

### ADCRL (A/D Converter Result Low Register) : 9CH

7	6	5	4	3	2	1	0	
ADDM3	ADDM2	ADDM1	ADDM0					
ADDL7	ADDL6	ADDL5	ADDL4	ADDL3	ADDL2	ADDL1	ADDL0	
R	R	R	R	R	R	R	R	
							nitial value : x	αH

ADDM[3:0]MSB align, A/D Converter Low result (4-bit)ADDL[7:0]LSB align, A/D Converter Low result (8-bit)

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# ADCM2 (A/D Converter Mode Register) : 9BH

7	6	5		4	3	2	1	0
EXTRG	TSEL2	TSEL1	TS	EL0	-	ALIGN	CKSEL1	CKSEL0
RW	RW	RW	R	Ŵ	RW	RW	RW	RW
								Initial value : 01
	I	EXTRG	A/D exte	ernal Tr	rigger			
			0	Extern	al Trigger di	sable		
			1	Extern	al Trigger er	nable		
	Т	SEL[2:0]	A/D Trig	ger So	urce selection	on		
			TSEL2	TSEL	_1 TSELC	Description		
			0	0	0	Ext. Interru	pt 0	
			0	0	1	Analog Cor	nparator Low	to High Transiti
			0	1	0	Analog Cor	nparator High	to Low Transition
			0	1	1	Timer1PW	A overflow eve	ent
			1	0	0	Timer1PW	A-ch event	compare match
			1	0	1	Timer1PW	A B-ch event	compare match
			1	1	0	Timer1PW	VIC-ch event	compare match
			1	1	1	Timer3(PW	M) interrupt	
		ALIGN	A/D Cor	nverter	data align se	election.		
			0	MSB a	lign (ADCRI	H[7:0], ADCRL[	7:4])	
			1	LSB al	lign (ADCRF	I[3:0], ADCRL[7	:0])	
	CI	KSEL[1:0]	A/D Co	nverter	r Clock seled	tion		
			CKSEL	_1 C	KSEL0 A	DC Clock	ADC V	DD
			0	0	fz	(/2	Test Or	nly
			0	1	fz	<i>ג</i> /4	3V~5V	
			1	0	fz	k/8	2.7V~3	V
			1	1	fz	(/32	2.4V~2	.7V
					Note) 1.	fx : system clo	ock	

Note) 1. fx : system clock

2. ADC clock have to be used 3MHz under

# **11.11 Analog Comparator**

### 11.11.1 Overview

The Analog Comparator compares the input values on the positive pin AC+ and the negative pin AC-. When the voltage on the positive pin AC+ is higher than the voltage on the negative pin AC-, the Analog Comparator output, ACOUT, is set.

## 11.11.2 Block Diagram



Figure 11-67 Analog Comparator Block Diagram

## 11.11.3 IN/OUT signal description

ACE : This enables Analog Comparator. When ACE is '0', the output of Comparator goes LOW.

BGR : Band Gap Reference Voltage

ACBG : This selects (-) input source between BGR and AC-. When ACBG is '1', the (-) input to AC is BGR.

AC- : This can be (-) input to the AC, and comes directly from external analog pad.

AC+ : This can be (+) input to the AC, and comes directly from external analog pad.

AMUXENB : This selects (+) input source between multiplexed output of ADC and AN5. AMUXENB is the inverted signal of AMUXEN bit in ADCM2 register. When AMUXENB is '0', the (+) input to AC comes from ADC module which is selected by ADSEL[3:0], the channel selection bits in ADCM register.

ACOUT : This is the output of Comparator.

ACO\_OUTEN : Analog Comparator output port Enable.

## 11.11.4 Register Map

Table 11-19 Register Map

Name	Address	Dir	Default	Description
ACCSR	E9	R/W	00H	Analog Comparator Control & Status Register

## 11.11.5 Analog Comparator Register description

Analog Comparator Register has one control register, Analog Comparator Control & Status Register (ACCSR). Note that AMUXENB is the inverted signal of AMUXEN bit which comes from ADC's ADCM2 register

# 11.11.6 Register description for Analog Comparator

7	6	5	4	3	2	1	0
ACE	ACBG	ACO	ACIF	ACIE	ACO_OUTEN	ACISM1	ACISM0
RW	RW	R	R	RW	RW	RW	RW
							Initial value : 00
		ACE	Enable Anal	og Comparat	or (AC).		
			0 Disa	able AC (pow	er down)		
			1 Ena	able AC			
		ACBG	Select (-) inp	out source of	AC, Band Gap Re	eference Volta	age or AN4.
			0 (-) i	nput is from A	N4		
			1 (-) i	nput is from E	and Gap Referer	nce Voltage	
		ACO		ampled by S	alue of ACOUT ( CLK, system cloo		
			0 Cor	nparator outp	ut is LOW		
			1 Cor	nparator outp	ut is HIGH		
		ACIF	according to	o the ACISI	an Analog Comp M[1:0] bits. This xecuted or '0' is w	bit is clear	red when Anal
			0 No	interrupt gene	erated or cleared		
			1 Inte	rrupt generat	ed		
		ACIE	Enable Anal	og Comparat	or Interrupt.		
			0 Disa	able Interrupt	, Polling mode op	eration	
			1 Ena	able Interrupt			
	AC	O_OUTEN	Analog Com	parator outpu	it port Enable		
			0 Disa	able			
			1 Ena	able			
	A	CISM[1:0]	Select Inter	rupt Mode of	Analog Comparat	tor.	
			ACISM1	ACISM0	Description		
			0	0	Reserved		
			0	1	Interrupt on falling	0 0	
			1	0	Interrupt on rising		
			1	1	Interrupt on both	edge of ACO	UT

### ACCSR (Analog Comparator Control & Status Register) : F9H

# **12. Power Down Operation**

## 12.1 Overview

The Z51F0811 has three power-down modes to minimize the power consumption of the device. In power down mode, power consumption is reduced considerably. The device provides three kinds of power saving functions, IDLE, STOP1 and STOP2 mode. In three modes, program is stopped.

## 12.2 Peripheral Operation in IDLE/STOP Mode

Peripheral	IDLE Mode	STOP1 Mode	STOP2 Mode
CPU	ALL CPU Operation are Disable	ALL CPU Operation are Disable	ALL CPU Operation are Disable
RAM	Retain	Retain	Retain
Basic Interval Timer	Operates Continuously	Operates Continuously	Stop
Watch Dog Timer	Operates Continuously	Operates Continuously	Stop
Watch Timer	Operates Continuously	Stop (Only operate in sub clock mode)	Stop (Only operate in sub clock mode)
TimerP0~4	Operates Continuously	Halted (Only when the Event Counter Mode is Enable, Timer operates Normally)	Halted (Only when the Event Counter Mode is Enable, Timer operates Normally)
ADC	Operates Continuously	Stop	Stop
BUZ	Operates Continuously	Stop	Stop
SPI/SCI	Operates Continuously	Only operate with external clock	Only operate with external clock
I <sup>2</sup> C	Operates Continuously	Stop	Stop
Internal OSC (8MHz)	Oscillation	Stop	Stop
Main OSC (1~12MHz)	Oscillation	Stop	Stop
Sub OSC (32.768kHz)	Oscillation	Oscillation	Oscillation
Internal RCOSC (125kHz)	Oscillation	Oscillation	Stop
I/O Port	Retain	Retain	Retain
Control Register	Retain	Retain	Retain
Address Data Bus	Retain	Retain	Retain
Release Method	By RESET, all Interrupts	By RESET, Timer Interrupt (EC0, EC2), SIO (External clock), External Interrupt, UART by ACK PCI, I <sup>2</sup> C (slave mode), WT (sub clock),WDT, BIT	By RESET, Timer Interrupt (EC0, EC2), SIO (External clock), External Interrupt, UART by ACK PCI, I <sup>2</sup> C (slave mode), WT (sub clock)

Table 12-1 Peripheral Operation during Power Down Mode.

### 12.3 IDLE mode

The power control register is set to '01h' to enter the IDLE Mode. In this mode, the internal oscillation circuits remain active. Oscillation continues and peripherals are operated normally but CPU stops. It is released by reset or interrupt. To be released by interrupt, interrupt should be enabled before IDLE mode. If using reset, because the device becomes initialized state, the registers have reset value.



Figure 12-1 IDLE Mode Release Timing by External Interrupt



Figure 12-2 IDLE Mode Release Timing by /RESET

(Ex) MOV PCON, #0000\_0001b ; setting of IDLE mode : set the bit of STOP and IDLE Control register (PCON)

### 12.4 STOP mode

The power control register is set to '03h' to enter the STOP Mode. In the stop mode, the main oscillator, system clock and peripheral clock is stopped, but watch timer continue to operate. With the clock frozen, all functions are stopped, but the on-chip RAM and control registers are held.

The source for exit from STOP mode is hardware reset and interrupts. The reset re-defines all the control registers.

When exit from STOP mode, enough oscillation stabilization time is required to normal operation. Figure 12-3 shows the timing diagram. When released from STOP mode, the Basic interval timer is activated on wake-up. Therefore, before STOP instruction, user must be set its relevant prescaler divide ratio to have long enough time (more than 20msec). this guarantees that oscillator has started and stabilized.



Figure 12-3 STOP Mode Release Timing by External Interrupt



Figure 12-4 STOP Mode Release Timing by /RESET

## 12.5 Release Operation of STOP1, 2 Mode

After STOP1, 2 mode is released, the operation begins according to content of related interrupt register just before STOP1, 2 mode start (Figure 12-5). Interrupt Enable Flag of All (EA) of IE should be set to `1`. Released by only interrupt which each interrupt enable flag = `1`, and jump to the relevant interrupt service routine.



Figure 12-5 STOP1, 2 Mode Release Flow

#### 12.5.1 Register Map

Table 12-2 Register Map

Name	Address	Dir	Default	Description
PCON	87H	R/W	00H	Power Control Register

#### 12.5.2 Power Down Operation Register description

The Power Down Operation Register consists of the Power Control Register (PCON).

#### 12.5.3 Register description for Power Down Operation

#### PCON (Power Control Register) : 87H



Note)

- 1. To enter IDLE mode, PCON must be set to '01H'.
- To STOP1,2 mode, PCON must be set to '03H'. (In STOP1,2 mode, PCON register is cleared automatically by interrupt or reset)
- 3. When PCON is set to '03H', if SCCR[7] is set to '1', it enters the STOP1 mode. if SCCR[7] is cleared to '0', it enters the STOP2 mode
- 4. The different thing in STOP 1,2 is only clock operation of internal 125kHz-OSC during STOP mode operating.
# 13. RESET

## 13.1 Overview

The Z51F0811 has reset by external RESETB pin. The following is the hardware setting value.

Table 13-1	Reset state
------------	-------------

On Chip Hardware	Initial Value		
Program Counter (PC)	0000h		
Accumulator	00h		
Stack Pointer (SP)	07h		
Peripheral Clock	On		
Control Register	Peripheral Registers refer		
Brown-Out Detector	Enable		

### 13.2 Reset source

The Z51F0811 has five types of reset generation procedures. The following is the reset sources.

- External RESETB
- Power ON RESET (POR)
- WDT Overflow Reset (In the case of WDTEN = `1`)
- BOD Reset (In the case of BODEN = `1 `)
- OCD Reset

### 13.3 Block Diagram



Figure 13-1 RESET Block Diagram

# 13.4 RESET Noise Canceller

The Figure 13-2 is the Noise canceller diagram for Noise cancel of RESET. It has the Noise cancel value of about 7us ( $@V_{DD}=5V$ ) to the low input of System Reset.



Figure 13-2 Reset noise canceller time diagram

# **13.5 Power ON RESET**

When rising device power, the POR (Power ON Reset) have a function to reset the device. If using POR, it executes the device RESET function instead of the RESET IC or the RESET circuits. And External RESET PIN is able to use as Normal I/O pin.



Figure 13-3 Fast VDD rising time



Figure 13-4 Internal RESET Release Timing On Power-Up



Figure 13-5 Configuration timing when Power-on



Figure 13-6 Boot Process Waveform

Table 13-2 Boot Process Descrip	otion
---------------------------------	-------

Process	Description	Remarks
1	-No Operation	
2	-1st POR level Detection -Internal OSC (125KHz) ON	-about 1.4V ~ 1.5V
3	<ul> <li>- (INT-OSC125KHz/32)×30h Delay section (=12ms)</li> <li>-VDD input voltage must rise over than flash operating voltage for Config read</li> </ul>	-Slew Rate >= 0.025V/ms
4	- Config read point	-about 1.5V ~ 1.6V -Config Value is determined by Writing Option
5	- Rising section to Reset Release Level	-16ms point after POR or Ext_reset release
6	<ul> <li>Reset Release section (BIT overflow)</li> <li>i) after16ms, after External Reset Release (External reset)</li> <li>ii) 16ms point after POR (POR only)</li> </ul>	- BIT is used for Peripheral stability
0	-Normal operation	

### **13.6 External RESETB Input**

The External RESETB is the input to a Schmitt trigger. A reset in accomplished by holding the reset pin low for at least 7us over, within the operating voltage range and oscillation stable, it is applied, and the internal state is initialized. After reset state becomes '1', it needs the stabilization time with 16ms and after the stable state, the internal RESET becomes '1'. The Reset process step needs 5 oscillator clocks. And the program execution starts at the vector address stored at address 0000H.



Figure 13-7 Timing Diagram after RESET



Figure 13-8 Oscillator generating waveform example

Note) as shown Figure 13-8, the stable generating time is not included in the start-up time.

# **13.7 Brown Out Detector Processor**

The Z51F0811 has an On-chip Brown-out detection circuit for monitoring the VDD level during operation by comparing it to a fixed trigger level. The trigger level for the BOD can be selected by BODLS[1:0] bit to be 1.6V, 2.5V, 3.6V or 4.2V. In the STOP mode, this will contribute significantly to the total current consumption. So to minimize the current consumption, the BODEN bit is set to off by software.



Figure 13-9 Block Diagram of BOD





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Figure 13-11 Configuration timing when BOD RESET

### 13.7.1 Register Map

Table 13-3 Register Map

Name	Address	Dir	Default	Description
BODR	86H	R/W	81H	BOD Control Register

### 13.7.2 Reset Operation Register description

Reset control Register consists of the BOD Control Register (BODR).

### 13.7.3 Register description for Reset Operation

### BODR (BOD Control Register) : 86H

7	6	5	4	3	2	1	0	
PORF	EXTRF	WDTRF	OCDRF	BODRF	BODLS[1]	BODLS[0]	BODEN	
RW	RW	RW	RW	RW	RW	RW	RW	
						I	nitial value : 87	1H

**PORF** Power-On Reset flag bit. The bit is reset by writing '0' to this bit.

	0	No detection	
	1	Detection	
EXTRF	Externa ON res	0	The bit is reset by writing '0' to this bit or by Power
	0	No detection	
	1	Detection	
WDTRF		Dog Reset flag I ON reset.	bit. The bit is reset by writing '0' to this bit or by
	0	No detection	
	1	Detection	
OCDRF		ip Debug Reset fl ON reset.	ag bit. The bit is reset by writing '0' to this bit or by
	0	No detection	
	1	Detection	
BODRF		Out Reset flag b ON reset.	bit. The bit is reset by writing '0' to this bit or by
	0	No detection	
	1	Detection	
BODLS[1:0]	BOD I	evel Voltage	
	BODL	S1 BODLS0	Description
	0	0	1.6V
	0	1	2.5V
	1	0	3.6V
	1	1	4.2V
BODEN	BOD op	peration	
	0	BOD disable	
	1	BOD enable	

# 14. On-chip Debug System

### 14.1 Overview

### 14.1.1 Description

On-chip debug System (OCD) of Z51F0811 can be used for programming the non-volatile memories and on-chip debugging. Detailed descriptions for programming via the OCD interface can be found in the following chapter.

Figure 14-1 shows a block diagram of the OCD interface and the On-chip Debug system.

### 14.1.2 Feature

- Two-wire external interface: 1-wire serial clock input, 1-wire bi-directional serial data bus
- Debugger Access to:
  - All Internal Peripheral Units
  - Internal data RAM
  - Program Counter
  - Flash and Data EEPROM Memories
- Extensive On-chip Debug Support for Break Conditions, Including
  - Break Instruction
  - Single Step Break
  - Program Memory Break Points on Single Address
  - Programming of Flash, EEPROM, Fuses, and Lock Bits through the two-wire Interface
  - On-chip Debugging Supported by Dr.Choice®
- Operating frequency

Supports the maximum frequency of the target MCU



Figure 14-1 Block Diagram of On-chip Debug System

### 14.2 Two-pin external interface

### 14.2.1 Basic transmission packet

- 10-bit packet transmission using two-pin interface.
- 1-packet consists of 8-bit data, 1-bit parity and 1-bit acknowledge.
- Parity is even of '1' for 8-bit data in transmitter.

• Receiver generates acknowledge bit as '0' when transmission for 8-bit data and its parity has no error.

- When transmitter has no acknowledge (Acknowledge bit is '1' at tenth clock), error process is executed in transmitter.
- When acknowledge error is generated, host PC makes stop condition and transmits command which has error again.
- Background debugger command is composed of a bundle of packet.

• Star condition and stop condition notify the start and the stop of background debugger command respectively.



Figure 14-2 10-bit transmission packet

### 14.2.2 Packet transmission timing

### 14.2.2.1 Data transfer





### 14.2.2.2 Bit transfer



Figure 14-4 Bit transfer on the serial bus

### 14.2.2.3 Start and stop condition



Figure 14-5 Start and stop condition

### 14.2.2.4 Acknowledge bit



Figure 14-6 Acknowledge on the serial bus



Figure 14-7 Clock synchronization during wait procedure

### 14.2.3 Connection of transmission

Two-pin interface connection uses open-drain (wire-AND bidirectional I/O).



Figure 14-8 Connection of transmission

# **15. Memory Programming**

### 15.1 Overview

### 15.1.1 Description

Z51F0811 incorporates flash and data EEPROM memory to which a program can be written, erased, and overwritten while mounted on the board. Also, data EEPROM can be programmed or erased in user program. Flash area can be programmed in only OCD or parallel ROM mode.

Serial ISP modes and byte-parallel ROM writer mode are supported.

### 15.1.2 Features

- Flash Size : 8Kbytes
- Single power supply program and erase
- Command interface for fast program and erase operation
- Up to 10,000 program/erase cycles at typical voltage and temperature for flash memory
- Up to 100,000 program/erase cycles at typical voltage and temperature for data EEPROM memory
- Security feature

### 15.2 Flash and EEPROM Control and status register

Registers to control Flash and Data EEPROM are Mode Register (FEMR), Control Register (FECR), Status Register (FESR), Time Control Register (FETCR), Address Low Register (FEARL), Address Middle Register (FEARM), address High Register (FEARH) and Data Register (FEDR). They are mapped to SFR area and can be accessed only in programming mode.

### 15.2.1 Register Map

Table 15-1 Register Map

Name	Address	Dir	Default	Description
FEMR	EAH	R/W	00H	Flash and EEPROM Mode Register
FECR	EBH	R/W	03H	Flash and EEPROM Control Register
FESR ECH		R/W	80H	Flash and EEPROM Status Register
FETCR	EDH	R/W	00H	Flash and EEPROM Time Control Register
FEARL	F2H	R/W	00H	Flash and EEPROM Address Low Register
FEARM	F3H	R/W	00H	Flash and EEPROM Address Middle Register
FEARH F4H		R/W	00H	Flash and EEPROM Address High Register
FEDR	F5H	R/W	00H	Flash and EEPROM Data Register

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### 15.2.2 Register description for Flash and EEPROM

7	6	5	4	3	2	1	0		
FSEL	ESEL	PGM	ERASE	PBUFF	OTPE	VFY	FEEN		
RW	RW	RW	RW	RW	RW	RW	RW		
							Initial value : 00		
		FSEL	Select flash m	emory.					
			0 Dese	lect flash mem	ory				
			1 Selec	t flash memory	/				
	ESEL Select data EEPROM								
			0 Dese	lect data EEPF	ROM				
			1 Selec	t data EEPRO	M				
		PGM	Enable progra	m or program	verify mode w	ith VFY			
			0 Disat	le program or	program verify	/ mode			
			1 Enab	le program or p	program verify	mode			
		ERASE	Enable erase	or erase verify	mode with VF	Ϋ́			
			0 Disat	le erase or era	ase verify mod	е			
			1 Enab	le erase or era	se verify mode	e			
		PBUFF	Select page b	uffer					
			0 Dese	lect page buffe	r				
			1 Selec	t page buffer					
		OTPE		ea instead of p	program memo	ory			
				lect OTP area					
			1 Selec	t OTP area					
		VFY	Set program or erase verify mode with PGM or ERASE						
			Program Verif	y: PGM=1, VF	Y=1				
			Erase Verify:	ERASE=1, VF	Y=1				
		FEEN		m and erase or read as normal		ata EEPRON	M. When inactive		
			0 Disat	le program an	d erase				
			1 Enab	le program and	derase				

### FEMR (Flash and EEPROM Mode Register) : EAH

### FECR (Flash and EEPROM Control Register) : EBH

7	6	5	4	3	2	1	0		
AEF	AEE	EXIT1	EXIT0	WRITE	READ	nFERST	nPBRST		
RW	RW	RW	RW	RW	RW	RW	RW		
						I	nitial value : 03		
		AEF	Enable flash	n bulk erase mo	de				
			0 Dis	able bulk erase	mode of Flash	memory			
			1 Ena	able bulk erase	mode of Flash	memory			
		AEE	Enable data EEPROM bulk erase mode						
			0 Dis	able bulk erase	mode of data l	EEPROM			
			1 Enable bulk erase mode of data EEPROM						
	I	<b>EXIT[1:0]</b> Exit from program mode. It is cleared automatically after 1 clock							
			EXIT1	EXITO D	escription				
			0	0 D	on't exit from p	rogram mode			
			0	1 D	on't exit from p	rogram mode			
			1	0 D	on't exit from p	rogram mode			
			1	1 E	xit from progra	m mode			
		WRITE		ogram or erase y after 1 clock	e of Flash and	d data EEPR	OM. It is clear		
			0 No	operation					
			1 Sta	rt to program o	erase of Flash	and data EEI	PROM		
		READ	Start auto-v after 1 clock		or data EEPR	OM. It is clea	ared automatica		
			0 No	operation					
			1 Sta	rt auto-verify of	Flash or data I	EPROM			
		nFERST	Reset Flash after 1 clock		ROM control lo	ogic. It is clea	ared automatica		
			0 No operation						
			1 Reset Flash or data EEPROM control logic.						
	I	nPBRST	Reset page	buffer with PBL	JFF. It is cleare	d automaticall	y after 1 clock		
			PBUFF	nPBRST	Description				
			0	0	Page buffer re	eset			
			1	0	Write checksu	ım reset			

WRITE and READ bits can be used in program, erase and verify mode with FEAR registers. Read or writes for memory cell or page buffer uses read and write enable signals from memory controller. Indirect address mode with FEAR is only allowed to program, erase and verify

### FESR (Flash and EEPROM Status Register) : ECH

7	6	5	4	3	2	1	0		
PEVBSY	VFYGOOD	-	-	ROMINT	WMODE	EMODE	VMODE		
R	RW	R	R	RW	R	R	R		
						I	nitial value : 80	)H	
	P		Operation status flag. It is cleared automatically when operation starts. Operations are program, erase or verification						
			0 Busy (	Operation pro	cessing)				
			1 Complete Operation						
	VF	YGOOD	Auto-verification result flag.						
			0 Auto-verification fails						
			1 Auto-v	erification suc	cesses				
	F		Flash and Da program/erase				uto-cleared wh fy completion	hen	
				errupt request					
			1 Interrupt request.						
	V	VMODE	Write mode fla	g					
	E	EMODE	Erase mode fla	ag					
	١	/MODE	Verify mode fla	g					

### FEARL (Flash and EEPROM address low Register) : F2H

7	6	5	4	3	2	1	0	
ARL7	ARL6	ARL5	ARL4	ARL3	ARL2	ARL1	ARLO	
W	W	W	W	W	W	W	W	
						I	nitial value : 0	ΟH



#### FEARM (Flash and EEPROM address middle Register) : F3H

7	6	5	4	3	2	1	0
ARM7	ARM6	ARM5	ARM4	ARM3	ARM2	ARM1	ARMO
W	W	W	W	W	W	W	W
						I	nitial value : 00

ARM[7:0] Flash and EEPROM address middle

### FEARH (Flash and EEPROM address high Register) : F4H

7	6			3	2	1	0
ARH7	ARH6	ARH5	ARH4	ARH3	ARH2	ARH1	ARH0
W	W	W	W	W	W	W	W
						I	nitial value : 00H

ARH[7:0] Flash and EEPROM address high

FEAR registers are used for program, erase and auto-verify. In program and erase mode, it is page address and ignored the same least significant bits as the number of bits of page address. In auto-verify mode, address increases automatically by one.

FEARs are write-only register. Reading these registers returns 24-bit checksum result

### FEDR (Flash and EEPROM data control Register) : F5H

7	6	5	4	3	2	1	0
FEDR7	FEDR6	FEDR5	FEDR4	FEDR3	FEDR2	FEDR1	FEDR0
W	W	W	W	W	W	W	W
						I	nitial value : 00H

FEDR[7:0] Flash and EEPROM data

Data register. In no program/erase/verify mode, READ/WRITE of FECR read or write data from EEPROM or Flash to this register or from this register to Flash or EEPROM.

The sequence of writing data to this register is used for EEPROM program entry. The mode entrance sequence is to write 0xA5 and 0x5A to it in order.

#### FETCR (Flash and EEPROM Time control Register) : EDH





Program and erase time is controlled by setting FETCR register. Program and erase timer uses 10bit counter. It increases by one at each divided system clock frequency(=SCLK/128). It is cleared when program or erase starts. Timer stops when 10-bit counter is same to FETCR. PEVBSY is cleared when program, erase or verify starts and set when program, erase or verify stops.

Max program/erase time at 12Mhz system clock : (255+1) \* 2 \* (83.3ns \* 128) = 5.459ms

In the case of 10% of error rate of counter source clock, program or erase time is 5.0~5.9ms

\* \* Program/erase time calculation

for page write or erase, Tpe = (TCON+1) \* 2 \* (SCLK \* 128)

for bulk erase, Tbe = (TCON+1) \* 4 \* (SCLK \* 128)

Table 15-2 Program/erase Time

	Min	Тур	Max	Unit
program/erase Time	2.4	2.5	2.6	ms

※ Recommended program/erase time at 12Mhz (FETCR = 75h)

### 15.3 Memory map

### 15.3.1 Flash Memory Map

Program memory uses 8-Kbyte of Flash memory. It is read by byte and written by byte or page. One page is 32-byte









### 15.3.2 Data EEPROM Memory Map

Data EEPROM memory uses 512-byte of EEPROM. It is read by byte and written by byte or page. One page is 16-byte. It is mapped to external data memory of 8051



Figure 15-3 Data EEPROM memory map



Figure 15-4 Address configuration of data EEPROM

## 15.4 Serial In-System Program Mode

Serial in-system program uses the interface of debugger which uses two wires. Refer to chapter 14 in details about debugger

### 15.4.1 Flash operation

Configuration (This Configuration is just used for follow description)

7	6	5	4	3	2	1	0
-	FEMR[4] & [1]	FEMR[5] &[1]	-	-	FEMR[2]	FECR[6]	FECR[7]
-	ERASE&VFY	PGM&VFY	-	-	OTPE	AEE	AEF



Figure 15-5 The sequence of page program and erase of Flash memory

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Figure 15-6 The sequence of bulk erase of Flash memory

### 15.4.1.1 Flash Read

- Step 1. Enter OCD(=ISP) mode.
- Step 2. Set ENBDM bit of BCR.
- Step 3. Enable debug and Request debug mode.
- Step 4. Read data from Flash.

### 15.4.1.2 Enable program mode

- Step 1. Enter OCD(=ISP) mode.<sup>1</sup>
- Step 2. Set ENBDM bit of BCR.
- Step 3. Enable debug and Request debug mode.
- Step 4. Enter program/erase mode sequence.<sup>2</sup>
  - (1) Write 0xAA to 0xF555.

(2) Write 0x55 to 0xFAAA.

(3) Write 0xA5 to 0xF555.

<sup>1</sup> Refer to how to enter ISP mode..

<sup>2</sup> Command sequence to activate Flash write/erase mode. It is composed of sequentially writing data of Flash memory.

### 15.4.1.3 Flash write mode

Step 1. Enable program mode.

Step 2. Reset page buffer. FEMR: 1000\_0001 FECR:0000\_0010

Step 3. Select page buffer. FEMR:1000\_1001

Step 4. Write data to page buffer.(Address automatically increases by twin.)

Step 5. Set write mode. FEMR:1010\_0001

Step 6. Set page address. FEARH:FEARM:FEARL=20'hx\_xxxx

Step 7. Set FETCR.

Step 8. Start program. FECR:0000\_1011

Step 9. Insert one NOP operation

Step 10. Read FESR until PEVBSY is 1.

Step 11. Repeat step2 to step 8 until all pages are written.

### 15.4.1.4 Flash page erase mode

- Step 1. Enable program mode.
- Step 2. Reset page buffer. FEMR: 1000\_0001 FECR:0000\_0010
- Step 3. Select page buffer. FEMR:1000\_1001
- Step 4. Write 'h00 to page buffer. (Data value is not important.)
- Step 5. Set erase mode. FEMR:1001\_0001

Step 6. Set page address. FEARH:FEARM:FEARL=20'hx\_xxxx

Step 7. Set FETCR.

Step 8. Start erase. FECR:0000\_1011

- Step 9. Insert one NOP operation
- Step 10. Read FESR until PEVBSY is 1.

Step 11. Repeat step2 to step 8 until all pages are erased.

### 15.4.1.5 Flash bulk erase mode

Step 1. Enable program mode.

Step 2. Reset page buffer. FEMR: 1000\_0001 FECR:0000\_0010

Step 3. Select page buffer. FEMR:1000\_1001

Step 4. Write 'h00 to page buffer. (Data value is not important.)

Step 5. Set erase mode. FEMR:1001\_0001.

(Only main cell area is erased. For bulk erase including OTP area, select OTP area.(set FEMR to 1000\_1101.)

Step 6. Set FETCR

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Step 7. Start bulk erase. FECR:1000\_1011 Step 8. Insert one NOP operation Step 9. Read FESR until PEVBSY is 1.

### 15.4.1.6 Flash OTP area read mode

Step 1. Enter OCD(=ISP) mode.

Step 2. Set ENBDM bit of BCR.

Step 3. Enable debug and Request debug mode.

Step 4. Select OTP area. FEMR:1000\_0101

Step 5. Read data from Flash.

### 15.4.1.7 Flash OTP area write mode

Step 1. Enable program mode. Step 2. Reset page buffer. FEMR: 1000\_0001 FECR:0000\_0010 Step 3. Select page buffer. FEMR:1000\_1001 Step 4. Write data to page buffer.(Address automatically increases by twin.) Step 5. Set write mode and select OTP area. FEMR:1010\_0101 Step 6. Set page address. FEARH:FEARM:FEARL=20'hx\_xxxx Step 7. Set FETCR. Step 8. Start program. FECR:0000\_1011 Step 9. Insert one NOP operation Step 10. Read FESR until PEVBSY is 1.

### 15.4.1.8 Flash OTP area erase mode

Step 1. Enable program mode.
Step 2. Reset page buffer. FEMR: 1000\_0001 FECR:0000\_0010
Step 3. Select page buffer. FEMR:1000\_1001
Step 4. Write 'h00 to page buffer. (Data value is not important.)
Step 5. Set erase mode and select OTP area. FEMR:1001\_0101
Step 6. Set page address. FEARH:FEARM:FEARL=20'hx\_xxxx
Step 7. Set FETCR.
Step 8. Start erase. FECR:0000\_1011
Step 9. Insert one NOP operation
Step 10. Read FESR until PEVBSY is 1.

### 15.4.1.9 Flash program verify mode

Step 1. Enable program mode.

Step 2. Set program verify mode. FEMR:1010\_0011

Step 3. Read data from Flash.

### 15.4.1.10 OTP program verify mode

Step 1. Enable program mode. Step 2. Set program verify mode. FEMR:1010\_0111 Step 3. Read data from Flash.

### 15.4.1.11 Flash erase verify mode

Step 1. Enable program mode. Step 2. Set erase verify mode. FEMR:1001\_0011 Step 3. Read data from Flash.

### 15.4.1.12 Flash page buffer read

Step 1. Enable program mode. Step 2. Select page buffer. FEMR:1000\_1001 Step 3. Read data from Flash.

### 15.4.2 Data EEPROM operation

Program and erase operation of Data EEPROM are executed by direct and indirect address mode. Direct address mode uses external data area of 8051. Indirect address mode uses address register of SFR area..

### 15.4.2.1 Data EEPROM Read

- Step 1. Enter OCD(=ISP) mode.
- Step 2. Set ENBDM bit of BCR.
- Step 3. Enable debug and Request debug mode.
- Step 4. Read data from Data EEPROM.

### 15.4.2.2 Enable program mode

- Step 1. Enter OCD(=ISP) mode.<sup>1</sup>
- Step 2. Set ENBDM bit of BCR.
- Step 3. Enable debug and Request debug mode.
- Step 4. Enter program/erase mode sequence.<sup>2</sup>
  - (1) Write 0xA5 to FEDR.
  - (2) Write 0x5A to FEDR.

<sup>1</sup> Refer to how to enter ISP mode..

<sup>2</sup> Command sequence to activate data EEPROM write/erase mode. It is composed of sequentially writing to data register(FEDR)

### 15.4.2.3 EEPROM write mode

- Step 1. Enable program mode.
- Step 2. Reset page buffer. FEMR: 0100\_0001 FECR:0000\_0010
- Step 3. Select page buffer. FEMR:0100\_1001
- Step 4. Write data to page buffer.(Address automatically increases by twin.)
- Step 5. Set write mode. FEMR:0110\_0001
- Step 6. Set page address. FEARH:FEARM:FEARL=20'hx\_xxxx
- Step 7. Set FETCR.
- Step 8. Start program. FECR:0000\_1011
- Step 9. Insert one NOP operation
- Step 10. Read FESR until PEVBSY is 1.
- Step 11. Repeat step2 to step 8 until all pages are written.

### 15.4.2.4 EEPROM page erase mode

- Step 1. Enable program mode.
- Step 2. Reset page buffer. FEMR: 0100\_0001 FECR:0000\_0010
- Step 3. Select page buffer. FEMR:0100\_1001
- Step 4. Write 'h00 to page buffer. (Data value is not important.)
- Step 5. Set erase mode. FEMR:0101\_0001
- Step 6. Set page address. FEARH:FEARM:FEARL=20'hx\_xxxx
- Step 7. Set FETCR.
- Step 8. Start erase. FECR:0000\_1011
- Step 9. Insert one NOP operation
- Step 10. Read FESR until PEVBSY is 1.
- Step 11. Repeat step2 to step 8 until all pages are erased.

### 15.4.2.5 EEPROM bulk erase mode

- Step 1. Enable program mode.
- Step 2. Reset page buffer. FEMR: 0100\_0001 FECR:0000\_0010
- Step 3. Select page buffer. FEMR:0100\_1001
- Step 4. Write 'h00 to page buffer. (Data value is not important.)
- Step 5. Set erase mode. FEMR:0101\_0001.
- Step 6. Set FETCR
- Step 7. Start bulk erase. FECR:0100\_1011
- Step 8. Insert one NOP operation
- Step 9. Read FESR until PEVBSY is 1.

### 15.4.2.6 Data EEPROM program verify mode

 Step 1. Enable program mode.

 Step 2. Set program verify mode. FEMR:0110\_0011

 Step 3. Read data from Flash.

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### 15.4.2.7 Data EEPROM erase verify mode

Step 1. Enable program mode. Step 2. Set erase verify mode. FEMR:0101\_0011 Step 3. Read data from Flash.

### 15.4.2.8 Data EEPROM page buffer read

Step 1. Enable program mode. Step 2. Select page buffer. FEMR:0100\_1001 Step 3. Read data from Flash.

### 15.4.3 Summary of Flash and Data EEPROM Program/Erase Mode

Table 15-3 Operation Mode

Oper	ation mode	Description
	Flash read	Read cell by byte.
F	Flash write	Write cell by bytes or page.
L	Flash page erase	Erase cell by page.
А	Flash bulk erase	Erase the whole cells.
S	Flash program verify	Read cell in verify mode after programming.
н	Flash erase verify	Read cell in verify mode after erase.
	Flash page buffer load	Load data to page buffer.
	Data EEPROM read	Read cell by byte.
E	Data EEPROM write	Write cell by bytes or page.
E	Data EEPROM page erase	Erase cell by page.
P	Data EEPROM bulk erase	Erase the whole cells.
R O	Data EEPROM program verify	Read cell in verify mode after programming.
м	Data EEPROM erase verify	Read cell in verify mode after erase.
	Data EEPROM page buffer load	Load data to page buffer.

### 15.5 Parallel Mode

### 15.5.1 Overview

Parallel program mode transfers address and data by byte. 3-byte address can be entered by one from the lease significant byte of address. If only LSB is changed, only one byte can be transferred. And if the second byte is changed, the first and second byte can be transferred. Upper 4-bit of the most significant byte selects memory to be accessed. Table 15-4 shows memory type to be accessible by parallel mode. Address auto-increment is supported when read or write data without address.

The erase and program sequence of Flash and data EEPROM is identical to that of ISP mode except the entrance of parallel mode. Refer to Table 15-5 for the entrance method for parallel mode.



Figure 15-7 Pin diagram for parallel programming

ADDRH[7	:4]			Memory Type
0	0 0 0		0	Program Memory
0	0 0 1		1	External Memory
0	0	1	0	SFR

Table 15-4	The selection	of memory type	by ADDRH[7:4]
10010 10 1	1110 001000001	or mornory type	<i>by reporting</i>

### 15.5.2 Parallel Mode instruction format

Instruction	Signal	Instr	uction	Sequen	ice										
	nALE	L		L		L		н		н		н		н	
n-byte data read	nWR	L	н	L	н	L	Н	н	Н	н	н	н	Н	Н	н
with 3-byte address	nRD	н	н	н	н	н	н	L	н	L	н	L	Н	L	н
	PDATA	ADDRL		ADD	RM	ADD	RH	DAT	A0	DAT	A1			DATAn	
	nALE	L		L		L		н		н		н		н	
n-byte data write	nWR	L	н	L	н	L	н	L	н	L	н	L	н	L	Н
with 3-byte address	nRD	н н		н	н	н	н	н	н	н	н	н	н	н	н
	PDATA	ADD	RL	ADD	RM	ADD	RH	DAT	A0	DAT	A1			DAT	'An
	nALE	L		L		н		н		н		н		н	
n-byte data read with 2-byte address	nWR	L	н	L	н	н	н	н	н	н	н	н	н	н	н
	nRD	нн		н	н	L	н	L	Н	L	н	L	н	L	н
	PDATA	ADDRL		ADDRM		DATA0		DATA1		DAT	A2			DAT	'An
	nALE	L		L		н		н		н		н		н	
n-byte data write with 2-byte	nWR	L	н	L	н	L	н	L	н	L	н	L	н	L	н
address	nRD	н	н	н	н	н	н	н	н	н	н	н	н	н	н
	PDATA	ADD	RL	ADD	RM	DATA0		DATA1		DATA2				DATAn	
	nALE	L		н	•	н		н		н		н		н	
n-byte data read	nWR	L	н	н	н	L	н	L	н	L	н	L	н	L	н
with 1-byte address	nRD	н	н	L	н	н	н	н	н	н	н	н	н	н	н
	PDATA	ADD	RL	DAT	A0	DAT	A1	DAT	A2	DATA3				DAT	'An
	nALE	L		н	T	н		н		н		н	н		
n-byte data write with 1-byte	nWR	L	н	L	н	L	н	L	н	L	н	L	н	L	н
address	nRD	н	н	н	н	н	Н	н	Н	н	н	н	Н	н	н
	PDATA	ADD	RL	DAT	A0	DAT	A1	DAT	A2	DAT	A3			DAT	'An

Table 15-5 Parallel mode instruction format





Figure 15-8 Parallel Byte Read Timing of Program Memory



Figure 15-9 Parallel Byte Write Timing of Program Memory

## 15.6 Mode entrance method of ISP and byte-parallel mode

### 15.6.1 Mode entrance method for ISP

TARGET MODE	DSDA	DSCL	DSDA
OCD(ISP)	'hC	'hC	'hC



Figure 15-10 ISP mode

### 15.6.2 Mode entrance of Byte-parallel

TARGET MODE	P0[3:0]	P0[3:0]	P0[3:0]
Byte-Parallel Mode	4'h5	4'hA	4'h5



Figure 15-11 Byte-parallel mode

# 15.7 Security

Z51F0811 provides Lock bits which can be left unprogrammed ("0") or can be programmed ("1") to obtain the additional features listed in Table 15-6. The Lock bits can be erased to "0" with only the bulk erase command and a value of more than 0x80 at FETCR.

	01	USER MODE										ISP/PMODE													
	OCK DE		FI	ash				ATA PROM	1	OTP Fi			Flash			DATA EEPROM				OTP					
LOC KE	LOC KF	R	w	P E	B E	R	w	P E	B E	R	w	P E	B E	R	w	P E	B E	R	w	P E	B E	R	w	P E	B E
0	0	0	0	0	Х	0	0	0	0	х	х	Х	Х	0	0	0	0	0	0	0	0	0	0	0	0
0	1	0	0	0	Х	0	0	0	0	х	х	Х	Х	х	Х	Х	0	0	0	0	0	0	х	Х	0
1	0	0	0	0	Х	0	0	0	0	х	х	Х	Х	0	$\diamond$	$\diamond$	$\diamond$	х	х	Х	0	0	$\diamond$	$\diamond$	$\diamond$
1	1	0	0	0	Х	0	0	0	0	х	х	Х	Х	Х	Х	Х	$\diamond$	х	Х	Х	0	0	х	Х	$\diamond$

Table 15-6 Security policy using lock-bits

- LOCKF: Lock bit of Flash memory
- LOCKE: Lock bit of data EEPROM
- R: Read
- W: Write
- PE: Page erase
- BE: Bulk Erase
- O: Operation is possible.
- X: Operation is impossible.

# 16. Configure option

# 16.1 Configure option Control Register

### FUSE\_CONF (Pseudo-Configure Data) : 2F5DH

7	6	5	4	3	2	1	0
BSIZE1	BSIZE0	SXINEN	XINENA	RSTDIS	LOCKB	LOCKE	LOCKF
R	R	R	R	R	R	R	R
						ļ	Initial value : 00
		BSIZE	Boot Area Har	d Lock Size Bi	it		
			00 0100H	l~01FFH (defa	ault)		
			01 0100H	l~03FFH			
			10 0100H	l~07FFH			
			11 0100H	I~0FFFH			
	5	SXINEN	External Sub C	Oscillator Enab	ole Bit		
			0 Sub C	SC disable (d	efault)		
			1 Sub C	SC Enable			
	)	<b>(INENA</b>	External Main	Oscillator Ena	ble Bit		
			0 Main (	OSC disable (	default)		
			1 Main (	OSC Enable			
	F	RSTDIS	External RESE	TB disable Bi	t		
			0 Extern	al RESET en	able		
			1 Extern	al RESET dis	able		
	I	LOCKB	Boot Area Har	d LOCK(prote	ction) Bit		
			0 Hard I	OCK Disable			
			1 Hard I	OCK Enable			
	I	LOCKE	DATA memory	LOCK bit			
			0 LOCK	Disable			
			1 LOCK	Enable			
	I	LOCKF	CODE memory	/ LOCK bit			
			0 LOCK	Disable			
			1 LOCK	Enable			

# 17. APPENDIX

# A. Instruction Table

Instructions are either 1, 2 or 3 bytes long as listed in the 'Bytes' column below. Each instruction takes either 1, 2 or 4 machine cycles to execute as listed in the following table. 1 machine cycle comprises 2 system clock cycles.

	ARITHMETIC			
Mnemonic	Description	Bytes	Cycles	Hex code
ADD A,Rn	Add register to A	1	1	28-2F
ADD A,dir	Add direct byte to A	2	1	25
ADD A,@Ri	Add indirect memory to A	1	1	26-27
ADD A,#data	Add immediate to A	2	1	24
ADDC A,Rn	Add register to A with carry	1	1	38-3F
ADDC A,dir	Add direct byte to A with carry	2	1	35
ADDC A,@Ri	Add indirect memory to A with carry	1	1	36-37
ADDC A,#data	Add immediate to A with carry	2	1	34
SUBB A,Rn	Subtract register from A with borrow	1	1	98-9F
SUBB A,dir	Subtract direct byte from A with borrow	2	1	95
SUBB A,@Ri	Subtract indirect memory from A with borrow	1	1	96-97
SUBB A,#data	Subtract immediate from A with borrow	2	1	94
INC A	Increment A	1	1	04
INC Rn	Increment register	1	1	08-0F
INC dir	Increment direct byte	2	1	05
INC @Ri	Increment indirect memory	1	1	06-07
DEC A	Decrement A	1	1	14
DEC Rn	Decrement register	1	1	18-1F
DEC dir	Decrement direct byte	2	1	15
DEC @Ri	Decrement indirect memory	1	1	16-17
INC DPTR	Increment data pointer	1	2	A3
MUL AB	Multiply A by B	1	4	A4
DIV AB	Divide A by B	1	4	84
DA A	Decimal Adjust A	1	1	D4

	LOGICAL				
Mnemonic	Description	Bytes	Cycles	Hex code	
ANL A,Rn	AND register to A	1	1	58-5F	
ANL A,dir	AND direct byte to A	2	1	55	
ANL A,@Ri	AND indirect memory to A	1	1	56-57	
ANL A,#data	AND immediate to A	2	1	54	
ANL dir,A	AND A to direct byte	2	1	52	
ANL dir,#data	AND immediate to direct byte	3	2	53	
ORL A,Rn	OR register to A	1	1	48-4F	
ORL A,dir	OR direct byte to A	2	1	45	
ORL A,@Ri	OR indirect memory to A	1	1	46-47	
ORL A,#data	OR immediate to A	2	1	44	
ORL dir,A	OR A to direct byte	2	1	42	
ORL dir,#data	OR immediate to direct byte	3	2	43	
XRL A,Rn	Exclusive-OR register to A	1	1	68-6F	
XRL A,dir	Exclusive-OR direct byte to A	2	1	65	

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XRL A, @Ri	Exclusive-OR indirect memory to A	1	1	66-67
XRL A,#data	Exclusive-OR immediate to A	2	1	64
XRL dir,A	Exclusive-OR A to direct byte	2	1	62
XRL dir,#data	Exclusive-OR immediate to direct byte	3	2	63
CLR A	Clear A	1	1	E4
CPL A	Complement A	1	1	F4
SWAP A	Swap Nibbles of A	1	1	C4
RL A	Rotate A left	1	1	23
RLC A	Rotate A left through carry	1	1	33
RR A	Rotate A right	1	1	03
RRC A	Rotate A right through carry	1	1	13

DATA TRANSFER				
Mnemonic	Description	Bytes	Cycles	Hex code
MOV A,Rn	Move register to A	1	1	E8-EF
MOV A,dir	Move direct byte to A	2	1	E5
MOV A,@Ri	Move indirect memory to A	1	1	E6-E7
MOV A,#data	Move immediate to A	2	1	74
MOV Rn,A	Move A to register	1	1	F8-FF
MOV Rn,dir	Move direct byte to register	2	2	A8-AF
MOV Rn,#data	Move immediate to register	2	1	78-7F
MOV dir,A	Move A to direct byte	2	1	F5
MOV dir,Rn	Move register to direct byte	2	2	88-8F
MOV dir,dir	Move direct byte to direct byte	3	2	85
MOV dir,@Ri	Move indirect memory to direct byte	2	2	86-87
MOV dir,#data	Move immediate to direct byte	3	2	75
MOV @Ri,A	Move A to indirect memory	1	1	F6-F7
MOV @Ri,dir	Move direct byte to indirect memory	2	2	A6-A7
MOV @Ri,#data	Move immediate to indirect memory	2	1	76-77
MOV DPTR,#data	Move immediate to data pointer	3	2	90
MOVC A,@A+DPTR	Move code byte relative DPTR to A	1	2	93
MOVC A,@A+PC	Move code byte relative PC to A	1	2	83
MOVX A,@Ri	Move external data(A8) to A	1	2	E2-E3
MOVX A,@DPTR	Move external data(A16) to A	1	2	E0
MOVX @Ri,A	Move A to external data(A8)	1	2	F2-F3
MOVX @DPTR,A	Move A to external data(A16)	1	2	F0
PUSH dir	Push direct byte onto stack	2	2	C0
POP dir	Pop direct byte from stack	2	2	D0
XCH A,Rn	Exchange A and register	1	1	C8-CF
XCH A,dir	Exchange A and direct byte	2	1	C5
XCH A,@Ri	Exchange A and indirect memory	1	1	C6-C7
XCHD A,@Ri	Exchange A and indirect memory nibble	1	1	D6-D7

BOOLEAN				
Mnemonic	Description	Bytes	Cycles	Hex code
CLR C	Clear carry	1	1	C3
CLR bit	Clear direct bit	2	1	C2
SETB C	Set carry	1	1	D3
SETB bit	Set direct bit	2	1	D2
CPL C	Complement carry	1	1	B3
CPL bit	Complement direct bit	2	1	B2
ANL C,bit	AND direct bit to carry	2	2	82

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ANL C,/bit	AND direct bit inverse to carry	2	2	B0
ORL C,bit	OR direct bit to carry	2	2	72
ORL C,/bit	OR direct bit inverse to carry	2	2	A0
MOV C,bit	Move direct bit to carry	2	1	A2
MOV bit,C	Move carry to direct bit	2	2	92

BRANCHING				
Mnemonic	Description	Bytes	Cycles	Hex code
ACALL addr 11	Absolute jump to subroutine	2	2	11→F1
LCALL addr 16	Long jump to subroutine	3	2	12
RET	Return from subroutine	1	2	22
RETI	Return from interrupt	1	2	32
AJMP addr 11	Absolute jump unconditional	2	2	01→E1
LJMP addr 16	Long jump unconditional	3	2	02
SJMP rel	Short jump (relative address)	2	2	80
JC rel	Jump on carry = 1	2	2	40
JNC rel	Jump on carry = 0	2	2	50
JB bit,rel	Jump on direct bit = 1	3	2	20
JNB bit,rel	Jump on direct bit = 0	3	2	30
JBC bit,rel	Jump on direct bit = 1 and clear	3	2	10
JMP @A+DPTR	Jump indirect relative DPTR	1	2	73
JZ rel	Jump on accumulator = 0	2	2	60
JNZ rel	Jump on accumulator $\neq 0$	2	2	70
CJNE A,dir,rel	Compare A, direct jne relative	3	2	B5
CJNE A,#d,rel	Compare A, immediate jne relative	3	2	B4
CJNE Rn,#d,rel	Compare register, immediate jne relative	3	2	B8-BF
CJNE @Ri,#d,rel	Compare indirect, immediate jne relative	3	2	B6-B7
DJNZ Rn,rel	Decrement register, jnz relative	3	2	D8-DF
DJNZ dir,rel	Decrement direct byte, jnz relative	3	2	D5

	MISCELLANEOUS			
Mnemonic	Description	Bytes	Cycles	Hex code
NOP	No operation	1	1	00

ADDITIONAL INSTRUCTIONS (selected through EO[7:4])				
Mnemonic	Description	Bytes	Cycles	Hex code
MOVC @(DPTR++),A	M8051W/M8051EW-specific instruction supporting software download into program memory	1	2	A5
TRAP	Software break command	1	1	A5

In the above table, an entry such as E8-EF indicates a continuous block of hex opcodes used for 8 different registers, the register numbers of which are defined by the lowest three bits of the corresponding code. Non-continuous blocks of codes, shown as  $11 \rightarrow F1$  (for example), are used for absolute jumps and calls, with the top 3 bits of the code being used to store the top three bits of the destination address.

The CJNE instructions use the abbreviation #d for immediate data; other instructions use #data.s

## B. Instructions on how to use the input port.

Error occur status

- Using compare jump instructions with input port, it could cause error due to the timing conflict inside the MCU.
- Compare jump Instructions which cause potential error used with input port condition:

JB bit, rel ; jump on direct bit=1 JNB bit, rel ; jump on direct bit=0 JBC bit, rel ; jump on direct bit=1 and clear CJNE A, dir, rel ; compare A, direct jne relative DJNZ dir, rel ; decrement direct byte, jnz relative

- It is only related with Input port. Internal parameters, SFRs and output bit ports don't cause an y error by using compare jump instructions.
- If input signal is fixed, there is no error in using compare jump instructions.
- Error status example

while(1){
if (P00==1){ P10=1; }
else { P10=0; }
P11^=1;
}

zzz: JNB	080.0, xxx ;it possible to be error
SETB	088.0
SJMP	ууу
xxx: CLR	088.0
yyy: MOV	C,088.1
CPL	С
MOV	088.1,C
SJMP	ZZZ
L	

unsigned char ret_bit_err(void)
{
return <b>!P00</b> ;
}

MO	V R7, #000
JB	080.0, xxx ; it possible to be error
MOV	R7, #001
xxx: RET	

- Preventative measures (2 cases)
  - Do not use input bit port for bit operation but for byte operation. Using byte operation instead
    of bit operation will not cause any error in using compare jump instructions for input port.

	zzz: MOV	A, 080	; read as byte
	JNB 0	DE0.0, xxx	; compare
	SETB	088.0	
[]	SJMP	ууу	
while(1){	xxx: CLR	088.0	
if ((P0&0x01)==0x01){ P10=1; }	yyy: MOV	C,088.1	
else { P10=0; }	CPL (	С	
P11^=1;	MOV	088.1,C	
}	SJMP	ZZZ	

	zzz: MOV C,080.0 ; input port use i parameter	internal
	MOV 020.0, C ; move	
	JB 020.0, xxx ; compare	
bit tt;	SETB 088.0	
while(1){	SJMP yyy	
tt=P00;	xxx: CLR 088.0	
if (tt==0){ P10=1;}	yyy: MOV C,088.1	
else { P10=0;}	CPL C	
P11^=1;	MOV 088.1,C	
}	SJMP zzz	

 If you use input bit port for compare jump instruction, you have to copy the input port as intern al parameter or carry bit and then use compare jump instruction.



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