

LED Driver for Lighting Series Electrolytic Capacitor Free Buck Converter LED Driver



BD555A1AFV

General Description

BD555A1AFV is a LED driver best for LED lighting applications. It supports dimming. Constant current switching controller for AC/DC buck converter is accumulated inside. By choosing external MOS Transistor, small~large power of LED can be driven. The driver is adoptable to a wide range of lighting from small light such as spotlights to large one like base lights.

With digital power control, the average value of LED current is stably adjustable to requesting current. By only primary sense resistor, LED current is available to feedback LED current is useful without feedback circuits, reduce parts. Input characteristics and output characteristics is good by precisely digital power control.

It's support electrolytic free design by capacitor controller which is detect AC voltage and control ceramic capacitor.

Key Specifications

- AC Input range
- Input Voltage range(AUX pin)
- Operating temperature
- Accuracy of output current

Features

- Highly efficient Buck AC/DC converter
- Primary control without feedback circuit
- High accuracy LED current output by LED average control
- Capacitor controller for Electrolytic free design
- Fixed switching frequency
- Built-in regulator for inner power supply
- Built-in LED open detection (shutdown type)
- Built-in UVLO detection
- Built-in thermal shut-down function

Packages

SSOP-B14



Applications

- Spot light without dimming
- Desk ramp without dimming
- Down light without dimming
- Base light without dimming

Typical Application Circuits



80 to 275VAC

-40 to 100°C

±1.5 %(Typ.)

10 to 38V

●Absolute Maximum Ratings (Ta=25℃)

Parameter	Symbol	Ratings	Unit	Conditions
		4.5		VDD2, DVDD, DVDD2,
Maximum Applied Voltage 1	VMAX1	4.5	v	OSC, SOFT, ISENSE PIN
Maximum Applied Voltage 2	VMAX2	15.5	V	VDET, VDD1, VOUT,
Maximum Applied Voltage 2	V IVI/ (/\Z	10.0	v	CSEL, TESTO PIN
Maximum Applied Voltage 3	VMAX3	40.0	V	AUX PIN
Allowable Power Dissipation ^{*1}	Pd1	874.7	mW	
Operating Temperature	Topr	-40 ~ +100	°C	
Storage temperature	Tstg	-55 ~ +150	C°	

*1

When being mounted on 1 layer substrate(ROHM typical board). Copper layer area is 20.2mm². When being used at over Ta=25°C, Pd1 decreases by about 7.00mW/°C.

●Recommended Operating Range (Ta=-40℃ ~ +100℃)

Deremeter	Symbol		Ratings		Linit	Conditions
Parameter	Symbol	Min.	Тур.	Max.	Unit	Conditions
Operating supply voltage	AUX	10	16	38	V	

●Electrical Characteristics (Unless specified, AUX=16V, Ta = +25℃)

Deverseter	Current of	Ratings		Unit	Conditions		
Parameter	Symbol	Min.	Тур.	Max.	Unit	Conditions	
[Circuit Current]							
Operating Current 1	ldd1	7.0	14.0	22.0	μA	At Start-up, AUX=7.0V	
Operating Current 2	ldd2	1.5	2.5	3.5	mA	When stopping switching	
[Regulator]							
VDD1 Voltage	VVDD1	11	12	13	V	With no-load	
VDD1 Low Voltage Detection Voltage	Vuvlo1	4.0	5.0	6.0	V	When AUX is falling	
VDD2 Voltage	VVDD2	2.9	3.3	3.7	V	With no-load	
AUX Start-Up Voltage	Vstup	7.45	8.50	10.00	V	When AUX is rising	
AUX Low Voltage Detection Voltage	Vuvloaux	6.00	6.50	7.45	V	When AUX is falling	
[Switching Regulator]							
Oscillating Frequency	Fosc	90	100	110	kHz	Rosc=Open	
Maximum Duty	Dmax	-	75.0	-	%		
Average Current Sense Voltage	Visns	475	500	525	mV		
Current Sense Blank-Time	Tisns _t	360	400	440	nsec		
VOUT High On-Resistance	Rvouth	3.0	7.0	16.0	Ω	IVOUT=20mA Source	
VOUT Low On-Resistance	Rvoutl	2.0	4.5	12.0	Ω	IVOUT=20mA Sink	
[Capacitor Controller]							
AC Input "H" Detection Voltage	Deth	340	420	500	mV	When VDET is rising	
AC Input "L" Detection Voltage	Detl	240	300	360	mV	When VDET is falling	
VDET Input Voltage Range	VVDET	-0.3	-	9.0	V		
CSEL On-Resistance	RonCS	-	10	100	Ω	ICSEL=2mA Sink	
CSEL Off-Leakage	lleakCS	-	-	1.0	μA	CSEL=10V	

• Pin Description

No	端子名称	I/O	機能	等価回路
1	DVDD	In	Digital Power Supply	С
2	VDET	In	AC voltage detection PIN(for Electrolytic free)	А
3	AUX	In	Power Supply Input Pin	A
4	VDD1	Out	Regulator Output 1 / Inner Power Supply 12.0V	С
5	VDD2	Out	Regulator Output 2 / Inner Power Supply 3.3V	С
6	OSC	In	Switching Frequency Setting Pin	С
7	DVDD2	In	Digital Power Supply	С
8	SOFT	In	Soft-start · Slope Time Setting Pin	С
9	VOUT	Out	Switching MOS Gate Driver Pin	D
10	ISENSE	In	Current Sense Pin	С
11	GND	-	GND Pin	В
12	GNDA	-	GND Pin	D
13	CSEL	Out	Capacitor Selector Pin(for Electrolytic free)	А
14	TESTO	Out	Test Output Pin	С

Pin Configuration



• Equivalent Circuit Diagram



Fig.3 Equivalent Circuit Diagram

Block Diagram





Description of Blocks

1) Regulator

Regulator with output of typ.3.3V is built in, which places AUX pin at input. When AC power source is input, according to clamp voltage of Zener Diode (D_{AUX}), power is supplied to AUX pin through start-up resistor (R_{AUX}). Until AUX pin gets typ.8.5V, switching operation is kept stopped. Circuit current is typ.10µA, so that voltage drop of R_{AUX} can be kept small. When AUX pin becomes typ.8.5V, switching operation starts. After the start of switching operation, it is recommended that

power is supplied to AUX pin by fly-back method through transformer (L_{AUX}). Output voltage of VDD2 is used at circuit inner power supply. When VDD2 pin becomes below typ.2.45V, UVLO detection is carried out and it leads to a non-operating state.

2) Soft Start

Soft start bring good effect which is smoothing light on at power on and prevent the LED short protection in power on, if output capacitor is big.

Soft start limits Duty of switching and increases the set value of LED peak current from low level by slope function.

Soft start is run only one time when the AC power is supplied. When VDD2 pin detect low voltage, this function is reset and run again when next AC power source is input (when low voltage detection is canceled).



Soft start time can be set by resistor (R_{SOFT}) connecting to SOFT pin. Soft start time, T_{SOFT} is shown as in below chart.

Soft-Start/	
Slope Time	
100msec	
800msec	
1600msec	
0msec	
(No Slope)	

RSOFT	, Setting	of Sof	t-Start	∕Slope	Time

3) Digitally Controlled Switching Converter

Operation Description of Buck Converter

When AUX becomes typ.8.5V, switching controller starts its switching operation. Switching frequency is determined depending on the value of resistor (R_{osc}) connecting to OSC pin.

At the beginning of switching cycle, FET (TR1) for switching turns ON. Coil current flows into R_{ISET} and current is detected by ISENSE pin. However even TR1 turns ON, current detection is not carried out for a definite period of time, typ.400nsec. It prevents a malfunction caused by reverse recovery current of freewheel diode (D_{OUT}), when TR1turns ON. TR1 turns OFF, so that the voltage of ISENSE pin increases linearly and LED average current is adjusted to requested current.

Switching duty(D) is generally shown as below formula, considering input voltage as VBUCK and output voltage as VLED.

$$D = \frac{VLED}{VBUCK}$$

When switching duty (D) reaches to maximum 75.0% (typ.), the status is forced to change into OFF. Moreover minimum On width is 400nsec (typ.). For 400nsec, ON status is maintained compulsorily.

Setting of LED Current

Switching controller adjusts the average current of coil to a value which is set by R_{ISET}, monitoring coil current at switching "ON" from ISENSE pin. Since the average current of coil and LED current are of equivalent value, setting of LED current is determined by controlling the average current of coil.

LED average current, I_{AVE} is determined by ripple current ΔI_L of coil. Peak Current is shown as below formula in continuous current mode (CCM).

Compute ΔI_L from the data sampled with A/D converter from ISENSE pin, and calculate peak current I_{PEAK}, so that average current becomes requested value.

Input peak current I_{PEAK} into D/A converter and treat it as reference voltage of comparator COMP.

Since ΔI_{L} is calculated and feed-backed, average current control is available without influence of input voltage \cdot output voltage \cdot constant of coil \cdot switching frequency which are change by ΔI_{L}



Fig.6 Waveform of LED Current (in continuous current mode)



Fig.7 Block Diagram of Average Current Control

LED Average Current I_{AVE} is set by resistor (R_{ISET}) connecting ISENSE pin. Each set-up current is indicated as below formula. Setting current is chosen drive current of external MOS FET. Average current sense voltage, Visns is set to 500mV.

$$I_{AVE} = \frac{Visns}{R_{ISET}}$$

RISET, Set-Up Examples of LED Average Current

RISET	LED	
	Average Current	
5.10Ω	98mA	
2.00Ω	250mA	
1.50Ω	333mA	
1.00Ω	500mA	
0.75Ω	667mA	
0.68Ω	735mA	

Fig.8 shows input voltage characteristics when R_{ISET} is 1 Ω . And Fig.9 indicates output voltage characteristics.



Fig.8 Input voltage (VPRI) vs. LED Current



Under the following conditions, it is difficult to control average current.

1.) When On time is 400nsec.

On Time, $t_{\text{on,}}$ is indicated as follows, considering switching frequency as $f_{\text{SW}}.$

$$\mathbf{t}_{\mathsf{ON}} = \left(\frac{\mathsf{VLED}}{\mathsf{VBUCK}}\right) \times \frac{1}{\mathbf{f}_{\mathsf{SW}}}$$

When input voltage is high, output voltage is low or switching frequency is high, there is a possibility that on time becomes 400nsec.

Switching frequency is recommended to be set low.

When switching duty (D) is 75.0%.
Switching duty (D) is expressed as follows.

$$\mathsf{D} = \frac{\mathsf{VLED}}{\mathsf{VBUCK}}$$

When input voltage is low or output voltage is high, there is a possibility switching duty (D) exceeds 75.0%. It is necessary to keep input voltage high or make output voltage low.

3.) In case of DCM (discontinuous current mode)

In CCM, digitally controlled switching converter carries out average current control. In DCM, since current of section where no current flows cannot be controlled, LED current value becomes lower than set-up current value.



Fig.10 Waveform of LED Current (In discontinuous current mode)

DCM occurs when LED set-up current is low or switching frequency is low. Switching frequency is recommended to set high.

Setting of Switching Frequency

In setting of switching frequency, a relation of trade-off holds between power efficiency and the size (price) of external components. To improve power efficiency, slower switching frequency is better but in that case, the size of external parts get bigger.

Maximum switching frequency is decided by minimum ON time.

Minimum ON time ($t_{ON(MIN)}$) is available when input voltage VBUCK is highest voltage and output voltage VLED is minimum,. It is shown as below formula.

$$t_{\text{ON(MIN)}} = \left(\frac{\text{VLED}_{(\text{MIN})}}{\text{VBUCK}_{(\text{MAX})}}\right) \times \frac{1}{f_{\text{SW}}}$$

Switching duty (D) is simply shown as following formula.

$$D = \frac{VLED}{VBUCK} = t_{ON} \times f_{SW}$$

Set switching frequency within 20kHz \sim 300kHz. Determine it so that minimum ON time ($t_{ON(MIN)}$) gets more than 400nsec. Switching frequency f_{OSC} is determined by resistor R_{OSC} connecting to OSC pin.

Each set-up frequency is indicated as below formula. However if Rosc is not connected, it is set to 100 kHz. Moreover if Rosc is short-circuit to GND, it is set to 20kHz.

$$\mathbf{f}_{\mathsf{OSC}}[kHz] = 8192 / \operatorname{Rosc}[k\Omega]$$

	e el el la contening i requerie		
Rosc	Switching		
	Frequency		
300 kΩ	28kHz		
150 kΩ	50kHz		
68 kΩ	120kHz		
27 kΩ	300kHz		
No Connection	100kHz		

Rosc Setting Example of Switching Frequency

Protection Function

1) Detection of Abnormal Temperature

Thermal shut down starts to operate over 150°C(typ.), making IC from active status to non-active status. In the non-active status, switching operation stops. Gate driver output turns into low impedance against GND.

When the temperature of IC returns to normal level, IC recovers from initial mode to active mode.

2) VDD1 Low Voltage Detection

When VDD1 power supply voltage is low or any abnormal status occurs such as VDD1pin short, IC is turned to non-active status from active mode, since there is a possibility that IC does not operate properly. In non-active mode, switching operation stops. Gate driver output turns into low impedance against GND.

When VDD1power supply voltage returns to normal level, IC recovers from initial mode to active mode.

3) VDD2 Low Voltage Detection

When VDD2 power supply voltage is low or any abnormal problem occurs such as VDD2 pin short, IC is turned to non-active mode, since there is a possibility that IC does not operate properly. In non-active mode, switching operation stops. Gate driver output turns into low impedance against GND.

When VDD2power supply voltage returns to normal level, IC recovers from initial mode to active mode.

4) AUX Low Voltage Detection

When AUX power supply voltage is low or any abnormal problem occurs such as AUX pin short, IC is turned non-active, since there is a possibility that IC does not operate properly. In non-active mode, switching operation stops. Gate driver output turns into low impedance against GND.

When AUX power supply voltage returns to normal level, IC recovers from initial mode to active mode.

5) LED OPEN Detection

IC is turned from active mode to non-active mode, when ISENSE pin voltage does not reach to average current value for 52.4msec because of LED's OPEN defects, OPEN ISENSE pin, coil's OPEN defects and so on. In non-active mode, switching operation stops. Gate driver output turns into low impedance against GND.

LED OPEN detection holds non-active mode until VDD2 low voltage detection is carried out. IC recovers to active mode in following flow; switching operation stops \rightarrow VDD2 low voltage detection works \rightarrow LEDOPEN detection is re-set.

6) LED GND SHORT Detection

IC is turned from active mode to non-active mode, when ISENSE pin voltage does not reach to average current value for 52.4msec because of LED's GND SHORT defects, SHORT ISENSE pin. In non-active mode, switching operation stops. Gate driver output turns into low impedance against GND.

LED OPEN detection holds non-active mode until VDD2 low voltage detection is carried out. IC recovers to active mode in following flow; switching operation stops \rightarrow VDD2 low voltage detection works \rightarrow LED GND SHORT detection is re-set.

Typical Performance Curves



Fig.13 VDD2 Voltage

Fig.14 Switching Frequency



Fig.17 AC Input Voltage Characteristics

Fig.18 Output voltage Characteristics

Application Examples

1) Basic Circuit (Buck)

Connect about LEDs in series. Connect 1.5Ω resistor to R_{ISET} and set LED average current to 333mA. Select external MOS-FET depending on LED current. (recommendation : R5205CND:ROHM)

Set AUX power supply to over 12.0V. At start up, power is supplied through TR3. When AUX exceeds start-up voltage 8.5V, switching operation starts. In switching operation, power is supplied via transformer.

Connect VDET pin to VDD2. Connect DVDD and DVDD2 to VDD2, too. Switching frequency is set by connecting resistor between OSC pin and GND. Slope time of dimming can be set by connecting resistor between SOFT pin and GND.



Fig.19 Application Example (High-Power LED 333mA)

2) Design for more compact circuit, reducing number of components

Supply power through R_{AUX} . When AUX exceeds start-up voltage, 8.5V, switching operation starts. In steady state, set AUX voltage to over 12V. Since about 3mA power loss (0.3W: at AC100V) occurs at R_{AUX} , efficiency gets worse compared to the basic circuit.

Make OSC and SOFT pins Open. At this time, switching frequency becomes 100kHz and SOFT time is 0msec.



Fig.20 Application Example (Miniaturization : High-Power LED 333mA)

3) Circuit Example for Electrolytic free

Connect VDET and CSEL pin to the "Electrolytic Capacitor Less Circuit".

This application is not used Electrolytic-capacitor, is used two ceramic-capacitor. Two capacitor controls is available to stable constant LED current, when AC input voltage is low.



Fig.21 Application Example (Electrolytic free : High-Power LED 333mA)

Operational Notes

(1) Absolute Maximum Ratings

An excess in the absolute maximum ratings, such as supply voltage, temperature range of operating conditions, etc., can break down devices, thus making impossible to identify breaking mode such as a short circuit or an open circuit. If any special mode exceeding the absolute maximum ratings is assumed, consideration should be given to take physical safety measures including the use of fuses, etc.

(2) Operating conditions

These conditions represent a range within which characteristics can be provided approximately as expected. The electrical characteristics are guaranteed under the conditions of each parameter.

(3) Power supply line

Design PCB pattern to provide low impedance for the wiring between the power supply and the GND lines. In this regard, for the digital block power supply and the analog block power supply, even though these power supplies has the same level of potential, separate the power supply pattern for the digital block from that for the analog block, thus suppressing the diffraction of digital noises to the analog block power supply resulting from impedance common to the wiring patterns. For the GND line, give consideration to design the patterns in a similar manner.

Furthermore, for all power supply terminals to ICs, mount a capacitor between the power supply and the GND terminal. At the same time, in order to use an electrolytic capacitor, thoroughly check to be sure the characteristics of the capacitor to be used present no problem including the occurrence of capacity dropout at a low temperature, thus determining the constant.

(4) GND voltage

Make setting of the potential of the GND terminal so that it will be maintained at the minimum in any operating state. Furthermore, check to be sure no terminals are at a potential lower than the GND voltage including an actual electric transient.

(5) Short circuit between terminals and erroneous mounting

In order to mount ICs on a set PCB, pay thorough attention to the direction and offset of the ICs. Erroneous mounting can break down the ICs. Furthermore, if a short circuit occurs due to foreign matters entering between terminals or between the terminal and the power supply or the GND terminal, the ICs can break down.

(6) Operation in strong electromagnetic field

Be noted that using ICs in the strong electromagnetic field can malfunction them.

(7) Inspection with set PCB

On the inspection with the set PCB, if a capacitor is connected to a low-impedance IC terminal, the IC can suffer stress. Therefore, be sure to discharge from the set PCB by each process. Furthermore, in order to mount or dismount the set PCB to/from the jig for the inspection process, be sure to turn OFF the power supply and then mount the set PCB to the jig. After the completion of the inspection, be sure to turn OFF the power supply and then dismount it from the jig. In addition, for protection against static electricity, establish a ground for the assembly process and pay thorough attention to the transportation and the storage of the set PCB.

(8) Input terminals

In terms of the construction of IC, parasitic elements are inevitably formed in relation to potential. The operation of the parasitic element can cause interference with circuit operation, thus resulting in a malfunction and then breakdown of the input terminal. Therefore, pay thorough attention not to handle the input terminals, such as to apply to the input terminals a voltage lower than the GND respectively, so that any parasitic element will operate. Furthermore, do not apply a voltage to the input terminals when no power supply voltage is applied to the IC. In addition, even if the power supply voltage is applied, apply to the input terminals a voltage lower than the power supply voltage or within the guaranteed value of electrical characteristics.

(9) Ground wiring pattern

If small-signal GND and large-current GND are provided, It will be recommended to separate the large-current GND pattern from the small-signal GND pattern and establish a single ground at the reference point of the set PCB so that resistance to the wiring pattern and voltage fluctuations due to a large current will cause no fluctuations in voltages of the small-signal GND. Pay attention not to cause fluctuations in the GND wiring pattern of external parts as well.

(10) External capacitor

In order to use a ceramic capacitor as the external capacitor, determine the constant with consideration given to a degradation in the nominal capacitance due to DC bias and changes in the capacitance due to temperature, etc.

(11) Thermal shutdown circuit (TSD)

When junction temperatures become higher than detection temparatures, the thermal shutdown circuit operates and turns a switch OFF. The thermal shutdown circuit, which is aimed at isolating the LSI from thermal runaway as much as possible, is not aimed at the protection or guarantee of the LSI. Therefore, do not continuously use the LSI with this circuit operating or use the LSI assuming its operation.

(12) Thermal design

Perform thermal design in which there are adequate margins by taking into account the permissible dissipation (Pd) in actual states of use.

(13) Selection of coil

Select the low DCR inductors to decrease power loss for DC/DC converter.

(14) The temperature range of operation guarantees functional operation only. The life of LSI is not guaranteed in this range. The life of LSI has derating according to the environment, such as Ta, humidity, Voltage and so on. In performing an apparatus design, please perform the design in consideration of life derating of LSI.

(15) About the function description or technical note or more

The function description and the application notebook are the design materials to design a set. So, the contents of the materials aren't always guaranteed. Please design application by having fully examination and evaluation include the external elements.

Status of this document

The Japanese version of this document is formal specification. A customer may use this translation version only for a reference to help reading the formal version.

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Physical Dimension Tape and Reel Information

SSOP-B14



Marking Diagram



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